TABLE 1 Measured Distances for Multiple Targe

Real Distance (cm)	Measured Distance to the Target B (cm)	Measurement Error (%)	
50.0	47.51	4.98	
60.0	58.79	2.02	
70.0	65.3	6.7	
	69.19 (Removing the target A)	1.16	
90.	94.1	4.56	

power detectors is demonstrated via distance measurements to a single target and multiple targets. The RF power detector is designed with a dummy structure and a differential common emitter amplifier to cancel DC offset voltage. The maximum distance errors are 2% in a single target and 6.7% in multitarget environment using traveling wave antennas. The radar sensor has high-range accuracy with narrow bandwidth. The sensor for short-range detection has the advantages of easy designs to the all components in the sensor and high-range accuracy from zero to the maximum detectable range. The whole size of the radar sensor module included the six-port network made by HBT MMIC process, a signal conditioning block, and a patch antenna is $4 \times 4 \times 2$ cm³.

REFERENCES

- G.F. Engen, The six-port reflectometer: An alternative network analyzer, IEEE Trans Microwave Theory Tech MTT-25 (1977), 1077– 1079.
- C.G. Miguélez, A new automobile radar based on the six-port phase/ frequency discriminator, IEEE Trans Vehicular Technol 49 (2000), 1416–1423.
- 3. J. Wenger, Automotive radar—status and perspectives, IEEE CSIC Digest (2005), 21–24.
- V. Giammello, E. Ragonese, and G. Palmisano, A 24/77-GHz SiGe BiCMOS transmitter chipset for automotive radar, Microwave Opt Technol Lett 55 (2013), 782–786.
- F. Zhenghe, L. Yang, L. Liwan, P. Weifeng, and C. Yaqin, Requirement analysis of linearity for FMCW source using open-loop correction, In: Proceedings of International Conference on Microwave Millimeter Wave Technology, 2000, pp. 679–682.
- R.G. Meyer, Low-power monolithic RF peak detector analysis, IEEE J Solid-State Circuits 30 (1995), 65–67.
- J.-R. Yang, D.-W. Kim, and S. Hong, A calibration method of a range finder with a six-port network, IEEE Microwave Compon Lett 17 (2007), 549–551.
- J.R. James and P.S. Hall, Handbook of microstrip antennas, IEE Electromagnetic Waves Series, London, UK, vol. 28, 1989.

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STABILITY ANALYSIS OF THE PML SCHEME FOR CN-FDTD AND ADI-FDTD

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ABSTRACT: The alternating direction implicit-finite-difference timedomain (ADI-FDTD) method can be seen as a second-order perturbation of the Crank–Nicolson FDTD (CN-FDTD) scheme. When the PML is introduced for the ADI-FDTD method, the perturbation term will affect the stability of the ADI-FDTD method. In this work, the stability analysis of the PML schemes for the ADI-FDTD and CN-FDTD are demonstrated. It is found that the PML equations for ADI-FDTD method will be unstable due to the perturbation term. © 2014 Wiley Periodicals, Inc. Microwave Opt Technol Lett 56:2637–2642, 2014; View this article online at wileyonlinelibrary.com. DOI 10.1002/mop.28658

Key words: Crank–Nicolson FDTD; alternating direction implicit-finitedifference time-domain; PML; stability; perturbation term

1. INTRODUCTION

Finite-difference time-domain (FDTD) method has been widely used to analyze the electromagnetic problems [1]. Due to the explicit nature of this method, the time step size is restricted by the Courant, Friedrichs, and Lewy (CFL) stability condition. A stable alternating direction implicit (ADI) scheme was introduced for the FDTD method. The ADI-FDTD method is an attractive method due to its unconditional stability with large CFL number (CFLN) [2, 3]. When the ADI-FDTD method is used to simulate unbounded region problems, efficient absorbing boundary conditions must be used. The PML schemes were used for the ADI-FDTD method [4-6]. In [7, 8], the ADI-FDTD with CFS PML was also proposed to reduce the reflection error. However, the implementation of the PML in the ADI-FDTD method can affect the stability of this scheme. In [9], it is found that the ADI-FDTD method with split-field PML will lead to late-time instability. The Crank-Nicolson FDTD (CN-FDTD) is found to be another alternative unconditionally stable FDTD method. The ADI-FDTD can be seen as an approximation of the CN-FDTD scheme [10]. In [11, 12], the CN-FDTD with nearly PML and split-field PML were proposed. It is shown that the CN-FDTD can remain unconditionally stable with PML implementation.

In this article, the stability analysis of the PML schemes for the CN-FDTD and ADI-FDTD will be studied. The von Neumann analysis is used to determine the stability of these schemes. The difference between the CN-FDTD and ADI-FDTD is the Δt^2 perturbation term. From this study, it is found that the perturbation term will affect the stability of PML schemes for ADI-FDTD method.

