



US008253642B2

(12) **United States Patent**
Chang et al.

(10) **Patent No.:** **US 8,253,642 B2**
(45) **Date of Patent:** **Aug. 28, 2012**

(54) **LEAKY-WAVE ANTENNA CAPABLE OF MULTI-PLANE SCANNING**

7,868,829 B1 * 1/2011 Colburn et al. 343/700 MS
8,026,854 B2 * 9/2011 Sanada 343/700 MS
2010/0309079 A1 * 12/2010 Rofougaran et al. 343/777

(75) Inventors: **Rong-Yuan Chang**, Hsinchu (TW);
Fu-Chiang Chen, Hsinchu (TW)

FOREIGN PATENT DOCUMENTS

TW 465140 11/2001
TW M267648 6/2005

(73) Assignee: **National Chiao Tung University**,
Hsinchu (TW)

OTHER PUBLICATIONS

(*) Notice: Subject to any disclaimer, the term of this patent is extended or adjusted under 35 U.S.C. 154(b) by 358 days.

Nguyen et al., "Pencil-Beam Full-Space Scanning 2D CRLH Leaky-Wave Antenna Array," ISSSE, Montreal, 2007.

Lai et al., "Leaky-Wave Steering in a Two-Dimensional Metamaterial Structure Using Wave Interaction Excitation," Proc. Int. Microwave Symposium (IMS), 2006.

(21) Appl. No.: **12/757,058**

* cited by examiner

(22) Filed: **Apr. 9, 2010**

Primary Examiner — Tho G Phan

(65) **Prior Publication Data**
US 2011/0148727 A1 Jun. 23, 2011

(74) *Attorney, Agent, or Firm* — Jianq Chyun IP Office

(30) **Foreign Application Priority Data**
Dec. 23, 2009 (TW) 98144536 A

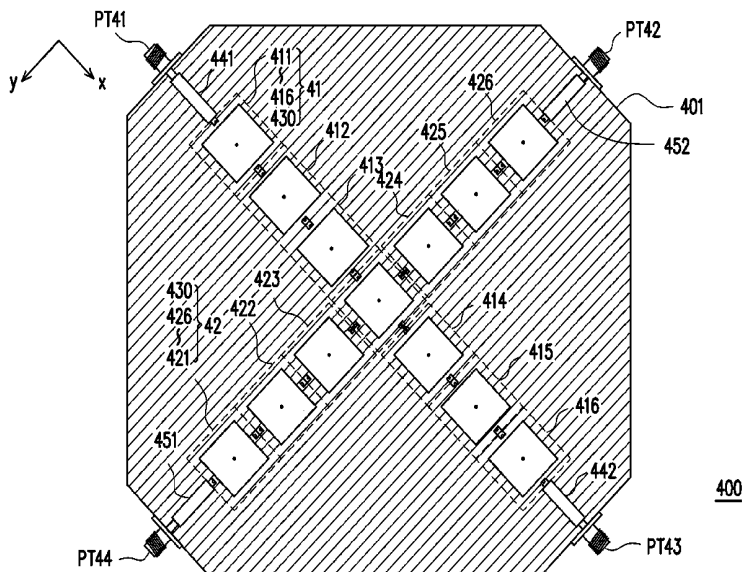
(57) **ABSTRACT**

A leaky-wave antenna capable of multi-plane scanning is provided. The leaky-wave antenna includes a substrate, a first antenna series, a second antenna series and a plurality of control units. The first antenna series intersects with the second antenna series to share a predetermined antenna unit among many antenna units. A part of the antenna units is connected in series to extend from a first and a second transmission lines of the predetermined antenna unit to compose the first antenna series, and the other antenna units are connected in series to extend from a third and a fourth transmission lines of the predetermined antenna unit to compose the second antenna series. The control units control the transmission paths between the first to the fourth transmission lines and the antenna units, and switch a leaky beam to different scanning planes, wherein the leaky beam scans with frequency variation through the antenna units.

(51) **Int. Cl.**
H01Q 13/20 (2006.01)
(52) **U.S. Cl.** 343/777; 343/700 MS; 343/756;
343/776
(58) **Field of Classification Search** 343/700,
343/756, 776, 777
See application file for complete search history.

(56) **References Cited**
U.S. PATENT DOCUMENTS
4,618,865 A * 10/1986 Lamensdorf et al. 343/776
6,317,095 B1 * 11/2001 Teshirogi et al. 343/785
7,253,780 B2 * 8/2007 Sievenpiper 343/745

