# Very High Density RF MIM Capacitor Compatible with VLSI

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*Abstract* **— We have fabricated RF MIM capacitors, using high-**κ **TiTaO as the dielectric, which show a record high density of 20 fF/**µ**m2 . In addition, the capacitors display a small capacitance reduction of only 3.6% over the 100 kHz to 20 GHz range, a low leakage current of 8 pA and a high Q of 120. This was for a typical large 8 pF TiTaO MIM capacitor. The small voltage dependence of the capacitance (**∆*C/C***) of 770 ppm at 2 GHz, shows that these MIM capacitors are useful for highprecision RF circuits.** 

*Index Terms* **— RF, capacitor, MIM***,* **TiTaO.**

## I. INTRODUCTION

To scale down the die size and cost of a RF IC, the areas of both the active and passive devices need to be reduced. Although the area of MOSFETs scale down by  $\sim$ 50% every 18-24 months, the passive RF devices do so at a much slower rate and occupy an increasing percentage of the IC area. Of all the passive devices the RF MIM capacitors [1]-[3] are the most frequently needed in an RF IC, being used for impedance matching, DC blocking and filtering. Therefore, to reduce the chip size and its cost, scaling down the MIM capacitor area is necessary. The practical way of achieving this is to increase the capacitance density  $(\varepsilon_0 \kappa / t_d)$  by using a high- $\kappa$  dielectric [4]-[7]. Decreasing the dielectric thickness  $(t_d)$  increases the leakage current, loss tangent and the voltage dependence of capacitance (∆*C/C*), all of which is undesirable. The planned approach is to evolve the high-κ dielectric in RF MIM capacitors from SiON ( $\kappa$ ~4-7) to Al<sub>2</sub>O<sub>3</sub> ( $\kappa$ =10) to HfO<sub>2</sub>  $(\kappa \sim 22)$  or Ta<sub>2</sub>O<sub>5</sub> ( $\kappa \sim 25$ ), according to the International Technology Roadmap for Semiconductors (ITRS). However, no practical high-κ dielectric for RF MIM capacitors has a κ value above 25. This is because increasing the  $\kappa$  value decreases the stability of the high κ dielectric, leading to unacceptably high leakage current. For example, although TiO<sub>2</sub> has a dielectric constant of  $\sim 80$ , TiO<sub>2</sub> RF MIM capacitors suffer from intolerably high leakage current, after the 400°C annealing needed for VLSI back-end process integration [4]-[7].

In this paper we report on record high-density (20  $fF/\mu m^2$ ) RF MIM capacitors, using TiTaO as the dielectric, where the TiTaO has a high- $\kappa$  value of  $\sim$ 45, which is greater

than the κ=25 barrier mentioned above. The motivation for using TiTaO is to take advantage of the high-κ property of the  $TiO<sub>2</sub>$ , but to reduce the DC leakage by using a mixed dielectric, similar to our previous AlTaO study [5]. We have obtained good RF performance, in terms of a small reduction  $(< 5\%)$  in the capacitor value, in the range 100 KHz to 20 GHz. Furthermore, ∆C/C was 770 ppm (representing a quadratic voltage coefficient of only 130 ppm/ $V^2$ ) in the RF frequency regime. Since these devices are fully processcompatible with current VLSI lines, our approach can achieve a potential reduction of the RF capacitor area of  $\sim$ 20 X, compared with technology currently available in foundries.

#### II. EXPERIMENTAL PROCEDURE

A thick  $SiO<sub>2</sub>$  isolation layer was deposited on standard Si wafers in a VLSI back-end process. The lower MIM capacitor electrode was then fabricated on the  $SiO<sub>2</sub>/Si$  substrates using a sputter-deposited TaN/Ta bi-layer and then patterning. The thick Ta in the bi-layer structure is needed to reduce the RF ohmic loss and the thin TaN acts as a barrier between the high-κ dielectric and the Ta electrode. Then a thin  $Ti<sub>x</sub>Ta<sub>1-x</sub>O$ (x~0.6) dielectric layer was deposited on the lower TaN/Ta electrode, followed by a 400°C oxidation and annealing step. This 400°C low-temperature thermal cycle is necessary for VLSI backend process integration. For comparison, we also fabricated capacitors using only  $TiO<sub>2</sub>$  as the dielectric. Finally, Al was deposited and patterned to form both the top capacitor electrode and for the RF transmission lines.