2. PML FOR CN-FDTD AND ADI-FDTD

The ADI-FDTD can be seen as a second-order perturbation of CN-FDTD. Their formulations are described briefly. The Max-well's curl equations are given by

$$\partial_t \vec{E} = \frac{1}{\varepsilon} \tilde{\Re} \vec{H} \quad \partial_t \vec{H} = -\frac{1}{\mu} \tilde{\Re} \vec{E}$$
(1)

where ε and μ are the permittivity and permeability, respectively. \vec{E} and \vec{H} are the electric and magnetic fields, and $\tilde{\Re}$ is the curl operator. One can replace all the derivatives by centered difference operator and average in time the field affected by the curl operator to derive the CN-FDTD scheme,

$$(\overline{\Psi}^{n+1} - \overline{\Psi}^{n}) = \widetilde{\Re}_{\mathrm{T}} \frac{\Delta t}{2} (\overline{\Psi}^{n+1} + \overline{\Psi}^{n})$$
(2)

where $\vec{\Psi} = (E_x \quad E_y \quad E_z \quad H_x \quad H_y \quad H_z)^{\mathrm{T}}$ is the numerical vector field and the operator $\hat{\Re}_{\mathrm{T}}$ is the numerical counterpart of $\hat{\Re}$.

Equation (2) is the CN-FDTD scheme. It will require large computational resources to solve this scheme. Nevertheless, the

CN-FDTD can be reformulated to ADI-FDTD scheme that can be solved efficiently. Based on [10], the CN-FDTD method can be split into two-step procedure for ADI-FDTD.

The space operator is $\tilde{\Re}_T$ is decomposed as

Ã

$$+\tilde{B} = \tilde{\Re}_{\mathrm{T}} \tag{3}$$

and (2) can be rewritten as

$$\left(\tilde{I} - \frac{\Delta t}{2}(\tilde{A} + \tilde{B})\right)\tilde{\Psi}^{n+1} = \left(\tilde{I} + \frac{\Delta t}{2}(\tilde{A} + \tilde{B})\right)\tilde{\Psi}^{n}$$
(4)

The CN-FDTD scheme (4) can be rewritten as

$$\begin{pmatrix} \tilde{I} - \frac{\Delta t}{2}\tilde{A} \end{pmatrix} \begin{pmatrix} \tilde{I} - \frac{\Delta t}{2}\tilde{B} \end{pmatrix} \overrightarrow{\Psi}^{n+1} = \begin{pmatrix} \tilde{I} + \frac{\Delta t}{2}\tilde{A} \end{pmatrix} \begin{pmatrix} \tilde{I} + \frac{\Delta t}{2}\tilde{B} \end{pmatrix} \overrightarrow{\Psi}^{n} + \frac{\Delta t^{2}}{4}\tilde{A}\tilde{B} (\overrightarrow{\Psi}^{n+1} - \overrightarrow{\Psi}^{n})$$

$$(5)$$

If we neglect the Δt^2 perturbation term, we can get the approximation of CN-FDTD as ADI-FDTD

$$\left(\tilde{I} - \frac{\Delta t}{2}\tilde{A}\right) \left(\tilde{I} - \frac{\Delta t}{2}\tilde{B}\right) \overline{\Psi}^{n+1} = \left(\tilde{I} + \frac{\Delta t}{2}\tilde{A}\right) \left(\tilde{I} + \frac{\Delta t}{2}\tilde{B}\right) \overline{\Psi}^{n} \quad (6)$$

Therefore, the ADI-FDTD scheme (6) can be seen as a secondorder perturbation of the CN-FDTD (2)

Equation (6) can be further split into two updating steps

$$\left(\tilde{I} - \frac{\Delta t}{2}\tilde{A}\right) \overrightarrow{\Psi}^{n*} = \left(\tilde{I} + \frac{\Delta t}{2}\tilde{B}\right) \overrightarrow{\Psi}^{n},$$

$$\left(\tilde{I} - \frac{\Delta t}{2}\tilde{B}\right) \overrightarrow{\Psi}^{n+1} = \left(\tilde{I} + \frac{\Delta t}{2}\tilde{A}\right) \overrightarrow{\Psi}^{n*}$$

$$(7)$$

where $\overline{\Psi}^{n*}$ is an auxiliary intermediate vector field. The ADI-FDTD Eq. (7) can be solved efficiently.

To simulate unbounded region problems, the PML formulations should be implemented for CN-FDTD and ADI-FDTD. We can use the same procedure to reformulate the PML equations. For the PML equations, the PML conductivity is incorporated into matrix $\tilde{\Re}_T$. Once we derive the operator $\tilde{\Re}_T$ and split into two operators \tilde{A} and \tilde{B} , we can reformulate the PML equations for CN-FDTD and ADI-FDTD. In this study, the unsplitfield PML, split-field PML, and CFS PML formulations are investigated.

