

Envelope ADI-FDTD Method and Its Application in Three-Dimensional Nonuniform Meshes

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Abstract—The envelope alternating-direction-implicit finite difference time domain (ADI-FDTD) method in 3-D nonuniform meshes was proposed and studied. The phase velocity error for the envelope ADI-FDTD and ADI-FDTD methods in uniform and nonuniform meshes and different temporal increments were studied. A cavity problem was studied using the envelope ADI-FDTD and ADI-FDTD methods in graded meshes and the conventional FDTD method in a uniform mesh. The simulation results show that the envelope ADI-FDTD performs better than the ADI-FDTD in numerical accuracy.

Index Terms—Courant-Friedrich-Levy (CFL) stability condition, envelope alternating-direction-implicit finite difference time domain (ADI-FDTD) method, phase velocity.

I. INTRODUCTION

THE idea of coupling the wave-envelope technique [1] with the ADI-FDTD method was proposed [2]–[4]. The numerical characteristic of the 3-D complex envelope ADI-FDTD method was derived in analytic form [5] and studied by solving a cavity problem [6]. However, the performance and efficiency of the envelope ADI-FDTD method with nonuniform mesh distribution has not been studied. In this letter, the numerical accuracy of the 3-D envelope ADI-FDTD method with uniform and nonuniform meshes were studied, and the simulation results of a cavity problem are shown.

II. METHOD

With the envelope ADI-FDTD method, the wave envelopes of the electromagnetic-field components are modeled, as opposed to the ADI-FDTD method, which simulates the actual electromagnetic-field components. Applying the wave envelope technique [1], the electromagnetic-field components are defined as

$$\vec{E}(x,y,z,t) = \vec{E}_a(x,y,z,t) \cdot e^{j\omega_c t}, \quad \vec{H}(x,y,z,t) = \vec{H}_a(x,y,z,t) \cdot e^{j\omega_c t} \quad (1)$$

where \vec{E}_a and \vec{H}_a are the wave envelopes and ω_c is the carrier angular frequency. Substituting (1) into Maxwell's curl equations and expanding in Cartesian coordinate into twelve differential equations. Following the FDTD method and ADI technique, the process of updating the electromagnetic-field components in a

time step in the FDTD method are divided into two half-time steps. These updating equations at each time step are shown as

$$\varepsilon_0(D_t + j\omega_c)E_{xay}^{N_1} + \sigma_y E_{xay}^{N_1} = D_y(H_{zax}^{N_2} + H_{zay}^{N_2}) \quad (2a)$$

$$\varepsilon_0(D_t + j\omega_c)E_{xaz}^{N_1} + \sigma_z E_{xaz}^{N_1} = -D_z(H_{yax}^{N_3} + H_{yaz}^{N_3}) \quad (2b)$$

$$\varepsilon_0(D_t + j\omega_c)E_{yaz}^{N_1} + \sigma_z E_{yaz}^{N_1} = D_z(H_{xay}^{N_2} + H_{xaz}^{N_2}) \quad (2c)$$

$$\varepsilon_0(D_t + j\omega_c)E_{yax}^{N_1} + \sigma_x E_{yax}^{N_1} = -D_x(H_{zax}^{N_3} + H_{zay}^{N_3}) \quad (2d)$$

$$\varepsilon_0(D_t + j\omega_c)E_{zax}^{N_1} + \sigma_x E_{zax}^{N_1} = D_x(H_{yax}^{N_2} + H_{yaz}^{N_2}) \quad (2e)$$

$$\varepsilon_0(D_t + j\omega_c)E_{zay}^{N_1} + \sigma_y E_{zay}^{N_1} = -D_y(H_{xay}^{N_3} + H_{xaz}^{N_3}) \quad (2f)$$

$$\mu_0(D_t + j\omega_c)H_{xay}^{N_1} + \sigma_y \eta_0^2 H_{xay}^{N_1} = -D_y(E_{zax}^{N_3} + E_{zay}^{N_3}) \quad (2g)$$

$$\mu_0(D_t + j\omega_c)H_{xaz}^{N_1} + \sigma_z \eta_0^2 H_{xaz}^{N_1} = D_z(E_{yax}^{N_2} + E_{yaz}^{N_2}) \quad (2h)$$