12 Claims, 8 Drawing Sheets



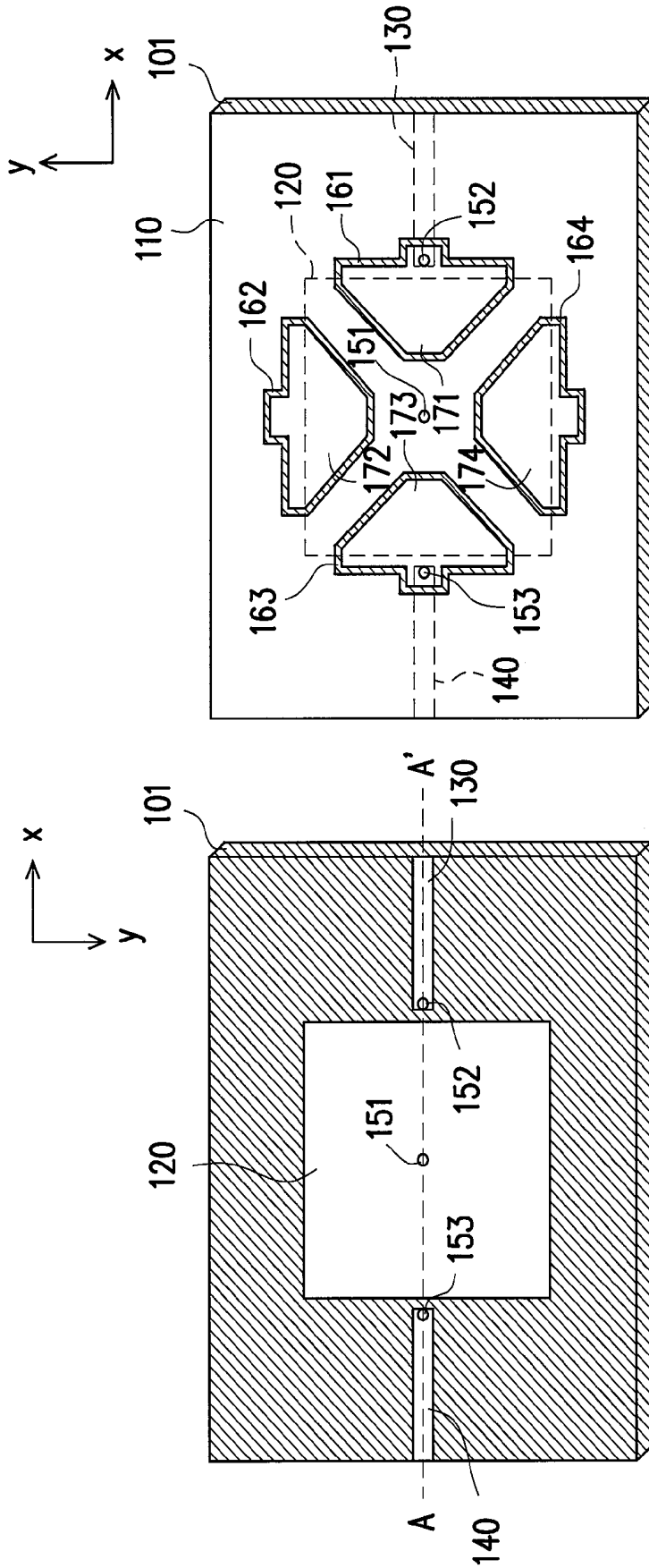


FIG. 1

100

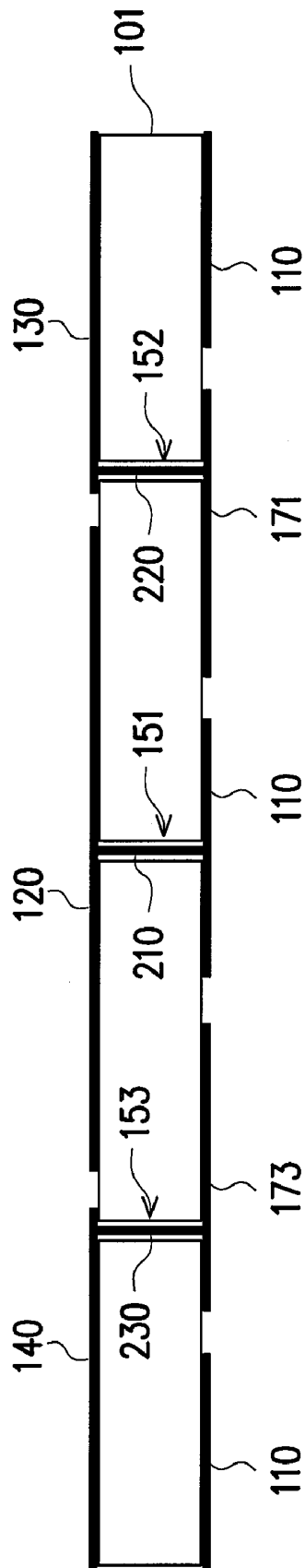


FIG. 2

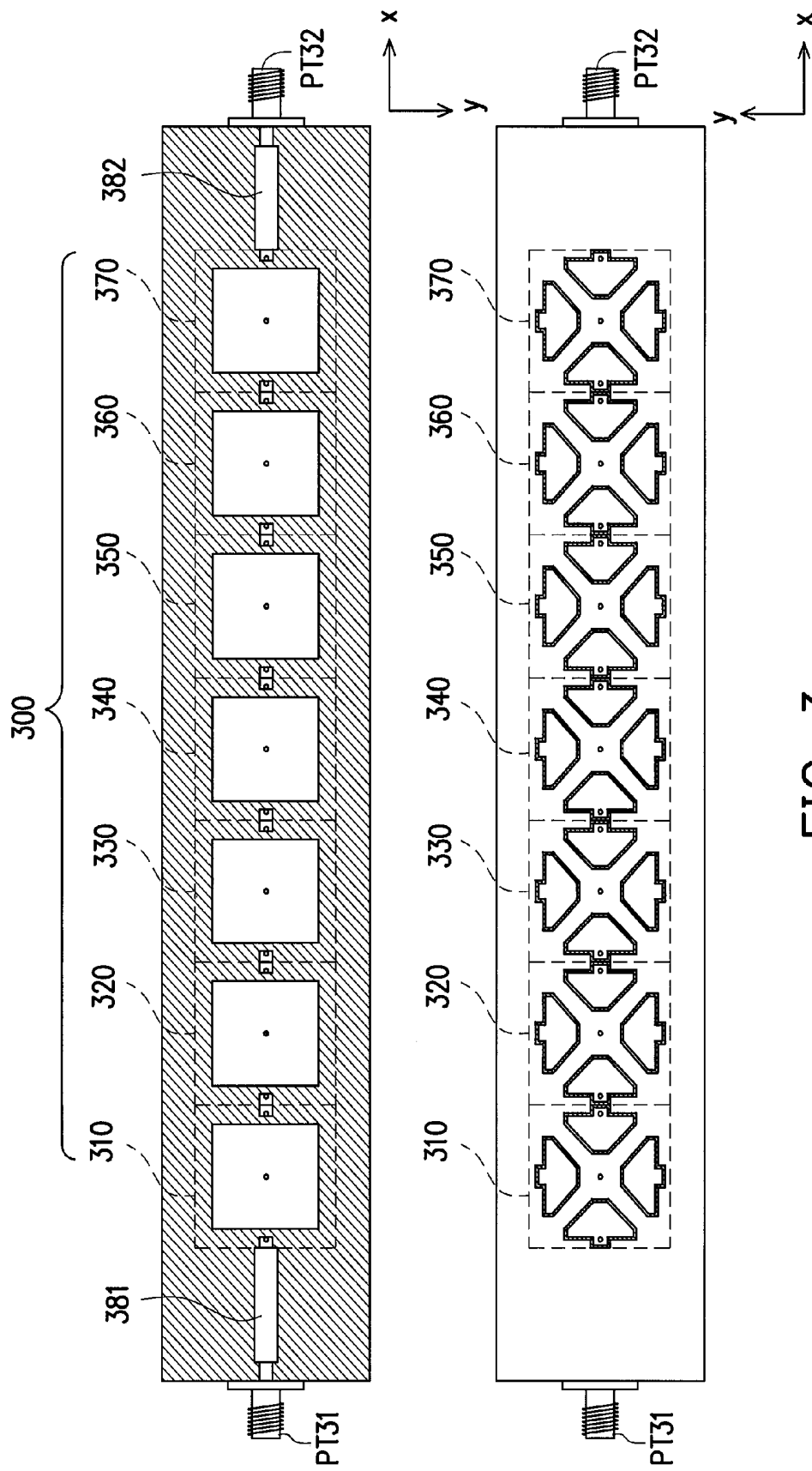


FIG. 3

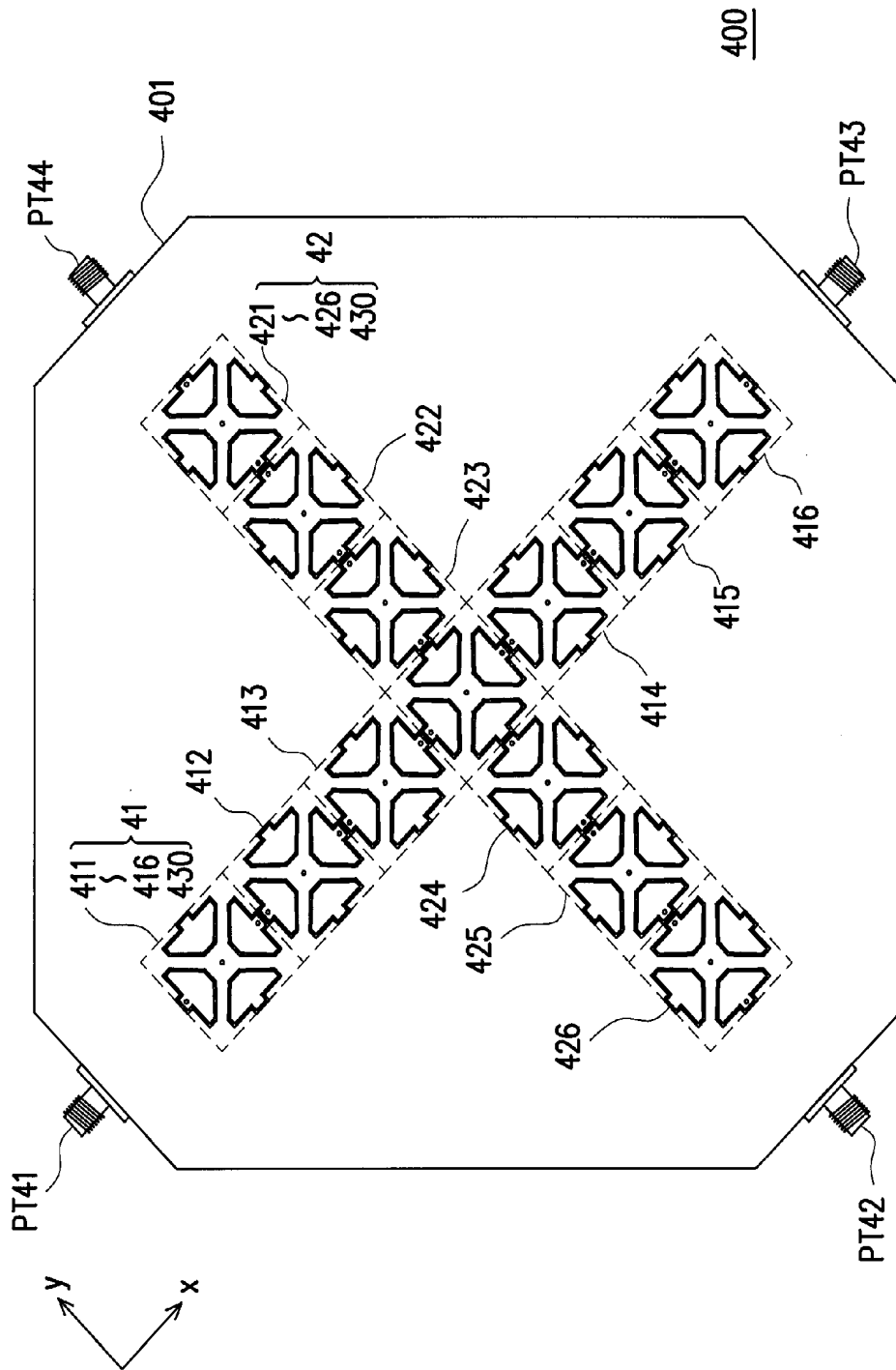


FIG. 5

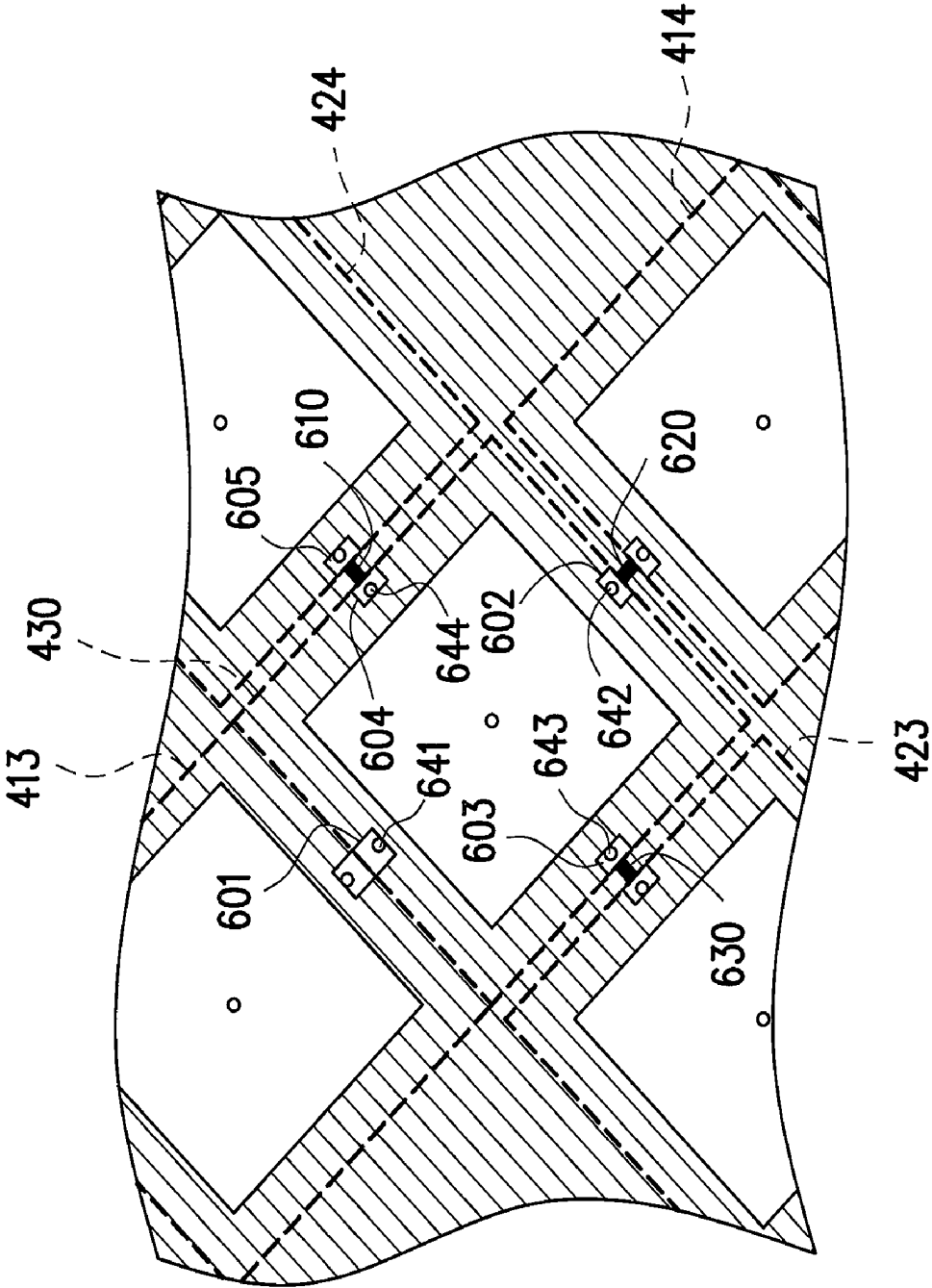


FIG. 6

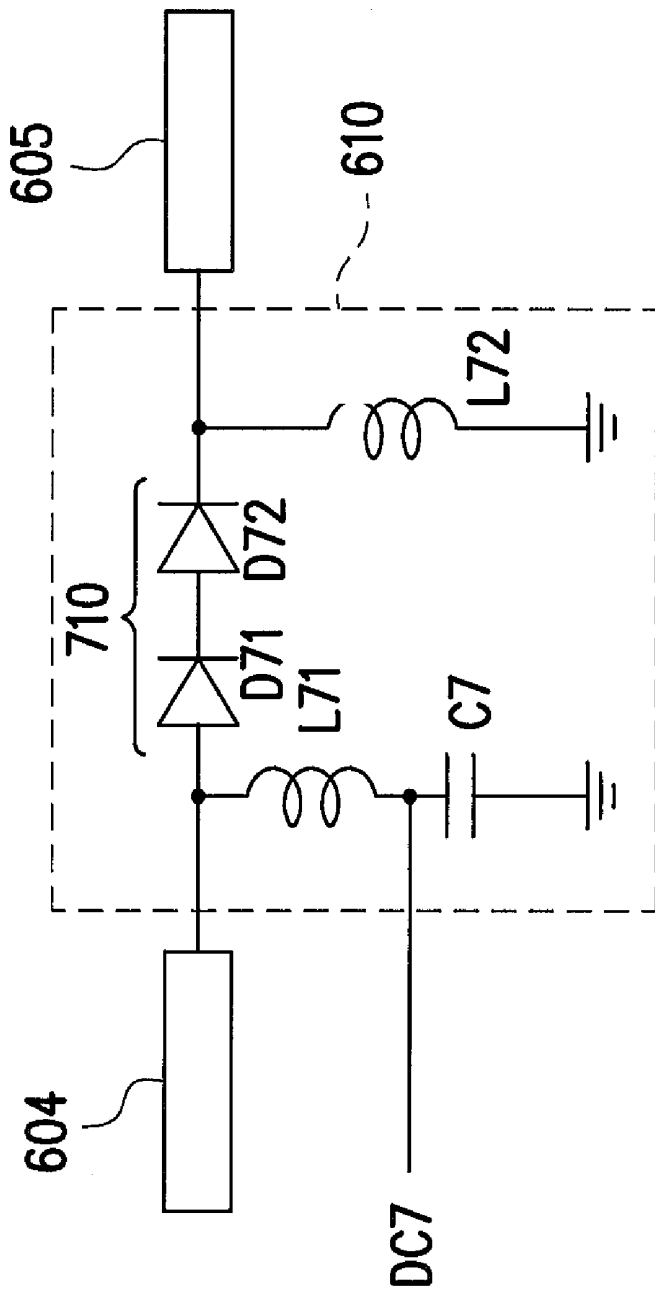


FIG. 7

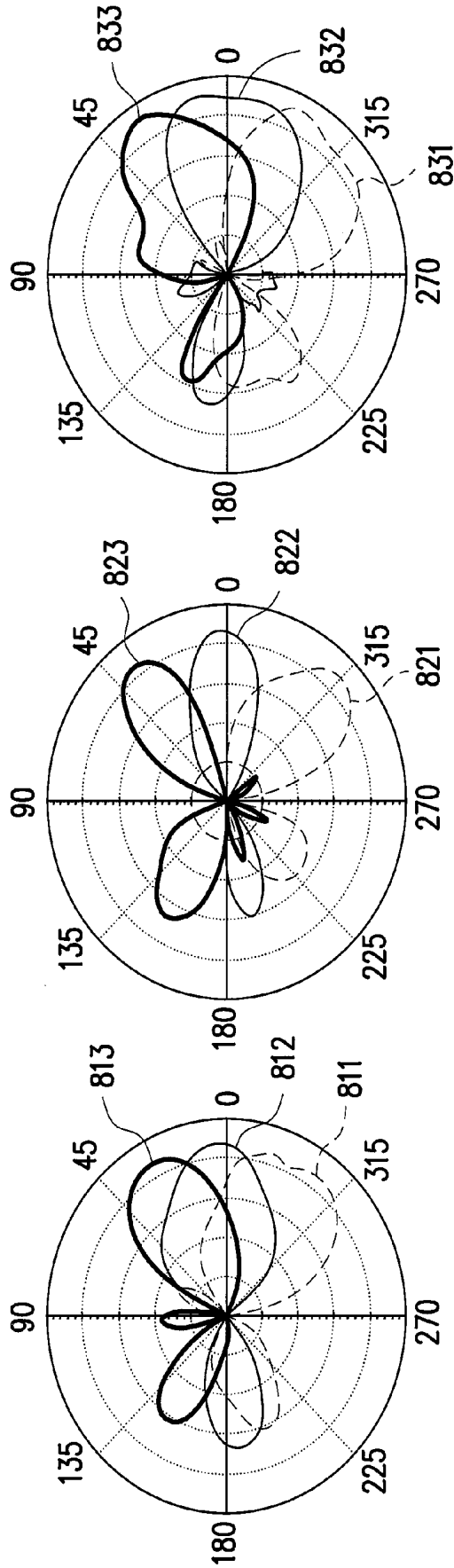


FIG. 8A

FIG. 8B

FIG. 8C

LEAKY-WAVE ANTENNA CAPABLE OF MULTI-PLANE SCANNING

CROSS-REFERENCE TO RELATED APPLICATION

This application claims the priority benefit of Taiwan application serial no. 98144536, filed on Dec. 23, 2009. The entirety of the above-mentioned patent application is hereby incorporated by reference herein and made a part of specification.

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to an antenna. More particularly, the present invention relates to a leaky-wave antenna capable of multi-plane scanning.

2. Description of Related Art