The fabricated RF MIM capacitors were characterized using an HP4284A precision LCR meter from 10 KHz to 1 MHz, and an HP8510C network analyzer was used to measure the S-parameters from 200 MHz to 20 GHz (followed by a standard de-embedding procedure from a dummy open device [8]-[9]). The series inductance was also de-embedded with a through transmission line. The RF frequency capacitance was extracted from the measured S-parameters using an equivalent circuit model [5]. The RF frequency ∆*C/C* and quadratic voltage coefficient were then calculated using circuit theory, from the measured S-parameter data.

# III. RESULTS AND DISCUSSION

## *A. Low frequencies characteristics*

Fig.1 shows the  $C-V$  characteristics for  $TiO<sub>2</sub>$  and  $TiTaO$ MIM capacitors, where record high capacitance densities of 21.5 and 20  $fF/\mu m^2$  respectively, were found at 1 MHz. Current MIM capacitors provided by foundries have only  $\sim$ 1  $fF/\mu m^2$ . Hence the TiTaO capacitor provides a 20 times area reduction compared with current technology. The  $TiO<sub>2</sub>$  MIM capacitor suffers from a large capacitance variation at voltages above 0.75 V. In contrast, constant capacitance values with little voltage and frequency dependence are shown by the TiTaO MIM capacitor – this is important for analog/RF circuit applications. TiTaO shows a high- $\kappa$  value of  $\sim$ 45 – this is beyond the  $\kappa$ ~22-25 range displayed by HfO<sub>2</sub> and Ta<sub>2</sub>O<sub>5</sub>, used in current DRAM manufacturing processes (Fig. 2).



Fig.1 The  $C-V$  characteristics for  $TiO<sub>2</sub>$  and  $TiTaO$  MIM capacitors measured at different frequencies.



Fig.2 The evolution of high-κ values for Analog/RF MIM and DRAM capacitors. TiTaO provides twice the  $\kappa$  value than HfO<sub>2</sub>.

To investigate the abnormal large capacitance variation in  $TiO<sub>2</sub>$  capacitors at applied voltage > 0.75 V, we have measured the DC Current-Density and Voltage (*J-V*) characteristics. As shown in Fig. 3, the TiO<sub>2</sub> capacitor leakage current increases rapidly with increasing voltage, giving rise to the abnormal capacitance increase in the *C-V* curve in Fig. 1. This large leakage current is related to the poly-crystallization of the TiO<sub>2</sub> from its amorphous state, after the 400°C O<sub>2</sub> annealing, which generates a high current conduction path through poly-grain boundaries. In contrast the TiTaO MIM capacitors provide  $\sim 6-7$  orders of magnitude lower leakage current than do the  $TiO<sub>2</sub>$  ones, and a low leakage current of  $2.0 \times 10^{-6}$  A/cm<sup>2</sup> is measured at 1V. The leakage current for a capacitor (*C*) can be obtained from  $J \times C/C_a$ , where *J* is shown in Fig. 3 and  $C_d$  is the capacitor density (20 fF/ $\mu$ m<sup>2</sup>) from Fig. 1. For the relatively large 8 pF capacitor used in this work, the leakage current is as low as 8 pA, due to the very small area with dimensions of 20  $\mu$ m × 20  $\mu$ m.



Fig. 3. The  $J-V$  characteristics for TiO<sub>2</sub> and TiTaO MIM capacitors, after the 400°C  $O_2$  anneal for 10 min. The large leakage current in the  $TiO<sub>2</sub>$  device is due to the formation of poly-crystalline grains. The leakage current is 6-7 orders of magnitude lower in the TiTaO MIM capacitors.

#### *B. High frequencies characteristics*

We have measured the S-parameters to investigate the RF frequency characteristics beyond the 1 MHz upper limit of precision LCR meter. Fig. 4 shows the S-parameters for a TiTaO MIM capacitor. The RF frequency capacitance can be extracted from the S-parameters using the equivalent-circuit model shown in Fig. 5 [6]. The *C*,  $R_P$ ,  $L_S$ , and  $R_S$  in this model are the capacitor, parallel resistance, parasitic series inductance, and parasitic series resistance, respectively, where the  $R<sub>S</sub>$  and  $L<sub>S</sub>$  are from the connected transmission lines. The extracted capacitor values, at different bias voltages and at RF frequencies of 0.2, 2.0 and 20 GHz, are also shown in Fig. 1.



