

A LAB-SCALE DIGITAL ACOUSTIC EMISSION SYSTEM FOR SOURCE LOCATION

signal processing (DSP) software packages for PC are avail- . 2 shows a set of simulated signals, with the *AE* from (1, 21, and critical factors involved are presented. **20, and 50 kHz**) are shown in Figure 3-5. It was found that

coustic emission *(AE)* techniques have been used \cdot of arrival-time differences *(delta-T's)* for each mesh point. To in the study of crack location, cracking mecha- \cdot calculate the source position for a set of burstcoustic emission (AE) techniques have been used
in the study of crack location, cracking mechanical calculate the source position for a set of burst-type AE signisms, deformation of metals and leak detection.
In these tech nisms, deformation of metals and leak detection. nals, a straightforward "shift and add" method was used.
In these techniques, AE sensors hit by transient . First, assuming the AE source is at a mesh point inside the stress waves generate *AE* signals, and then the *AE* signals . circle, the signal waveform of each channel was shifted forward in the time domain, with the corresponding delta-T tems only record some parameters of the AE signals, such value. Each time-shifted waveform was added to that of the as peak amplitude, ringdown counts, duration, risetime, and * first-hit channel, and then the resultant additive waveform so on. On the other hand, a fully digital *AE* system can dig- ' was squared to give the signal power from that mesh point. itize the waveforms of AE signals and perform various anal-
yses. Since many versatile and relatively cheap personal points, a power map can be created, and the peak of power yses. Since many versatile and relatively cheap personal : points, a power map can be created, and the peak of power computers (PC), analog-to-digital (A/D) cards, and digital : map is considered the closest point to the A computers (PC), analog-to-digital (A/D) cards, and digital map is considered the closest point to the AE source. Figures signal processing (DSP) software packages for PC are avail- 2 shows a set of simulated signals, with able, a fully digital and PC-based lab-scale AE system for 3- . propagating at 1480 m/s, of 10 kHz carrier frequency and D source location can be set up. With the results of DSP for \cdot sampled at 100 kHz. With this sampl sampled at 100 kHz. With this sampling rate, a series of simulated and real AE signals, features of this AE system \cdot location results for various carrier frequencies of signal (10,

per channel for a total number of 8 channels. This means an : approximate 7-times sampling of signals and a time resolu-
tion of 5 microseconds. Digitized AE signals were stored in . If a signal appears with infinite duration and excellent repsoftware, so this system was highly flexible. BE493 A/D cards, working at a sampling rate of 200 kHz same result was obtained when evaluating with cross-

SAMPLING RATE

For a digital AE system to perform properly, sampling at a rate high enough is a prerequisite. Because of the nature of irreproducibility of *AE* phenomena, *AE* signals simulated by software were used to evaluate the influence of sampling rate. Assume a circle with a diameter of 12 meters is on the *XY* plane and centered at the origin. *Six AE* sensors were assumed to have been mounted, equally spaced along the circumference of this circle. Then, an imaginary square mesh with an edge of 12 meters, 0.5-meter interval, also centered at the origin, is placed on the *XY* plane. This results in 25 \times 25 mesh points. Since the propagation speed of AE burst is known, the time needed for an *AE* burst from any mesh point to reach each sensor is known. That is, there is a set

when the sampling rate was sults were good no matter kHz) AE sensors were used
in our leak testing. The A/D **Fig. 1: The configuration of AE system**
and the loosting the loosting the loosting in our leak testing. The A/D **Fig. I: The configuration of AE system** and a marginally incessing the Ny result was incorrect. The control of the control of the system and the control of the control of two Nicolet was incorrect. The control of the cont

correlation method.

an ordinary WINTEL PC, and MATLAB was used for data . The signal appears what intimite datation and exclude ϵ reprocessing. Not only the AE sensors and amplifiers were examples often shown in textbooks when processing. Not only the *AE* sensors and amplifiers were ex-
changeable, but also the PC, A/D cards and the analysis the comic frequency is sufficient. In AE testing an uniqu the carrier frequency is sufficient. In *AE* testing or our sim-

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Fig. 3: Source location result when carrier frequency is 10 kHz