With booming development of wireless communication technology and liberalization of telecommunication, various communication protocol specifications and techniques are provided for achieving a better communication quality in an effective bandwidth. Moreover, since an antenna is one of indispensable elements in a wireless communication system, to design an antenna capable of improving a system performance is an important issue.

In a current communication system, the antenna generally has a characteristic of a wide beam pattern, for example, an omni directional and single beam. Generally, signals transmitted by an omni directional antenna and a directional antenna are liable to be influenced by multi-path fading and similar signals, so that a communication quality is influenced. To resolve the above problem, development of a smart antenna is one of the most promising technologies. Generally, the smart antennas can be group into switched beam antennas and scanning beam antennas. The switched beam antenna can change a beam shape and a beam direction of the antenna to increase an antenna gain and reduce the noise interference. The scanning beam antenna is implemented with assistant of active components or implemented by a leaky-wave antenna.

Presently, it is known that designs of the switched-beam antenna or the scanning beam antenna are approximately categorized into following types. The first type is to use a 90-degree coupler to feed into an antenna array, and different ports of the coupler are changed to serve as an input port, so as to achieve a beam switching effect. The second type is to design a Yagi antenna, and a PIN diode is added into a parasitic device, so that a length of the parasitic device is changed according to whether the PIN diode is conducted for serving as a reflection device or a guided-wave device, so as to achieve an effect of switching the beam direction. The third type is to use a Butler matrix to match an antenna array, and use a beam-forming technique for implementation.