Fig. 4. The measured scattering parameters of a TiTaO MIM capacitor, from 200 MHz to 20 GHz.



Fig. 5. The equivalent-circuit model used for capacitance extraction from the measured S-parameters. The  $C$  and  $R<sub>P</sub>$  are the capacitor and its parallel resistance, and the  $L_S$  and  $R_S$  are the parasitic inductance and resistance, obtained from transmission lines connected between the capacitor body and probing pads.

The frequency dependence of the capacitance is shown in Fig. 6. Only a small variation of 3.6% occurs over the frequency range up to 20 GHz. In addition, a high quality (Q) factor of ~120 is shown at 1 GHz for an 8 pF capacitor. The nearly constant capacitance values, as shown in Fig. 6, in combination with the high Q-factor are important for analog/RF ICs incorporating such high density capacitors and operating from the IF to the RF regime.

The variation of capacitance (∆*C/C),* is one of the most important parameters for analog/RF IC applications. The value of ∆*C/C* below 1 MHz can be calculated from the *C* values, shown in Fig. 1 and measured by a precision LCR meter. However, the accuracy of  $\Delta$ *C*/*C* at RF frequencies is insufficient if obtained from the equivalent-circuit model. To achieve high precision  $\Delta C/C$  at the RF frequencies, we used equation (1), previously derived [5] for this purpose:

$$
\frac{\Delta C}{C} = \frac{Z_0 [2 + Z(\omega)/Z_0]^2}{2R_p^2} j\omega C (R_p + 1/j\omega C)^2 \Delta(S_{21})
$$
 (1)



Fig. 6. The frequency-dependence of the capacitance for a TiTaO MIM capacitor. The capacitance reduction is only observed at low frequency, since the trapped carriers in the high-κ dielectric of MIM capacitor can not follow the high frequency signal.

 $Z(\omega)$  in equation (1) is the total frequency-dependent impedance between the two ports with  $Z_0$  being the characteristic impedance of the transmission line. Based on the circuit theory derived equation (1), the ∆*C/C* can be obtained by differentiating the measured  $S_{21}$ , i.e.,  $\Delta S_{21}$  to enhance the small variations from the S-parameters. The RF frequency ∆*C/*C is plotted in Fig. 7. The ∆*C/*C data below 1 MHz, measured using the precision LCR meter, is also plotted for comparison. A rapid decrease of ∆C/C is observed with increasing frequency in the range 1 MHz to 200 MHz.



Fig. 7. The voltage dependence of the capacitance (∆*C/C*) for a TiTaO MIM capacitor. The low frequency data  $\leq 1$  MHz is from the *C-V* measurements in Fig. 1, and the RF frequency data >0.1 GHz is derived by eq. (1) from measured S-parameters.

To demonstrate the frequency dependence, we plot ∆*C/C* from 100 kHz to 20 GHz, in Fig. 8. Most of the decrease of <sup>∆</sup>*C/C* with increasing frequency occurs in the 1 MHz to 200 MHz range. Because the typical carrier lifetime from traprelated Shockley-Read-Hall (SHR) recombination is in the order of ~µs, the reduction of ∆*C/C* within the 1MHz to 200 MHz range may be due to the trapped carriers being unable to follow the high frequency signal [10]-[12]. The observed small ∆*C/C* of 770 ppm at 2GHz, or equivalent to a low quadratic voltage coefficient  $\alpha$  of 130 ppm/V<sup>2</sup>, is sufficient to meet the tight requirements of high-speed analog/RF IC applications.



Fig. 8.  $\Delta C/C$  for a 20 fF/ $\mu$ m<sup>2</sup> TiTaO MIM capacitor, biased at 2V, as a function of frequency. The line was drawn by hand and guided by the eye.

# IV. CONCLUSIONS

We have demonstrated TiTaO MIM capacitors that have a high capacitance density of 20 fF/ $\mu$ m<sup>2</sup>, a small capacitance reduction of 3.6%, and small  $\Delta$ C/C of 770 ppm at 2 GHz. The good MIM capacitor integrity makes them suitable for precision circuit applications at RF frequencies.

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