Fig. 4: Source location result when carrier frequency is 20 kHz

Fig. 5: Source location result when carrier frequency is 50 kHz

ulation where signals are of finite duration and consist of noise, a sampling rate of Nyquist frequency seems insufficient. As a rough approximation, an AE signal $h(t)$ can be simulated by a sinusoidal function with an exponential decay envelope along with noise, that is, $h(t) = Ae^{-\alpha t} \cos(2\pi ft)$ $+$ *noise*, where A and α are constants, t is time, and f is the carrier frequency. The frequency spectrum of this signal exhibits a bell-shape (due to the exponential decay), and is shifted and centered at *f* (due to modulation). **As** the signal becomes shorter (greater α), plus the effect of noise, the bellshaped curve spreads. This means more frequency (and time) information will be lost if the sampling rate is fixed at Nyquist frequency, then the result of data processing will tend to be incorrect.

DIGITAL SIGNAL PROCESSING

To pinpoint the position of an *AE* source, the propagation speed of *AE* burst, the hit-sequence, and the delta-T's must be known. Conventionally, the "threshold-crossing method" is used: when the *AE* signal crosses a preset threshold of a channel, this channel is considered "hit" by the *AE* burst, and the time of crossing is defined as the arrival time. Then the hit-sequence and delta-T's are deduced thereafter. Threshold setting is user-dependent and somewhat subjective, but usually it is set at **6** dB (two times) higher than background noise. Because of the random nature of noise, in practice the threshold-crossing method is susceptible to noise, therefore leading to erroneous location results.

To make our system more robust to noise, a series of digital signal processing techniques were used. First, a digital bandpass (BP) filtering was applied to each digitized waveform of AE signal, according to the nominal frequency of AE sensors used. For instance, a BP filter of 20-40 **kHz** is used for AE sensors of **30** kHz. Second, filtered *AE* signals were transformed to their corresponding analytic signals'. With an analytic signal, the information of amplitude vs. time is kept but the oscillation term is suppressed. Third, the thresholdcrossing method was applied to determine arrivals. Finally, the cross-correlation between two analytic signals was performed to give the relative arrival-time between two channels, then the hit-sequence and delta-T's were determined. Since the maximal value of cross-correlation indicates the relative time-shift between two main energy packets, correct hit-sequence and delta-T's can be deduced even in the presence of noises. Figure **6** shows a set of *AE* signals due to the leakage through a drilled hole in a leak experiment. *As* shown in Figure **6,** in source location calculation, it is a common practice to set a certain time window which at least is equal to the maximal possible arrival-time difference between *AE* sensors, and then analyze the signals within the time window only. This also reduces calculation time significantly. The corresponding normalized analytic signals within the time window are shown in Figure **7.** The noise was rather low, so the main energy packets (designated by **"P)** can be easily identified. Figure 8 shows another set of AE signals that are weaker and severely "contaminated" by noise. Since there are many "noise packets" (designated by "n" in Figure **91,** conventional threshold-crossing method will lead to incorrect result. However, by using cross-correlation the correct result still can be obtained.

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Fig. 6: A set of AE signals obtained in a leak experiment. The partial loss of couplant resulted in the lower signal amplitude of channel 6.

Fig. 7: The normalized analytic signals of AE signals within the time window in Figure 6. The main energy packets ("P") can be easily identified.

Fig. 8: Another set of AE signals obtained in a leak experiment

Fig. 9: The normalized analytic signals of AE signals within the time window in Figure 8. Many noises ("n") are present.

Fig. 10: Some location results of the AE signals in the leak experiment

3-D SOURCE LOCATION

When hit-sequence and delta-T data are available, calculations can be conducted to determine the location of an AE source. Various algorithms of AE source calculation have being developed, but the Spherical Interpolation (SI) method for 3-D location by Smith and Abel² was adopted. The SI method consists of a series of matrix computations, and is implemented with MATLAB. Some location results of the AE signals from leakage are shown in Figure 10, which is the projection of 3-D locations on the XY-plane. The calculated positions show good consistency and are very close to the leak hole.

CONCLUSIONS

A PC-based, fully digital and lab-scale acoustic emission system for 3-D source location was set up. To get good results, a sampling rate higher than the Nyquist frequency is preferred, and is at least five times the carrier frequency of signal for our experiment. With digital filtering, analytical signals, and cross-correlation technique, the source location operation can be made more robust to noise. Versatile 3-D source location calculation can be implemented with numerical and DSP software packages.

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