Although the current techniques can achieve the beam switching effect, there is a plurality of shortages. For example, regarding the technique of changing different ports of the coupler to serve as the input port, the unused ports have to be connected with matched impedances for normal operation, which may lead to an operation inconvenience. Regarding the Yagi antenna designed according to a monopole antenna technique, etc., the antenna does not have a low profile characteristic, and slimness of the antenna cannot be implemented. Moreover, regarding the beam-forming technique, a complicated and large-area feed-in network and an antenna array have to be used to implement switching of the multiple beam directions, so that miniaturization thereof is

hard to be achieved. Moreover, none of the above methods can achieve a beam scanning function.

SUMMARY OF THE INVENTION

The present invention is directed to a leaky-wave antenna capable of multi-plane scanning, which has a beam scanning function and a beam switching function, and also has an advantage of miniaturization due to usage of a planar structural design.

The present invention provides a leaky-wave antenna capable of multi-plane scanning. The leaky-wave antenna includes a substrate, a first antenna series, a second antenna series and a plurality of control units. The first antenna series and the second antenna series are disposed on the substrate, and include a plurality of antenna units. Moreover, the first antenna series intersects with the second antenna series to share a predetermined antenna unit among the antenna units. A part of the antenna units are connected in series to extend outwards from a first and a second transmission lines of the predetermined antenna unit to form the first antenna series, and the other antenna units are connected in series to extend outwards from a third and a fourth transmission lines of the predetermined antenna unit to form the second antenna series. The control units are disposed at peripheral of the predetermined antenna unit for controlling transmission paths between the first to the fourth transmission lines and the antenna units, and switching a leaky beam to one of a plurality of scanning planes, wherein the leaky beam performs scanning along with a frequency variation through the antenna units.

In an embodiment of the present invention, the antenna units respectively comprise a metal ground layer, a metal sheet, a first conductive via, a fifth transmission line, a second conductive via, a sixth transmission line, and a third conductive via. The metal ground layer is disposed on a first surface of the substrate, and has a plurality of slots for dividing a plurality of metal blocks that are not electrically connected with each other. The metal sheet is disposed on a second surface of the substrate, and is partially overlapped to the metal blocks in a vertical projection plane. The first conductive via penetrates through the metal ground layer, the substrate and the metal sheet, and the metal sheet is electrically connected to the metal ground layer through a first conductive pole in the first conductive via.

Moreover, the fifth transmission line is disposed on the second surface of the substrate and located at a side of the metal sheet, and the fifth transmission line is partially overlapped to a first metal block of the metal blocks on the vertical projection plane. The second conductive via penetrates through the fifth transmission line, the substrate and the first metal block, and the fifth transmission line is electrically connected to the first metal block through a second conductive pole in the second conductive via. The sixth transmission line is disposed on the second surface of the substrate and located at another side of the metal sheet, and the sixth transmission line is partially overlapped to a second metal block of the metal blocks on the vertical projection plane. The third conductive via penetrates through the sixth transmission line, the substrate and the second metal block, and the sixth transmission line is electrically connected to the second metal block through a third conductive pole in the third conductive via.

In an embodiment of the present invention, the antenna units are respectively equivalent to a composite right/left-hand (CRLH) transmission line, and a balance frequency point of the CRLH transmission line relates to sizes of the

metal sheet, the first conductive pole, the second conductive pole and the metal blocks. Moreover, the metal sheet, the first conductive pole and the metal ground layer are equivalent to a left-hand inductor of the CRLH transmission line, and the fifth transmission line, the first metal block and the second conductive pole are equivalent to a left-hand capacitor of the CRLH transmission line.

According to the above descriptions, in the present invention, the antenna series having the beam scanning function are disposed in intersection, and the control units are used to control conduction states of the transmission paths. In this way, the leaky beam radiated by the leaky-wave antenna capable of multi-plane scanning is switched to one of the scanning planes, and an original sweep-frequency characteristic of the antenna series is maintained. Moreover, the leaky-wave antenna capable of multi-plane scanning has a planar structural design, so that it can be miniaturized. Since the antenna series are disposed in intersection, the circuit features of the leaky paths of the antenna are similar, so that usage of complicated matching circuits is unnecessary.

In order to make the aforementioned and other features and advantages of the present invention comprehensible, several exemplary embodiments accompanied with figures are described in detail below.

BRIEF DESCRIPTION OF THE DRAWINGS

The accompanying drawings are included to provide a further understanding of the invention, and are incorporated in and constitute a part of this specification. The drawings illustrate embodiments of the invention and, together with the description, serve to explain the principles of the invention.

FIG. 1 is a schematic diagram illustrating a structure of an antenna unit according to an embodiment of the present invention.

FIG. 2 is a cross-sectional view of an antenna unit of FIG. 1 along a line A-A'.

FIG. 3 is a structural schematic diagram illustrating an antenna series according to an embodiment of the present invention.

FIG. 4 and FIG. 5 are respectively a top view and a bottom view of a leaky-wave antenna capable of multi-plane scanning according to an embodiment of the present invention.

FIG. 6 is a partial enlarged diagram illustrating a leaky-wave antenna capable of multi-plane scanning of FIG. 5.

FIG. 7 is a circuit schematic diagram illustrating a control unit according to an embodiment of the present invention.

FIG. 8A is a measuring diagram of far-field radiation patterns of a 45-degree scanning plane of an antenna.

FIG. 8B is a measuring diagram of far-field radiation patterns of a 0-degree scanning plane of an antenna.

FIG. 8C is a measuring diagram of far-field radiation patterns of a -45-degree scanning plane of an antenna.

DESCRIPTION OF THE EMBODIMENTS

Reference will now be made in detail to the present preferred embodiments of the invention, examples of which are illustrated in the accompanying drawings. Wherever possible, the same reference numbers are used in the drawings and the description to refer to the same or like parts.

In the present invention, antenna series are disposed in intersection to form a leaky-wave antenna capable of multi-plane scanning, and each of the antenna series is formed by serially connecting a plurality of antenna units. Moreover, control units are disposed aside the intersection of the antenna series, so that a leaky beam radiated by the leaky-wave

antenna capable of multi-plane scanning can be switched to one of scanning planes. Moreover, each of the antenna units is designed to be a composite right/left-hand (CRLH) transmission line structure, so that the leaky beam can implement a sweep-frequency mechanism through the antenna units. To further convey the spirit of the present invention to those skilled in the art, a structure of the antenna unit is first described below, and the antenna series formed by serially connecting the antenna units, and the leaky-wave antenna capable of multi-plane scanning that is formed by the antenna series and the control units are described in succession.

FIG. 1 is a schematic diagram illustrating a structure of an antenna unit according to an embodiment of the present invention. Referring to FIG. 1, the antenna unit 100 has a planar structural design, which is disposed on a substrate 101. The substrate 101 has a first surface and a second surface. Moreover, the antenna unit 100 includes a metal ground layer 110, a metal sheet 120, a transmission line 130, a transmission line 140 and conductive vias 151-153.