2.1. Unsplit-Field PML Scheme

The unsplit-field PML scheme is based on the formulations derived in [6]. The unsplit form PML equations are

$$\partial_t \vec{E} = \frac{1}{\varepsilon} \tilde{\Re} \vec{H} - \frac{1}{\varepsilon} \tilde{o} \vec{E} - \frac{1}{\varepsilon} (\tilde{e} - \tilde{o}) \vec{E}_e$$
(8a)

$$\partial_t \vec{H} = -\frac{1}{\mu} \tilde{\Re} \vec{E} - \frac{1}{\varepsilon} \tilde{\epsilon} \vec{H} - \frac{1}{\varepsilon} (\tilde{\epsilon} - \tilde{o}) \vec{H}_o$$
(8b)

$$\partial_t \vec{E}_e = \frac{1}{\varepsilon} \tilde{\Re}_e \vec{H} - \frac{1}{\varepsilon} \tilde{\epsilon} \vec{E}_e$$
 (8c)

$$\partial_t \vec{H}_{\rm o} = -\frac{1}{\mu} \hat{\Re}_{\rm o} \vec{E} - \frac{1}{\varepsilon} \tilde{}_{\rm o} \vec{H}_{\rm o}$$
(8d)

where ∂_u represents the partial derivative with respective to *u* direction, \vec{E}_e and \vec{H}_o are two auxiliary fields, and

$$\tilde{\Re} = \begin{pmatrix} 0 & -\partial_z & \partial_y \\ \partial_z & 0 & -\partial_x \\ -\partial_y & \partial_x & 0 \end{pmatrix}, \quad \tilde{\Re} = \tilde{\Re}_e + \tilde{\Re}_o$$

$$\tilde{\Re}_e = \begin{pmatrix} 0 & 0 & \partial_y \\ \partial_z & 0 & 0 \\ 0 & \partial_x & 0 \end{pmatrix} \quad \tilde{\Re}_o = \begin{pmatrix} 0 & -\partial_z & \partial \\ 0 & 0 & -\partial_x \\ -\partial_y & 0 & 0 \end{pmatrix}$$

$$\tilde{\sigma}_e = \begin{pmatrix} \sigma_y & 0 & 0 \\ 0 & \sigma_z & 0 \\ 0 & 0 & \sigma_x \end{pmatrix} \quad \tilde{o} = \begin{pmatrix} \sigma_z & 0 & 0 \\ 0 & \sigma_x & 0 \\ 0 & 0 & \sigma_y \end{pmatrix}$$
(9)

(8) can be written in a compact form as

$$\partial_t \vec{\Psi}(t) = \tilde{\Re}_{\mathrm{T}} \vec{\Psi}(t) \tag{10}$$

where $\Psi(t)$ is the compound Cartesian vector

$$\overline{\Psi} = \begin{pmatrix} E_x & E_y & E_z & H_x & H_y & H_z \\ E_{\text{ex}} & E_{\text{ey}} & E_{\text{ez}} & H_{\text{ox}} & H_{\text{oy}} & H_{\text{oz}} \end{pmatrix}^{\text{T}}$$
(11)

And

$$\tilde{\Re}_{\mathrm{T}} = \begin{pmatrix} -\frac{1}{\varepsilon}\tilde{\sigma}_{\mathrm{o}} & \frac{1}{\varepsilon}\tilde{\Re} & -\frac{1}{\varepsilon}(\tilde{\sigma}_{\mathrm{e}}-\tilde{\sigma}_{\mathrm{o}}) & \tilde{0} \\ \frac{1}{\mu}\tilde{\Re} & -\frac{1}{\varepsilon}\tilde{\sigma}_{\mathrm{e}} & \tilde{0} & \frac{1}{\varepsilon}(\tilde{\sigma}_{\mathrm{e}}-\tilde{\sigma}_{\mathrm{o}}) \\ \tilde{0} & \frac{1}{\varepsilon}\tilde{\Re}_{\mathrm{e}} & -\frac{1}{\varepsilon}\tilde{\sigma}_{\mathrm{e}} & \tilde{0} \\ -\frac{1}{\mu}\tilde{\Re}_{\mathrm{o}} & \tilde{0} & \tilde{0} & -\frac{1}{\varepsilon}\tilde{\sigma}_{\mathrm{o}} \end{pmatrix}$$
(12)

where \tilde{I} and $\tilde{0}$ are 3 × 3 identity and null matrix, respectively. The operators \tilde{A} and \tilde{B} are chosen, so that

$$\tilde{A} + \tilde{B} = \tilde{\Re}_{\mathrm{T}} \tag{13}$$

A possible choice of \tilde{A} and \tilde{B} is given by

$$\tilde{A} = \begin{pmatrix} \tilde{0} & \frac{1}{\varepsilon} \tilde{\Re}_{e} & -\frac{1}{\varepsilon} \tilde{\sigma}_{e} & \tilde{0} \\ -\frac{1}{\mu} \tilde{\Re}_{o} & \tilde{0} & \tilde{0} & -\frac{1}{\varepsilon} \tilde{\sigma}_{o} \\ \tilde{0} & \frac{1}{\varepsilon} \tilde{\Re}_{e} & -\frac{1}{\varepsilon} \tilde{\sigma}_{e} & \tilde{0} \\ -\frac{1}{\mu} \tilde{\Re}_{o} & \tilde{0} & \tilde{0} & -\frac{1}{\varepsilon} \tilde{\sigma}_{o} \end{pmatrix}$$
(14)

$$\tilde{B} = \begin{pmatrix} -\frac{1}{\varepsilon}\tilde{\sigma}_{0} & \frac{1}{\varepsilon}\tilde{\Re}_{0} & \frac{1}{\varepsilon}\tilde{\sigma}_{0} & \tilde{0} \\ -\frac{1}{\varepsilon}\tilde{\Re}_{e} & -\frac{1}{\varepsilon}\sigma_{e} & \tilde{0} & \frac{1}{\varepsilon}\tilde{\sigma}_{e} \\ \tilde{0} & \tilde{0} & \tilde{0} & \tilde{0} \\ \tilde{0} & \tilde{0} & \tilde{0} & \tilde{0} \end{pmatrix}$$
(15)