$$\mu_0(D_t + j\omega_c)H_{yaz}^{N_1} + \sigma_z \eta_0^2 H_{yaz}^{N_1} = -D_z(E_{xay}^{N_3} + E_{xaz}^{N_3}) \quad (2i)$$

$$\mu_0(D_t + j\omega_c)H_{yax}^{N_1} + \sigma_x \eta_0^2 H_{yax}^{N_1} = D_x(E_{zax}^{N_2} + E_{zay}^{N_2}) \quad (2j)$$

$$\mu_0(D_t + j\omega_c)H_{zax}^{N_1} + \sigma_x \eta_0^2 H_{zax}^{N_1} = -D_x(E_{yax}^{N_3} + E_{yaz}^{N_3}) \quad (2k)$$

$$\mu_0(D_t + j\omega_c)H_{zay}^{N_1} + \sigma_y \eta_0^2 H_{zay}^{N_1} = D_y(E_{xay}^{N_2} + E_{xaz}^{N_2}) \quad (2l)$$

where D_x , D_y , D_z , and D_t represent the first-order central finite-difference operators in x , y , z directions, and time domain, respectively; η_0 is the free-space wave impedance, σ_s is electric conductivity, and $s = x, y$, or z . $E_{xay}^{N_1}$ in (2a) means the E_{xay} component locates at the N_1 time step. From the n time step to the $n+1$ time step, N_1, N_2, N_3 in the first half-time step are $n+1/4, n+1/2, n$, respectively; N_1, N_2, N_3 in the second half-time step are $n+3/4, n+1/2, n+1$, respectively. The electromagnetic-field components are distributed in graded meshes as shown in Fig. 1(a), and the discrete forms of (2e) and (2j) in the first half-time step are shown below as examples:

$$\begin{aligned} E_{zax,(i,j,k+1/2)}^{n+1/2} &= ca_{x,(i,j,k+1/2)} E_{zax,(i,j,k+1/2)}^n + cb_{x,(i,j,k+1/2)} \\ &\times \left[H_{yax,(i+1/2,j,k+1/2)}^{n+1/2} + H_{yaz,(i+1/2,j,k+1/2)}^{n+1/2} \right. \\ &\quad \left. - H_{yax,(i-1/2,j,k+1/2)}^{n+1/2} - H_{yaz,(i-1/2,j,k+1/2)}^{n+1/2} \right] \quad (3a) \end{aligned}$$

$$\begin{aligned} H_{yax,(i+1/2,j,k+1/2)}^{n+1/2} &= da_{x,(i+1/2,j,k+1/2)} H_{yax,(i+1/2,j,k+1/2)}^n \\ &\quad + db_{x,(i+1/2,j,k+1/2)} \\ &\times \left[E_{zax,(i+1,j,k+1/2)}^{n+1/2} + E_{zay,(i+1,j,k+1/2)}^{n+1/2} \right. \\ &\quad \left. - E_{zax,(i,j,k+1/2)}^{n+1/2} - E_{zay,(i,j,k+1/2)}^{n+1/2} \right] \quad (3b) \end{aligned}$$

where the coefficients are written as

$$\begin{aligned} ca_{s,(i,j,k)} &= \frac{4\varepsilon_0 - \sigma_{s,(i,j,k)}\Delta t - j\varepsilon_0\omega_c\Delta t}{4\varepsilon_0 + \sigma_{s,(i,j,k)}\Delta t + j\varepsilon_0\omega_c\Delta t} \quad (4a) \end{aligned}$$

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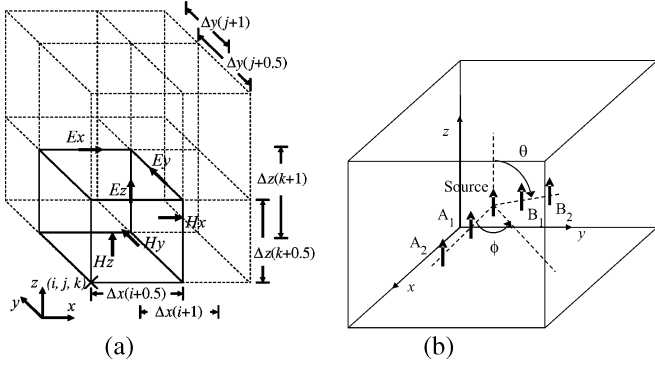


Fig. 1. In a 3-D domain: (a) a graded mesh and (b) a 3-D model used for studying the phase velocity error.