Further, the metal ground layer 110 is disposed on the first surface of the substrate, and has a plurality of slots 161-164. The slots 161-164 expose the first surface of the substrate, and respectively form a closed loop. Therefore, the slots 161-164 can divide the metal ground layer 110 into a plurality of metal blocks 171-174 that are not electrically connected with each other. Moreover, the metal sheet 120, the transmission line 130 and the transmission line 140 are all disposed on the second surface of the substrate. For simplicity's sake, relative positions of the metal sheet 120, the transmission line 130 and the transmission line 140 that are projected on the first surface of the substrate are further illustrated by dot lines in FIG. 1.

As shown in FIG. 1, if the first surface of the substrate is regarded as a vertical projection plane, regarding physical configurations, the metal sheet 120 is partially overlapped to the metal blocks 171-174 on the vertical projection plane. Moreover, the metal blocks 171-174 are mutually symmetric relative to a geometric center of the metal sheet 120, and the conductive via 151 is overlapped to the geometric center of the metal sheet 120 on the vertical projection plane. On the other hand, the transmission line 130 is located at a side of the metal sheet 120, and is partially overlapped to the metal block 171 on the vertical projection plane. Moreover, the transmission line 140 is located at another side of the metal sheet 120, and is partially overlapped to the metal block 173 on the vertical projection plane.

FIG. 2 is a cross-sectional view of the antenna unit 100 along a line A-A'. Referring to FIG. 1 and FIG. 2, the conductive via 151 penetrates through the metal ground layer 110, the substrate 101 and the metal sheet 120. Therefore, the metal sheet 120 can be electrically connected to the metal ground layer 110 through a conductive pole 210 in the first conductive via 151. Moreover, the conductive via 152 penetrates through the transmission line 130, the substrate 101 and the metal block 171. Therefore, the transmission line 130 can be electrically connected to the metal block 171 through a conductive pole 220 in the conductive via 152. Moreover, the conductive via 153 penetrates through the transmission line 140, the substrate 101 and the metal block 173. Therefore, the transmission line 140 can be electrically connected to the metal block 173 through a conductive pole 230 in the conductive via 153.

It should be noticed that according to the above configurations, the antenna unit 100 is equivalent to a CRLH transmission line. The metal sheet 120, the conductive pole 210 and the metal ground layer 110 are equivalent to a left-hand inductor of the CRLH transmission line, and the transmission line 130, the metal block 171 and the conductive pole 220 are

equivalent to a left-hand capacitor of the CRLH transmission line. Comparatively, a balance frequency point of the CRLH transmission line is determined according to sizes of the metal sheet **120**, the conductive pole **210**, the conductive pole **220** and the metal blocks **171-174**.

In other words, a part of the areas of the metal ground layer **110** is hollowed by the slots **161-164**, and the metal blocks **171-174** divided by the slots **161-164** are respectively used to form one piece of metal sheet of a metal-insulator-metal (MIM) capacitor. Namely, in the present embodiment, a mushroom-like structure in a meta-material is combined with the MIM capacitor to form the left-hand inductor and the left-hand capacitor additionally required by the CRLH transmission line. Therefore, compared to a conventional MIM capacitor structure which requires an additional substrate to support a metal sheet thereof, in the present embodiment, only one substrate is used to implement the circuit structure of the CRLH transmission line, so that a low profile characteristic of the antenna is achieved, and the antenna is easy to be integrated with a planar printed circuit board.

Furthermore, the antenna units **100** can be connected in series to form an antenna series. For example, FIG. 3 is a structural schematic diagram illustrating an antenna series according to an embodiment of the present invention. Referring to FIG. 3, the antenna series **300** includes a plurality of antenna units **310-370**, wherein configurations of the antenna units **310-370** are the same to that of the antenna unit **100** of FIG. 1. Here, the antenna units **310-370** are respectively connected to the tandem antenna units in series through the internal transmission lines thereof, so as to form the antenna series **300**. Moreover, since a Bloch impedance of the CRLH transmission line is about 20 ohms, two ends of the antenna series **300** can be electrically connected to matching wires **381** and **382** with a quarter wavelength, so as to match impedances of feed-in ports **PT31** and **PT32**.

During an actual operation, when energy is transmitted to one end of the antenna series **300** from the feed-in port **PT31**, the other end of the antenna series **300** is electrically connected to a terminator of 50 ohms through the feed-in port **PT32**, so as to form the leaky-wave antenna structure. Moreover, a leaky beam radiated by the antenna series **300** can perform a continuous scanning along with a frequency variation, i.e. from a backward radiation formed at a low frequency left-hand leaky area to a forward radiation formed at a high frequency right-hand leaky area, which includes a broadside radiation scanned at the balance frequency point. Namely, when an operation frequency of the antenna series **300** is less than the balance frequency point, the antenna series **300** works at the left-hand leaky area and generates the backward radiation. When the operation frequency of the antenna series **300** is the balance frequency point, the antenna series **300** generates the broadside radiation. When the operation frequency of the antenna series **300** is greater than the balance frequency point, the antenna series **300** works at the right-hand leaky area and generates the forward radiation.

It should be noticed that the leaky areas of the antenna series **300** are operated in a fundamental mode rather than a high-order mode, so that a scanning angle and a radiation characteristic thereof are all better than that of a conventional leaky-wave antenna. Moreover, although the antenna series **300** of FIG. 3 is composed of 7 antenna units, a number of the antenna units used for forming the antenna series is not limited thereto. Those skilled in the art can arbitrarily modify the number of the antenna units according to actual design requirements, and a radiation gain and a directivity of the antenna series are correspondingly increased as the number of the antenna units is increased.

Further, two sets of the aforementioned antenna series can be disposed in intersection, and control units can be disposed aside the intersection of the antenna series to form the leaky-wave antenna capable of multi-plane scanning. For example, FIG. 4 and FIG. 5 are respectively a top view and a bottom view of the leaky-wave antenna capable of multi-plane scanning according to an embodiment of the present invention. FIG. 6 is a partial enlarged diagram illustrating the leaky-wave antenna capable of multi-plane scanning of FIG. 5. Referring to FIGS. 4-6, the leaky-wave antenna capable of multi-plane scanning **400** includes a substrate **401**, a first antenna series **41**, a second antenna series **42** and a plurality of control units **610-630**.

The first antenna series **41**, the second antenna series **42** and the control units **610-630** are all disposed on the substrate **401**, and the first antenna series **41** and the second antenna series **42** are formed by a plurality of antenna units **411-416**, **421-426** and **430**. Wherein, the first antenna series **41** and the second antenna series **42** are disposed in intersection for sharing the antenna unit **430**. In an actual structure, configurations of the antenna units **411-416** and **421-426** are the same as that of the antenna unit **100** of FIG. 1. A configuration of the antenna unit **430** is similar to that of the antenna unit **100** of FIG. 1, and only corresponding transmission lines and conductive vias are added to serially connect different antenna series.