The field components \vec{E} , \vec{H} , \vec{E}_{e} , and \vec{H}_{o} can be solved by CN-FDTD (4) or by two updating steps ADI-FDTD method (7). The system Eq. (7) can be further triangularized to solve it. When we set the PML conductivity $\sigma_x = \sigma_y = \sigma_z = 0$, \vec{E} and \vec{H} formulations will be identical to the ADI-FDTD method.

2.2. Split-Field PML Scheme

The split-field PML formulations can also be expressed in the partial differential form (1) and solved by the CN-FDTD scheme. Based on [4], the field vector is defined as

$$\overrightarrow{\Psi} = \begin{pmatrix} E_{xy} & E_{yz} & E_{zx} & E_{xz} & E_{yx} & E_{zy} \\ H_{xy} & H_{yz} & H_{zx} & H_{xz} & H_{yx} & H_{zy} \end{pmatrix}^{\mathrm{T}}$$
(16)

and $\tilde{\Re}_T$ is a 12 \times 12 dimensional space operator

$$\tilde{\Re}_{\mathrm{T}} = \begin{pmatrix} -\frac{1}{\varepsilon} \tilde{\sigma}_{\mathrm{e}} & \tilde{0} & \frac{1}{\varepsilon} \tilde{R}_{\mathrm{e}} & \frac{1}{\varepsilon} \tilde{R}_{\mathrm{e}} \\ \tilde{0} & -\frac{1}{\varepsilon} \tilde{\sigma}_{\mathrm{o}} & \frac{1}{\varepsilon} \tilde{R}_{\mathrm{o}} & \frac{1}{\varepsilon} \tilde{R}_{\mathrm{o}} \\ -\frac{1}{\mu} \tilde{R}_{\mathrm{e}} & -\frac{1}{\mu} \tilde{R}_{\mathrm{e}} & -\frac{1}{\varepsilon} R \tilde{\sigma}_{\mathrm{e}} & \tilde{0} \\ -\frac{1}{\mu} \tilde{R}_{\mathrm{o}} & -\frac{1}{\mu} \tilde{R}_{\mathrm{o}} & \tilde{0} & -\frac{1}{\varepsilon} \tilde{\sigma}_{\mathrm{o}} \end{pmatrix}$$
(17)

 $\hat{\Re}_{T}$ is split into two operators to derive ADI-FDTD scheme and the operators \tilde{A} and \tilde{B} are given by

TABLE 1 Eigenvalues of Λ and (G for Unsplit PML Scheme
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$$\tilde{A} = \begin{pmatrix} -\frac{1}{2\varepsilon}\tilde{\sigma}_{e} & \tilde{0} & \frac{1}{\varepsilon}\tilde{R}_{e} & \frac{1}{\varepsilon}\tilde{R}_{e} \\ \tilde{0} & -\frac{1}{2\varepsilon}\tilde{\sigma}_{o} & \tilde{0} & \tilde{0} \\ \tilde{0} & \tilde{0} & -\frac{1}{2\varepsilon}\tilde{\sigma}_{e} & \tilde{0} \\ -\frac{1}{\mu}\tilde{R}_{o} & -\frac{1}{\mu}\tilde{R}_{o} & \tilde{0} & -\frac{1}{2\varepsilon}\tilde{\sigma}_{o} \end{pmatrix}$$

$$and \qquad (18)$$

$$\tilde{B} = \begin{pmatrix} -\frac{1}{2\varepsilon}\tilde{\sigma}_{e} & \tilde{0} & \tilde{0} & \tilde{0} \\ \tilde{0} & -\frac{1}{2\varepsilon}\tilde{\sigma}_{o} & \frac{1}{\varepsilon}\tilde{R}_{o} & \frac{1}{\varepsilon}\tilde{R}_{o} \\ -\frac{1}{\mu}\tilde{R}_{e} & -\frac{1}{\mu}\tilde{R}_{e} & -\frac{1}{2\varepsilon}\tilde{\sigma}_{e} & \tilde{0} \\ \tilde{0} & \tilde{0} & \tilde{0} & \frac{1}{2\varepsilon}\tilde{\sigma}_{o} \end{pmatrix}$$

Compared to the unsplit-field PML scheme, the split-field PML equation is less complicate and more straightforward. However, this formulation suffers from large reflection error, when attempting to absorb the low-frequency evanescent wave. The CFS PML scheme can accurately absorb the low-frequency wave and is discussed below.