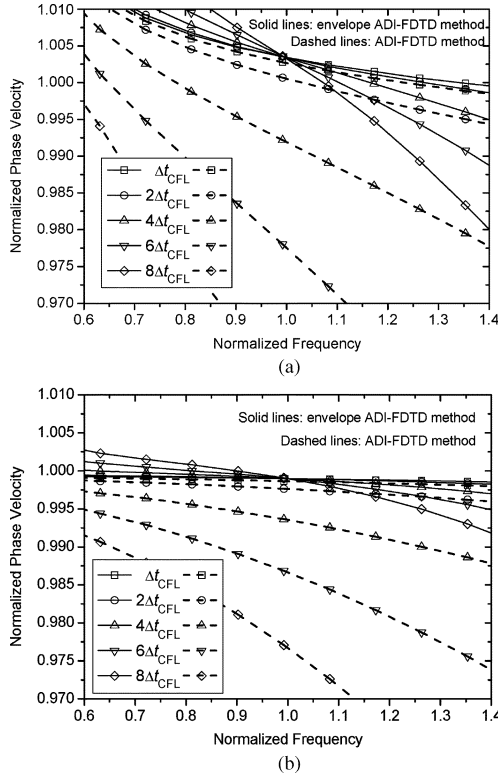


Fig. 2. Normalized phase velocity versus the normalized frequency. The cell size is $c_0/(40f_c)$. The propagation direction is (a) A_1 to A_2 and (b) B_1 to B_2 .

$$cb_{s,(i,j,k)} = 2\Delta t \cdot [\Delta s_{(i,j,k)} (4\epsilon_0 + \sigma_{s,(i,j,k)}\Delta t + j\epsilon_0\omega_c\Delta t)]^{-1} \quad (4b)$$

$$da_{s,(i,j,k)} = \frac{4\mu_0 - \sigma_{s,(i,j,k)}\eta_0^2\Delta t - j\mu_0\omega_c\Delta t}{4\mu_0 + \sigma_{s,(i,j,k)}\eta_0^2\Delta t + j\mu_0\omega_c\Delta t} \quad (4c)$$

$$db_{s,(i,j,k)} = 2\Delta t \cdot [\Delta s_{(i,j,k)} (4\mu_0 + \sigma_{s,(i,j,k)}\eta_0^2\Delta t + j\mu_0\omega_c\Delta t)]^{-1}. \quad (4d)$$

In these half-time steps, the implicit formulations can be solved with a tridiagonal matrix [7]. At the conclusion of the simulation, the wave envelope is modulated by the carrier to obtain the actual wave propagation.

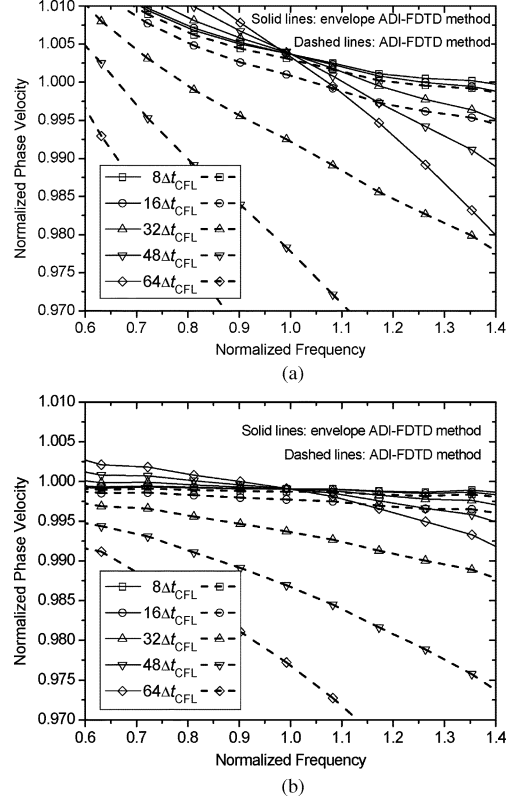


Fig. 3. Normalized phase velocity versus the normalized frequency. The maximum cell size is $c_0/(40f_c)$. The propagation direction is (a) A_1 to A_2 and (b) B_1 to B_2 .