As shown in FIG. 6, the antenna unit **430** includes transmission lines **601-604** and conductive vias **641-644** corresponding to the transmission lines **601-604**. The antenna units **411-416** are connected in series to extend outwards from the transmission lines **601** and **602** of the antenna unit **430** to form the first antenna series **41**, and the antenna units **421-426** are connected in series to extend outwards from the transmission lines **603** and **604** of the antenna unit **430** to form the second antenna series **42**. In other words, the first antenna series **41** is formed by the antenna units **411-416** and the antenna unit **430**, and the second antenna series **42** is formed by the antenna units **421-426** and the antenna unit **430**. Moreover, two ends of the antenna series **41** are electrically connected to matching wires **441** and **442** with a quarter wavelength, so as to match impedances of feed-in ports **PT41** and **PT43**. Comparatively, two ends of the antenna series **42** are electrically connected to matching wires **451** and **452** with a quarter wavelength, so as to match impedances of feed-in ports **PT42** and **PT44**.

Further, as shown in FIG. 6, the transmission lines **602-604** of the antenna unit **430** are electrically connected to the antenna units **414**, **423** and **424** through the control units **610-630**. Therefore, the leaky-wave antenna capable of multi-plane scanning **400** can control conduction states of transmission paths between the transmission lines **602-604** and the antenna units **414**, **423** and **424** through the control units **610-630**. FIG. 7 is a circuit schematic diagram illustrating a control unit according to an embodiment of the present invention. Referring to FIG. 7, taking the control unit **610** as an example, the control unit **610** includes a diode series **710**, an inductor **L71**, a capacitor **C7** and an inductor **L72**. Here, the diode series **710** is formed by serially connected diodes **D71** and **D72**. An anode of the diode series **710** is electrically connected to the transmission line **604** of the control unit **430**, and a cathode of the diode series **710** is electrically connected to a transmission line **605** of the control unit **424**. A first end of the inductor **L71** is electrically connected to the anode of the diode series **710**, and a second end of the inductor **L71** is used for receiving a direct current (DC) voltage **DC7**.

A first end of the capacitor **C7** is electrically connected to the second end of the inductor **L71**, and a second end of the

capacitor **C7** is electrically connected to the ground. A first end of the inductor **L72** is electrically connected to the cathode of the diode series **710**, and a second end of the inductor **L72** is electrically connected to the ground. During an actual operation, when the DC voltage **DC7** is switched to a positive voltage level, the diode series **710** is conducted, so that a transmission path between the transmission line **604** and the transmission line **605** is conducted. Comparatively, when the DC voltage **DC7** is switched to a negative voltage level, the diode series **710** is not conducted, so that the transmission path between the transmission line **604** and the transmission line **605** is not conducted. To avoid the DC voltage **DC7** influencing the operation of the antenna, the DC voltage **DC7** is isolated from the ground through the capacitor **C7**, and is transmitted to the diode series **710** through the inductor **L71**. Moreover, the inductor **L72** is used for conducting the DC voltage **DC7** to the ground.

beam performs continuous scanning along with the frequency variation in any of the scanning planes. For example, FIG. **8A** is a measuring diagram of far-field radiation patterns of the 45-degree scanning plane of the antenna. Wherein, curves **811-813** are respectively the radiation patterns when an operation frequency f of the antenna **400** is 2.26 GHz, 2.48 GHz and 2.88 GHz. Moreover, FIG. **8B** is a measuring diagram of far-field radiation patterns of the 0-degree scanning plane of the antenna. Wherein, curves **821-823** are respectively the radiation patterns when the operation frequency f of the antenna **400** is 2.26 GHz, 2.48 GHz and 2.88 GHz. Further, FIG. **8C** is a measuring diagram of far-field radiation patterns of the -45-degree scanning plane of the antenna. Wherein, curves **831-833** are respectively the radiation patterns when the operation frequency f of the antenna **400** is 2.26 GHz, 2.48 GHz and 2.88 GHz.

TABLE ONE

	45-degree scanning plane		0-degree scanning plane		-45-degree scanning plane	
	Beam direction	Antenna gain	Beam direction	Antenna gain	Beam direction	Antenna gain
$f = 2.26$ GHz	-39 degrees	3.99 dBi	-45 degrees	5.3 dBi	-29 degrees	3.97 dBi
$f = 2.48$ GHz	5 degrees	4.1 dBi	1 degree	4.96 dBi	5 degrees	4.02 dBi
$f = 2.88$ GHz	26 degrees	4.2 dBi	38 degrees	3.89 dBi	43 degrees	4.1 dBi

In this way, by switching the DC voltage level, the control units **610-630** can control the conducting states of the transmission paths between the transmission lines **602-604** and the antenna units **414, 423** and **424**, so that the leaky beam radiated by the leaky-wave antenna capable of multi-plane scanning **400** can be switched to one of a plurality of scanning planes. For example, when the control unit **610** conducts the transmission paths between the transmission line **604** and the antenna unit **424**, and the control units **620** and **630** maintains the respective transmission paths thereof in a non-conducting state, the energy of the antenna is transmitted from the feed-in port **PT41** to the feed-in port **PT42**. Now, the leaky-wave antenna capable of multi-plane scanning **400** can be regarded as an orthogonal-type leaky-wave antenna. Therefore, the leaky beam is synthesized by two orthogonal sub leaky beams, and an angle Φ of the scanning plane is about 45 degrees.

When the control unit **620** conducts the transmission paths between the transmission line **602** and the antenna unit **414**, and the control units **610** and **630** maintains the respective transmission paths thereof in the non-conducting state, the energy of the antenna is transmitted from the feed-in port **PT41** to the feed-in port **PT43**. Now, the leaky-wave antenna capable of multi-plane scanning **400** can be regarded as a one-dimensional leaky-wave antenna, and the angle Φ of the scanning plane is about 0 degree. On the other hand, when the control unit **630** conducts the transmission paths between the transmission line **603** and the antenna unit **423**, and the control units **610** and **620** maintains the respective transmission paths thereof in the non-conducting state, the energy of the antenna is transmitted from the feed-in port **PT41** to the feed-in port **PT44**, and the angle Φ of the scanning plane is about -45 degree.

In other words, the leaky-wave antenna capable of multi-plane scanning **400** can control the control units **610-630** to switch the leaky beam to one of the three scanning planes. On the other hand, according to the sweep-frequency characteristic of the control units **411-416, 421-426** and **430**, the leaky

Referring to FIGS. **8A-8C**, the characteristic of the leaky-wave antenna capable of multi-plane scanning **400** is shown in the table one. When the operation frequency f of the antenna **400** is 2.26 GHz, the antenna **400** works at the left-hand leaky area and generates the backward radiation. Moreover, the measured main beam directions of the antenna **400** in the three scanning planes are respectively -39 degrees, -45 degrees and -29 degrees, the measured maximum antenna gains are respectively 3.99 dBi, 5.3 dBi and 3.97 dBi, and the measured half power beam-width are respectively 58 degrees, 37 degrees and 61 degrees. When the operation frequency f of the antenna **400** is 2.48 GHz, the antenna **400** works at the balance frequency point and generates the broadside radiation. Moreover, the measured main beam directions of the antenna **400** in the three scanning planes are respectively 5 degrees, 1 degree and 5 degrees, the measured maximum antenna gains are respectively 4.1 dBi, 4.96 dBi and 4.02 dBi, and the measured half power beam-width are respectively 44 degrees, 30 degrees and 59 degrees.