2.3. CFS PML Scheme

The PML equations can be expressed in the stretched coordinate space [7]. For example, considering the *x*-projection of Ampere's law

	CFLN=1	CFLN=5	CFLN=15
Split-field PML for	9.9422334 e -001	9.9590080 e −001	1.0066924 e +000
ADI-FDTD $\sigma = 0.1066 \lambda_{\Lambda}$	9.9422334 e −001	9.9590080 e −001	1.0066924 e +000
	9.9422334 e −001	9.9590080 e −001	1.0066924 e +000
	9.9422334 e −001	9.9590080 e −001	1.0066924 e +000
	9.8563297 e −001	9.4176082 e −001	8.3275850 e -001
	9.8563297 e −001	9.4176082 e −001	8.3275850 e -001
	9.8563297 e −001	9.4176082 e −001	8.3275850 e -001
	9.8563297 e −001	9.4176082 e −001	8.3275850 e -001
	1.0000000 e + 000	1.0000000 e + 000	9.9999972 e -001
	9.99999999 e −001	1.0000000 e +000	1.0000000 e +000
	1.0000000 e + 000	9.9999996 e −001	1.000000 e + 000
	1.0000000 e + 000	9.99999999 e −001	9.99999999 e -001
Split-field PML for	9.9614053 e −001	9.9851056 e −001	9.9948561 e -001
CN-FDTD σ =0.1066 $\lambda_{\rm G}$	9.9614053 e −001	9.9851056 e −001	9.9948561 e -001
	9.9614053 e −001	9.9851056 e −001	9.9948561 e -001
	9.9614053 e −001	9.9851056 e −001	9.9948561 e -001
	9.8468627 e −001	9.2575790 e −001	7.9378279 e -001
	9.8468627 e −001	9.2575790 e −001	7.9378279 e -001
	9.8468627 e −001	9.2575790 e −001	7.9378279 e -001
	9.8468627 e −001	9.2575790 e −001	7.9378279 e -001
	1.0000000 e + 000	1.0000000 e + 000	9.9999986 e -001
	1.0000000 e +000	1.0000000 e +000	9.9999996 e -001
	9.99999999 e −001	9.9999996 e −001	1.000000 e + 000
	1.0000000 e + 000	9.9999998 e -001	1.0000000 e + 000

$$j\omega\varepsilon E_x = \frac{1}{s_y}\frac{\partial}{\partial y}H_z - \frac{1}{s_z}\frac{\partial}{\partial_z}H_y$$
(19)

and the z-projection of Faraday's law

$$j\omega\mu H_z = \frac{1}{s_y} \frac{\partial}{\partial y} E_x - \frac{1}{s_x} \frac{\partial}{\partial x} E_y$$
(20)

where s_i is the stretched coordinate metric and is chosen to be

$$s_i = \kappa_i + \frac{\sigma_i}{\alpha_i + j\omega\varepsilon} \tag{21}$$

When (19) and (20) are transformed into time-domain, the right-hand side terms will become convolution. The convolutional CFS PML scheme is not suitable to formulate CN-FDTD method. In [10], the authors use the auxiliary differential equations to express the Eqs. (16) and (17) and formulate CFS PML scheme to be suitable for CN-FDTD scheme. The field vector is defined as

$$\overline{\Psi} = (E_x \quad E_y \quad E_z \quad f_{xy} \quad f_{xz} \quad f_{yz} \quad f_{yx} \quad f_{zx} \quad f_{zy}
H_x \quad H_y \quad H_z \quad g_{xy} \quad g_{xz} \quad g_{yz} \quad g_{yx} \quad g_{zx} \quad g_{zy})$$
(22)

where f and g are auxiliary variables and $\hat{\Re}_{T}$ is a 18 \times 18 dimensional space operator

$$\tilde{\Re}_{\mathrm{T}} = \begin{pmatrix} \tilde{0} & \frac{1}{\varepsilon} \tilde{C}_{\mathrm{ss}} & \frac{1}{\varepsilon} \tilde{R}_{\mathrm{k}} & \tilde{0} \\ \tilde{0} & -\tilde{\sigma}_{\mathrm{s}} & \tilde{R}_{\mathrm{ss}} & \tilde{0} \\ -\frac{1}{\mu} \tilde{R}_{\mathrm{k}} & \tilde{0} & \tilde{0} & -\frac{1}{\mu} \tilde{C}_{\mathrm{ss}} \\ \tilde{R}_{\mathrm{ss}} & \tilde{0} & \tilde{0} & -\tilde{\sigma}_{\mathrm{s}} \end{pmatrix}$$
(23)

After splitting $\tilde{\Re}_{T}$, we can derive the ADI-FDTD scheme. The operators \tilde{A} and \tilde{B} are given by

$$\tilde{A} = \begin{pmatrix} \tilde{0} & \frac{1}{\varepsilon} \tilde{C}_{se} & \frac{1}{\varepsilon} \tilde{R}_{ke} & \tilde{0} \\ \tilde{0} & -\frac{1}{2} \tilde{\sigma}_{s} & \tilde{R}_{sse} & \tilde{0} \\ \frac{1}{\mu} \tilde{R}_{ko} & \tilde{0} & \tilde{0} & -\frac{1}{\mu} \tilde{C}_{so} \\ \tilde{R}_{sso} & \tilde{0} & \tilde{0} & -\frac{1}{2} \tilde{\sigma}_{s} \end{pmatrix}$$
and
$$\tilde{B} = \begin{pmatrix} \tilde{0} & \frac{1}{\varepsilon} \tilde{C}_{so} & \frac{1}{\varepsilon} \tilde{R}_{ko} & \tilde{0} \\ \tilde{0} & -\frac{1}{2} \tilde{\sigma}_{s} & \tilde{R}_{sso} & \tilde{0} \\ \tilde{0} & -\frac{1}{2} \tilde{\sigma}_{s} & \tilde{R}_{sso} & \tilde{0} \\ -\frac{1}{\mu} \tilde{R}_{ke} & \tilde{0} & \tilde{0} & -\frac{1}{\mu} \tilde{C}_{se} \\ \tilde{R}_{sse} & \tilde{0} & \tilde{0} & -\frac{1}{2} \tilde{\sigma}_{s} \end{pmatrix}$$