III. RESULTS

A. Study of the Numerical Phase Velocity

The numerical phase velocity was studied by simulating the wave propagation in the free-space model as shown in Fig. 1(b). The dimension of the free space is $300 \times 300 \times 300$ cells. In order to obtain more accurate simulation results, 20 layers of the PML medium were imposed outside. A Gaussian pulse with bandwidth f_e of 6.35 GHz modulated with a carrier frequency f_c of 15 GHz was introduced at the source point which is located at (151, 151, 151) in free space. Points A_1 and A_2 , which are located at (231, 151, 151) and (251, 151, 151), respectively, are used to study the wave propagation along the x -axis ($\theta = 90^\circ$ and $\phi = 0^\circ$). Points B_1 and B_2 , which are located at (231, 231, 231) and (251, 251, 251), respectively, are used to model the oblique wave propagation in xyz domain ($\theta = 45^\circ$ and $\phi = 45^\circ$). At the end of the simulation, the numerical phase velocity can be computed [7].

In the following discussion, the free space is discretized in an uniform mesh (model 1) and a nonuniform mesh (model 2). In model 1, the cell size Δ is $c_0/(40f_c)$, and the maximum temporal increment under the CFL stability condition is $\Delta t_{CFL} = \Delta/(\sqrt{3}c_0)$. The model was simulated for the envelope ADI-FDTD and ADI-FDTD methods with different temporal increments, and the numerical phase velocity normalized by the speed of light versus frequency normalized by the carrier frequency are shown in Fig. 2. Here, the theoretical solution of the normalized phase velocity is 1. For the same temporal increment, we can find the numerical accuracy of the

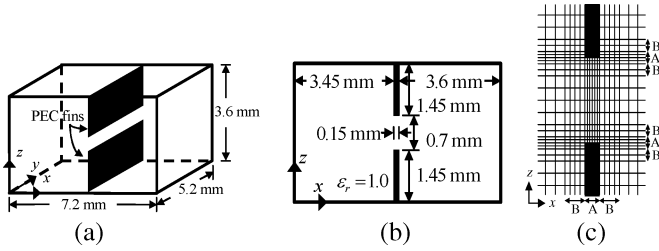


Fig. 4. A 3-D cavity model. (a) Bird's eye view. (b) Side view. (c) Under side view, the cell distribution near the PEC fins.

TABLE I
RESULTS OF A 3-D CAVITY IN FIG. 4

$\Delta t/\Delta t_{\text{CFL}}$	Resonant Frequency (GHz)			CPU time (s)
	1 st	2 nd		
envelope ADI-FDTD	8	51.096	55.296	101574.520
	16	51.096	55.296	52153.641
	24	51.096	55.296	33844.640
	32	51.139	55.296	25053.016
	40	51.139	55.296	20099.860
ADI-FDTD	48	51.144	55.258	16674.938
	8	51.052	55.166	38712.937
	16	50.922	54.863	19145.515
	24	50.703	54.427	12980.250
	32	50.359	53.867	9720.562
FDTD ⁽¹⁾	40	50.013	53.174	7786.750
	48	49.537	52.438	6483.531
FDTD ⁽²⁾	2	51.096	55.209	218253.122
	2	51.096	55.209	363966.750

diagonal propagation ($\theta = 45^\circ$ and $\phi = 45^\circ$) is better than that of the normal propagation ($\theta = 90^\circ$ and $\phi = 0^\circ$) in Fig. 2, and this result is the same as the theoretical analysis published in [5]. For the envelope ADI-FDTD method, the numerical error for all the temporal increments as shown in Fig. 2 is less than 2%, but the numerical error of the ADI-FDTD method is much larger than 2% when temporal increment is $8 \Delta t_{\text{CFL}}$ and normalized frequency is 1.0. In model 2, the distribution of a nonuniform mesh is

$$\begin{aligned} \Delta s(i) &= \Delta, & 1 \leq i \leq 237, & \quad 245 \leq i \leq 300 \\ \Delta s(i) &= \Delta/2, & i &= 238, 244 \\ \Delta s(i) &= \Delta/2^2, & i &= 239, 243 \\ \Delta s(i) &= \Delta/2^3, & 240 \leq i \leq 242 \end{aligned}$$

where $s = x, y,$ or z . Since the maximum cell size is $\Delta \times \Delta \times \Delta$ and the minimum cell size is $\Delta/8 \times \Delta/8 \times \Delta/8$, the Δt_{CFL} here is $1/8$ of Δt_{CFL} in model 1. This model was also modeled with the envelope ADI-FDTD and ADI-FDTD methods, and the simulation results are shown in Fig. 3. We can find each plot in Fig. 3 almost identical to the corresponding plots in Fig. 2, the envelope ADI-FDTD method also maintains good numerical accuracy in 3-D nonuniform mesh distribution.