Further, when the operation frequency f of the antenna **400** is 2.88 GHz, the antenna **400** works at the right-hand leaky area and generates the forward radiation. Moreover, the measured main beam directions of the antenna **400** in the three scanning planes are respectively 26 degrees, 38 degree and 34 degrees, the measured maximum antenna gains are respectively 4.2 dBi, 3.89 dBi and 4.1 dBi, and the measured half power beam-width are respectively 41 degrees, 26 degrees and 43 degrees. According another aspect, the antenna **400** can continuously scan for 65 degrees along with the frequency variation in the 45-degree scanning plane, and can continuously scan for 83 degrees along with the frequency variation in the 0-degree scanning plane, and can continuously scan for 63 degrees along with the frequency variation in the -45-degree scanning plane.

It should be noticed that in FIG. **4** and FIG. **5**, the metal sheets in the antenna units **411-416, 421-426** and **430** are squares, and the transmission lines **601-604** are respectively configured at four sides of the metal sheet of the middle

antenna unit **430**. Therefore, the antenna units **411-416** of the first antenna series **41** are connected in series along a first predetermined direction while taking the middle antenna unit **430** as a center, and the antenna units **421-426** of the second antenna series **42** are connected in series along a second predetermined direction while taking the middle antenna unit **430** as a center, wherein the first predetermined direction and the second predetermined direction are mutually perpendicular, so that the first antenna series **41** and the second antenna series **42** are intersected to form a cross structure, and generate the three scanning planes.

However, in an actual application, shapes of the metal sheets in the antenna units **411-416**, **421-426** and **430** can also be circular, hexagonal or octagonal. In case that the shape of the metal sheet is octagonal, the leaky-wave antenna capable of multi-plane scanning **400** further includes two additional sets of antenna series intersected with the antenna series **41** and **42**, so as to form a *-shape structure. Moreover, the leaky-wave antenna capable of multi-plane scanning **400** further includes four additional control units for controlling transmission paths formed at the intersection of the two added antenna series. In this way, the leaky-wave antenna capable of multi-plane scanning **400** can generate seven scanning planes. Deduced by analogy, the leaky-wave antenna capable of multi-plane scanning **400** may have a more omni-directional scanning function as the antenna series and the control units are increased.

In summary, in the present invention, the antenna series having the beam scanning function are disposed in intersection, and the control units are disposed at the intersection of the antenna series. In this way, the conduction states of the transmission paths provided by the control units can be controlled by switching the DC voltage level, so that the leaky beam radiated by the leaky-wave antenna capable of multi-plane scanning is switched to one of the scanning planes. Therefore, the leaky-wave antenna capable of multi-plane scanning may simultaneously have the beam scanning function and the beam switching function. Moreover, the leaky-wave antenna capable of multi-plane scanning has a planar structural design, so that it can be miniaturized, and since the antenna series are disposed in intersection, the circuit features of the leaky paths of the antenna are similar, so that usage of complicated matching circuits is unnecessary.

It will be apparent to those skilled in the art that various modifications and variations can be made to the structure of the present invention without departing from the scope or spirit of the invention. In view of the foregoing, it is intended that the present invention cover modifications and variations of this invention provided they fall within the scope of the following claims and their equivalents.

What is claimed is:

1. A leaky-wave antenna capable of multi-plane scanning, comprising:

a substrate;

a first antenna series and a second antenna series, disposed on the substrate, and comprising a plurality of antenna units, wherein the first antenna series intersects with the second antenna series to share a predetermined antenna unit among the antenna units, and a part of the antenna units are connected in series to extend outwards from a first and a second transmission lines of the predetermined antenna unit to form the first antenna series, and the other antenna units are connected in series to extend outwards from a third and a fourth transmission lines of the predetermined antenna unit to form the second antenna series; and

a plurality of control units, disposed at peripheral of the predetermined antenna unit for controlling transmission paths between the first to the fourth transmission lines and the antenna units, and switching a leaky beam to one of a plurality of scanning planes, wherein the leaky beam scans along with a frequency variation through the antenna units.

2. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 1, wherein the antenna units in the first antenna series are connected in series along a first predetermined direction while taking the predetermined antenna unit as a center, and the antenna units in the second antenna series are connected in series along a second predetermined direction while taking the predetermined antenna unit as a center.

3. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 2, wherein the first predetermined direction and the second predetermined direction are mutually perpendicular, so that the first antenna series and the second antenna series are intersected to form a cross structure.

4. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 1, wherein the antenna units respectively comprises:

a metal ground layer, disposed on a first surface of the substrate, and having a plurality of slots for dividing a plurality of metal blocks that are not electrically connected with each other;

a metal sheet, disposed on a second surface of the substrate, and being partially overlapped to the metal blocks in a vertical projection plane;

a first conductive via, penetrating through the metal ground layer, the substrate and the metal sheet, wherein the metal sheet is electrically connected to the metal ground layer through a first conductive pole in the first conductive via;

a fifth transmission line, disposed on the second surface of the substrate and located at a side of the metal sheet, wherein the fifth transmission line is partially overlapped to a first metal block of the metal blocks on the vertical projection plane;

a second conductive via, penetrating through the fifth transmission line, the substrate and the first metal block, wherein the fifth transmission line is electrically connected to the first metal block through a second conductive pole in the second conductive via;

a sixth transmission line, disposed on the second surface of the substrate and located at another side of the metal sheet, wherein the sixth transmission line is partially overlapped to a second metal block of the metal blocks on the vertical projection plane; and

a third conductive via, penetrating through the sixth transmission line, the substrate and the second metal block, wherein the sixth transmission line is electrically connected to the second metal block through a third conductive pole in the third conductive via.

5. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 4, wherein the metal blocks are mutually symmetric relative to a geometric center of the metal sheet.

6. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 4, wherein the first conductive via is overlapped to a geometric center of the metal sheet on the vertical projection plane.

7. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 4, wherein a shape of the metal sheet is rectangular, circular, hexagonal or octagonal.

11

8. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 4, wherein the antenna units are respectively equivalent to a composite right/left-hand (CRLH) transmission line, and a balance frequency point of the CRLH transmission line relates to sizes of the metal sheet, the first conductive pole, the second conductive pole and the metal blocks.

9. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 8, wherein the metal sheet, the first conductive pole and the metal ground layer are equivalent to a left-hand inductor of the CRLH transmission line.

10. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 8, wherein the fifth transmission line, the first metal block and the second conductive pole are equivalent to a left-hand capacitor of the CRLH transmission line.

11. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 1, wherein the control units respectively comprise:

12

a diode series, having an anode electrically connected to one of the first to the fourth transmission lines, and a cathode electrically connected to one of the antenna units;

a first inductor, having a first end electrically connected to the anode of the diode series;

a capacitor, having a first end electrically connected to a second end of the first inductor, and a second end electrically connected to ground; and

a second inductor, having a first end electrically connected to the cathode of the diode series, and a second end electrically connected to the ground.

12. The leaky-wave antenna capable of multi-plane scanning as claimed in claim 1, wherein the first antenna series and the second antenna series respectively comprise a first matching wire and a second matching wire, and the first matching wire and the second matching wire are electrically connected to two ends of the first antenna series and the second antenna series.

* * * * *