$$(24)$$

where $\tilde{R}_{ss} = \tilde{R}_{sse} + \tilde{R}_{sso}$, $\vec{C}_{ss} = \vec{C}_{sse} + \vec{C}_{sso}$, and $\vec{R}_{k} = \vec{R}_{ke} + \vec{R}_{ko}$. The matrices \tilde{R}_{ss} , \vec{C}_{ss} , and \vec{R}_{k} are related to PML parameters κ , σ , and α . Details of these matrices are shown in [10]. The ADI-FDTD with CFS PML is a simple extension of the ADI-FDTD method. This scheme can provide wideband absorption of the incident wave.

3. THEORETICAL STABILITY ANALYSIS

To study the stability of the PML schemes for the CN-FDTD and ADI-FDTD method, the von Neumann method is used. Following the similar procedures presented in [3], we assume that for each time step the field components are Fourier-transformed into the spatial spectral domain. From the system equations of (4), the CN-FDTD scheme can be written in the spatial spectral domain in a matrix form as

$$\mathbf{X}^{n+1} = G\mathbf{X}^n \tag{25}$$

where the vector X^n represents the field components and the auxiliary variables at the *n*th time step. From the system equations of (7), the ADI-FDTD scheme can also be written in the spatial spectral domain in matrix form as

$$M_1 \mathbf{X}^{n|} = P_1 \mathbf{X}^n \tag{26}$$

$$M_2 X^{n+1} = P_2 X^{n|} \tag{27}$$

for the n^* and n + 1 time steps, respectively. The entries for the matrices M and P are derived from updating equations. The two half time steps can be combined to one time step

$$X^{n+1} = M_2^{-1} P_2 M_1^{-1} P_1 X^n = \Lambda X^n$$
(28)

The stability criterion requires that the eigenvalues of amplification matrices G and Λ lie within or on the unit circle. An attempt to determine the eigenvalues of G and Λ symbolically was made. However, due to the complexity of the amplification matrix, it is difficult to get a simplified analytical expression for the eigenvalues. The maximum eigenvalues are numerically calculated by Matlab®. In this study, we set the cell size to be $\Delta x = \Delta y = \Delta z = 1.0$ mm and FDTD time step limit Δt max = 1.92 ps is used. The ratio of $\Delta t/\Delta t$ max is defined as the CFLN.

3.1. Unsplit-Field PML Scheme

The eigenvalues of G and Λ are computed when $\sigma_x = \sigma_z = 0$, and $\sigma_v = 10.66$ S/m. The time step sizes are chosen to be Δt_{max} , $2\Delta t_{max}$, and $5\Delta t_{max}$, respectively. Both the unsplit-field PML schemes for the CN-FDTD scheme and ADI-FDTD are investigated and the theoretical results are shown in Table 1. When CFLN = 1, both the ADI-FDTD and CN-FDTD PML formulations can be stable as the maximum eigenvalues for the two schemes are smaller than unity. However, it is found that the eigenvalues of the unsplit-field PML scheme for ADI-FDTD scheme are larger than unity when CFLN = 2 is used. Conversely, we use different CFLN and σ values and no instability was observed for CN-FDTD PML scheme. As shown in (5), the difference between CN-FDTD and ADI-FDTD is the Δt^2 perturbation term. The results indicate that perturbation term will affect the stability of PML scheme for ADI-FDTD when large CFLN is used.

3.2. Split-Field PML Scheme

The split-field PML schemes for CN-FDTD and ADI-FDTD are studied. In [9], it indicated that the split-field PML for the ADI-FDTD method will be unstable when the PML conductivity