B. Application of a 3-D Cavity Problem

The specifications of the structure are shown in Fig. 4. The thicknesses of the two fins are 0.15 mm. A Gaussian pulse with bandwidth 8 GHz modulated with a carrier frequency of 54 GHz was introduced, and this problem was solved by using

the ADI-FDTD, the envelope ADI-FDTD, and the traditional FDTD methods. The mesh size in y direction is uniform at 0.1 mm and in x, z directions are nonuniform. The mesh distribution near the PEC fin is shown in Fig. 4(c), where the definition of the mesh size in x, z directions are 0.025 mm in segment A, 0.050 mm in segment B, and 0.100 mm in other mesh. This problem was solved by the ADI-FDTD and the envelope ADI-FDTD methods in nonuniform mesh, and was also solved by the conventional FDTD method in uniform mesh where the cell size was $0.05 \times 0.05 \times 0.05 \text{ mm}^3$. Here, Δt_{CFL} is set as $(0.025 \times 10^{-3} \text{ m})/(c_0 \sqrt{3})$. The simulation time in the ADI-FDTD and the envelope ADI-FDTD methods is 7.698 ns, and that in the FDTD method is 7.698 ns⁽¹⁾ and 15.396 ns⁽²⁾. We recorded the values of E_z field at the observation point, then padded zeros of twice the simulation time were added, and applied the discrete Fourier transform to compute the resonant frequencies of the cavity. The resonant frequencies and CPU time for all three numerical methods are listed in Table I. Since the cavity was discretized by a uniform fine mesh in the FDTD method, we took the results of the FDTD method as the reference. For the first resonant frequency, it is found that the relative error of the envelope ADI-FDTD method is always less than that of the ADI-FDTD method with the same CPU time. The envelope ADI-FDTD method also takes less CPU time and obtains the same value of the first resonant frequency which was solved by the FDTD method. Besides, the numerical error of the ADI-FDTD method increases with the temporal increment, but the envelope ADI-FDTD method always maintains good performance even when the temporal increment is large.

IV. CONCLUSION

The numerical accuracy and efficiency of the 3-D envelope ADI-FDTD method were presented in this letter. Compared with the ADI-FDTD method, the numerical phase velocity for the envelope ADI-FDTD method is much more accurate than that of the ADI-FDTD method. The numerical solutions of a 3-D cavity problem show that the efficiency of the envelope ADI-FDTD method is better than that of the ADI-FDTD method.

REFERENCES

- [1] C. T. M. Choi and J. P. Webb, "Wave-envelope and transformation methods for finite element solution of unbounded electromagnetic wave problems," *IEEE Trans. Magn.*, vol. 32, no. 3, pp. 886–889, May 1996.
- [2] H. Rao, R. Scarmozzino, and R. M. Osgood, "An improved ADI-FDTD method and its application to photonic simulations," *IEEE Photon. Technol. Lett.*, vol. 14, no. 4, pp. 477–479, Apr. 2002.
- [3] C. T. M. Choi and S. H. Sun, "Numerical performance and applications of the envelope ADI-FDTD method," *IEEE Trans. Microw. Theory Tech.*, vol. 54, no. 1, pp. 256–264, Jan. 2006.
- [4] S. H. Sun and C. T. M. Choi, "Performance of the improved PML for the envelope ADI-FDTD method in two-dimensional domain," *IEEE Microw. Wireless Compon. Lett.*, vol. 15, no. 11, pp. 800–802, Nov. 2005.
- [5] C. Ma and Z. Chen, "Dispersion analysis of the three-dimensional complex envelope ADI-FDTD method," *IEEE Trans. Microw. Theory Tech.*, vol. 53, no. 3, pp. 971–976, Mar. 2005.
- [6] —, "The complex envelope (CE) ADI-FDTD method," in *IEEE MTT-S Int. Dig.*, Jun/2005, pp. 1119–1122.
- [7] T. Namiki and K. Ito, "Investigation of numerical errors of the two-dimensional ADI-FDTD method," *IEEE Trans. Microw. Theory Tech.*, vol. 48, no. 11, pp. 1950–1956, Nov. 2000.