TABLE 2 Eigenvalues of Λ and G for Split PML Scheme

	CFLN=1	CFLN=2	CFLN=5
Unsplit-field PML for	1.0000000 e +000	1.0000000 e + 000	1.0000000 e +000
ADI-FDTD $\sigma = 10.66 \lambda $	8.7747147 e −001	1.0000000 e + 000	1.0000000 e +000
	8.7747147 e −001	1.1404158 e+000	1.8425689 e +000
	8.7747147 e −001	1.1404158 e+000	1.8425689 e +000
	1.0000000 e + 000	1.0000000 e + 000	1.8425689 e +000
	1.0000000 e + 000	1.0000000 e + 000	1.8425689 e +000
	8.7747147 e −001	1.1404158 e+000	1.0000000 e + 000
	3.0475086 e −001	1.1404158 e +000	1.0000000 e + 000
	3.0475086 e −001	4.3679841 e -001	4.1123514 e -001
	3.0475086 e −001	4.3679841 e -001	4.1123514 e -001
	3.0475086 e −001	4.3679841 e -001	4.1123514 e -001
	1.0000000 e + 000	4.3679841 e -001	4.1123514 e -001
Unsplit-field PMLfor	1.0000000 e + 000	1.0000000 e + 000	1.0000000 e + 000
CN-FDTD σ =10.66 $\lambda_{\rm G}$	9.2402257 e −001	9.2903235 e −001	9.6195689 e −001
	9.2402257 e −001	9.2903235 e -001	9.6195689 e −001
	9.2402257 e −001	9.2903235 e −001	9.6195689 e −001
	9.2402257 e −001	9.2903235 e -001	9.6195689 e −001
	2.3259772 e −001	4.6207853 e −001	7.3695418 e -001
	2.3259772 e −001	4.6207853 e −001	7.3695418 e -001
	2.3259772 e −001	4.6207853 e −001	7.3695418 e -001
	2.3259772 e −001	4.6207853 e −001	7.3695418 e -001
	1.0000000 e + 000	1.0000000 e + 000	1.0000000 e + 000
	1.0000000 e + 000	9 9999998 e -001	9.9999997 e −001
	1.0000000 e + 000	1.0000000 e + 000	1.0000000 e + 000

profile is polynomial scaled and this scheme can be stable with constant PML conductivity. In this study, the smaller PML conductivity $\sigma_x = \sigma_z = 0$ and $\sigma_y = 0.1066$ S/m is used. From Table 2, it is found that the split-field PML scheme for ADI-FDTD can be stable when CFLN = 5 and will become unstable when large CFLN = 15 is used. No instability is observed when split-field PML for CN-FDTD scheme is used.

3.3. CFS PML Scheme

The convolutional CFS PML scheme is not suitable to formulate CN-FDTD method. It has also been proved in [7] that convolutional CFS PML for ADI-FDTD can be unconditionally stable. In [10], the CFS PML formulations can also be expressed for CN-FDTD and ADI-FDTD schemes, as discussed in Section 2. We find that the CFS PML for CN-FDTD can still be uncondi-

tionally stable. Table 3 shows the calculated eigenvalues of Λ with $\sigma_s = 10$ S/m, $\kappa_s = 5$, and $\alpha_s = 0.1$ S/m. When CFS PML is formulated for ADI-FDTD (7), this scheme will also lead to unstable condition when large CFLN = 15 is used.

4. NUMERICAL VERFICATION

Numerical simulations are performed by 3-D ADI-FDTD with unsplit-field PML and split-field PML to validate the instability of the two schemes. A uniform mesh with cell size $\Delta x = \Delta y =$ $\Delta z = 1.0$ mm and FDTD time step limit $\Delta t_{max} = 1.92$ ps are used. The computation domain is $42 \times 42 \times 42$. PML layers that are 10 cells thick terminated all six sides of the computation domain. A differential Gaussian pulse applied to H_x field is excited at the center position (21, 21, 21). The polynomial scaling is used for the PML conductivity profile

TABLE 3 Eigenvalues of Λ for CFS PML ADI-FDTD with ADE Formulation	TABLE 3	Eigenvalues of	Λ for CFS PML	ADI-FDTD with	ADE Formulation
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	CFLN=1	CFLN=5	CFLN=15
Unsplit-field PML	6.4610024 e -001	9.5826739 e −001	6.7860023 e -001
for ADI-FDTD $ \lambda_{\Lambda} $	6.4610024 e -001	9.5826739 e −001	6.7860023 e -001
	6.4610024 e -001	9.5826739 e −001	6.7860023 e -001
	6.4610024 e -001	9.5826739 e −001	6.7860023 e -001
	6.3629102 e -001	1.0000000 e + 000	1.1340967 e +000
	9.9044985 e −001	1.0000000 e + 000	1.1340967 e +000
	9.9044985 e −001	2.4086514 e -001	1.1340967 e +000
	9.9044985 e −001	2.4086514 e -001	1.1340967 e +000
	9.9044985 e −001	2.4086514 e -001	1.0000000 e + 000
	1.0000000 e + 000	2.4086514 e -001	1.0000000 e + 000
	1.0000000 e + 000	1.6715028 e -001	4.0697273 e -001
	6.3629102 e -001	1.6715028 e -001	4.0697273 e -001
	6.3629102 e -001	1.2104686 e -001	4.0697273 e -001
	6.3629102 e -001	1.2104686 e -001	4.0697273 e -001
	6.3267285 e -001	1.6715028 e -001	2.1521203 e -001
	6.3267285 e -001	1.6715028 e -001	2.1521203 e -001
	6.3267285 e -001	1.2104686 e -001	2.1521203 e -001
	6.3267285 e -001	1.2104686 e -001	2.1521203 e -001

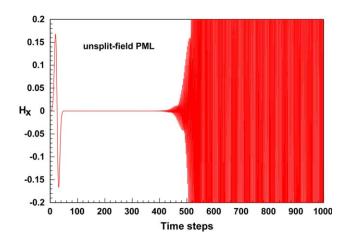


Figure 1 The H_x component for ADI-FDTD with unsplit-field PML (CFLN = 2). [Color figure can be viewed in the online issue, which is available at wileyonlinelibrary.com]

$$\sigma_{\text{smax}} = \sigma_{\text{opt}} \approx \frac{(m+1)}{150\pi\Delta s}$$

$$\sigma_s(s) = \frac{\sigma_{\text{smax}}|S - S_0|^m}{d_m} \quad s = x, y, z$$
(29)

where *d* is the thickness of PML absorber, Δs is the cell size, and s_0 represents the interface. In this simulation, we choose scaling factor m = 4 and $\sigma_{\text{max}} = 10.61$ S/m for optimum PML performance [4].

First, numerical simulation of the ADI-FDTD with unsplitfield PML is performed. The time step size in this study is chosen to be $2\Delta t_{max}$. Figure 1 shows the time-domain H_x fields recorded at the position (21, 20, 21). As shown in Figure 1, this scheme will become unstable after running 400 time steps.

Then, numerical simulation of the ADI-FDTD with splitfield PML is performed. The time step size in this study is $7\Delta t$ max. As shown in Figure 2, this scheme will become unstable after running 1200 time steps.

5. CONCLUSION

In this work, the stability analysis of the PML schemes for the ADI-FDTD and CN-FDTD are studied. The ADI-FDTD can be seen as a second-order perturbation of the CN-FDTD method.

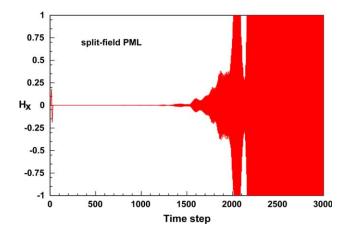


Figure 2 The H_x component for ADI-FDTD with split-field PML (CFLN = 7). [Color figure can be viewed in the online issue, which is available at wileyonlinelibrary.com]

From the theoretical stability analysis, it is found that the ADI-FDTD will be unstable while the CN-FDTD can remain stable when the PML medium is used. We find that the Δt^2 perturbation term will affect the stability of the PML schemes for ADI-FDTD.

REFERENCES

- A. Taflove and S. Hagness, Computational electrodynamics: the finite-difference time-domain method, 3rd ed., Artech House, Boston, MA, 2005.
- T. Namiki, A new FDTD algorithm based on alternating-direction implicit method, IEEE Trans Microwave Theory Tech 47 (1999), 2003–2007.
- F. Zheng, Z. Chen, and J. Zhang, Toward the development of a three-dimensional unconditionally stable finite-difference timedomain method, IEEE Trans Microwave Theory Tech 48 (2000), 1550–1558.
- G. Liu and S.D Gedney, Perfectly matched layer media for an unconditionally stable three-dimensional ADI-FDTD method, IEEE Microwave Guided Wave Lett 10 (2000), 261–263.
- C.C.-P. Chen, T.-W. Lee, N. Murugesan, and S.C. Hagness, Generalized FDTD-ADI: An unconditionally stable full-wave Maxwell's equations solver for VLSI interconnect modeling, In: Proceedings of International Conference on Computer-Aided Design, San Jose, CA, 2000, pp. 156–163.
- R.G. Rubio, S.G. Garcia, A.R. Bretones, and R.G. Martin, An unsplit Berenger-Like PML for the ADI-FDTD method, Microwave Opt Technol Lett 42 (2004), 466–469.
- S.D. Gedney, G. Liu, J.A. Roden, and A. Zhu, Perfectly matched layer media with CFS for an unconditionally stable ADI-FDTD method, IEEE Trans Antennas Propag 49 (2001), 1554–1559.
- L. Wang and C. Liang, A new implementation of CFS-PML for ADI-FDTD method, Microwave Opt Technol Lett 49 (2006), 1924– 1928.
- J.-N. Hwang and F.-C. Chen, A modified PML conductivity profile for the ADI-FDTD Method with split-field PML, IET Proc Microwave Antennas Propag 1 (2007), 1071–1077.
- R.G. Rubio, S.G. Garcia, A.R. Bretones, and R.G. Martin, Crank-Nicolson reformulation of ADI-FDTD PML extensions, Antennas Wireless Propag Lett 5 (2006), 357–360.
- D. Wu and J. Chen, Perfectly matched layer for Crank-Nicolson (CN) FDTD method, In IEEE Antennas Propagation Symposium. Digest, Monterey, CA, 2004, pp. 583–586.
- O. Ramadan. Unconditionally stable Crank-Nicolson nearly PML algorithm for truncating linear Lorentz dispersive FDTD domains, IEEE Trans Microwave Theory Tech 54 (2006), 2807–2812.

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REFRACTIVE INDEX SENSOR BASED ON MACH-ZEHNDER INTERFEROMETER FORMED BY TWO CASCADED SINGLE MODE FIBER CORNERS

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ABSTRACT: We reported a novel fiber sensor based on two cascaded single mode fiber corners, which has a high sensitivity for refractive index (RI) sensing and a low temperature crosstalk. RI sensing was experimentally demonstrated with an average sensitivity of 235 nm/RI