

國立交通大學

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碩士論文

複晶矽薄膜電晶體的充放電機制用於
記憶體應用之研究



**A Study of Trapping/De-trapping Mechanisms in Poly-Si
Thin Film Transistors for Memory Applications**

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中華民國九十七年六月

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摘要



在本論文研究中，我們在複晶矽薄膜電晶體傳導特性中觀察到一種有趣的遲滯(hysteresis)現象，並對此進行深入的探討。為了清楚的瞭解其發生的起因，本研究針對可能調控此現象的參數，包括：通道厚度、通道長度、通道寬度、電漿處理及溫度等，設計相關的實驗以解析其影響。根據實驗的結果，我們提出一物理模型，利用複晶矽薄膜電晶體中晶界充放電機制，說明在特定條件下造成此種遲滯現象的原因與過程。此外，本論文也探討不同材料與元件組態對於遲滯現象的影響。我們的研究也同時評估，複晶矽元件之遲滯特性對於將來運用於記憶體應用之可行性。

A Study of Trapping/De-trapping Mechanisms in Poly-Si Thin Film Transistors for Memory Applications

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Abstract

A new hysteresis phenomenon presenting in the current-voltage characteristics of poly-Si thin-film transistors (TFTs) was discovered, for the first time, in our experiments. Mechanisms reported in previous literatures fail to explain the finding. To clearly understand the root cause of the hysteresis, we examine several factors that might affect the phenomenon, including channel thickness, channel length, channel width, post-metal treatment, and temperature, etc. From the results, we propose a physical model, based on the trapping/detrapping events occurring in the grain boundaries of poly-Si thin film transistors (TFTs), to reasonably explain the hysteresis behavior of poly-Si TFTs. Besides, we also investigate the effects of Ge incorporation in the channel and structural factors on the hysteresis characteristics. Our findings indicate the potentiality of the poly-Si devices for future memory applications.

Acknowledgment

經過兩年的努力，即使途中經過了不少阻礙與挫折，最後還是將這篇論文給順利完成。在這過程中需要感謝的人太多，首先，我誠摯地感謝指導教授林鴻志博士和黃調元博士，兩位老師細心的教導與要求，使我在人生中最重要的一年中累積到許多寶貴的經驗，感謝林博指導我正確的研究方法與態度，讓我學習到更有效處理事情的態度，對於以後進入社會工作受益很大。感謝黃老師對我論文的修改，使得論文更佳的完善。

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誌於風城交大

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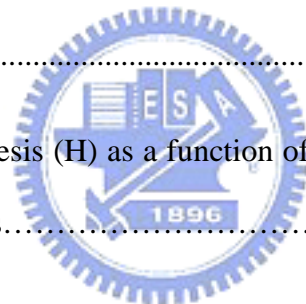
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Chapter 1

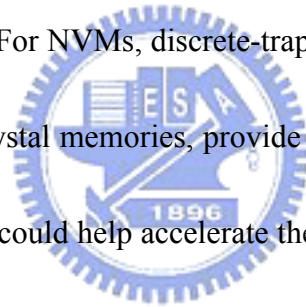
Introduction

1-1 Overview of Semiconductor Memory

The semiconductor market has been active for years. A significant portion of this market is given by the semiconductor memories. Semiconductor memories are usually classified into two major types: volatile and non-volatile memories. Volatile memories (VMs) lose data as the power is turned off. Most types of random access memory (RAM) fall into this category. RAM is further divided into dynamic RAM (DRAM) and static RAM (SRAM). DRAM requires refresh action to prevent charge loss from the storage capacitor through leakage current.

On the other hand, non-volatile memories (NVMs) can retain their stored information even after the power is turned off. There are many kinds of NVMs, including read only memory (ROM), mask read only memory (MASKROM), programmable read only memory (PROM), electrical programmable read only memory (EPROM), electrical erasable programmable read only memory (EEPROM), flash memory, etc. Owing to the requirements of increasing packing density, the devices size must be

scaled accordingly. In this regard, traditional VMs and NVMs have encountered certain problems. Fortunately, there are solutions to overcome these issues. For DRAM of VMs, to maintain sufficient charge storage capability as cell area is reduced, we can use high- k dielectric materials, such as $\text{Al}_2\text{O}_3 \cdot \text{Ta}_2\text{O}_5$ [1], to replace the dielectric of SiON in conventional DRAM capacitors, or use three dimensional structures to increase the effective capacitor area. Recently, a new concept of a capacitor-less 1-transistor dynamic memory (1T-DRAM) cell has been developed and it provides another possible approach for cell scaling. The detail of 1T-DRAM will be described in the next section. For NVMs, discrete-trap non-volatile memories, such as NROM, SONOS, and nanocrystal memories, provide particularly attractive feature of storing two bits per cell. This could help accelerate the effective scaling of NVM cells, and allows people to effectively cope with the difficulties in scaling the gate stack of traditional floating-gate VNM devices.



1-1-1 1T-DRAM

The demand for high density, large capacity, high speed, and low-power embedded memories are more important for system-on-chip (SoC) application. Embedded SRAM and embedded DRAM are the two types of embedded memories for practical application. The former can operate at a high speed but it occupies a

large area because it requires six transistors consisting of two p-channel MOSFET and four n-channel MOSFETs. Moreover, it is increasingly more difficult to scale down the 6T SRAM in consideration of matching among cell transistors. The latter, which integrates one transistor and one capacitor (1T/1C) in a cell, has a small cell size albeit slower operation speed. Nonetheless, one of the main issues lies in cell area reduction is the requirement of non-scalable capacitance of around 30fF/cell for maintaining sufficient operation margin.

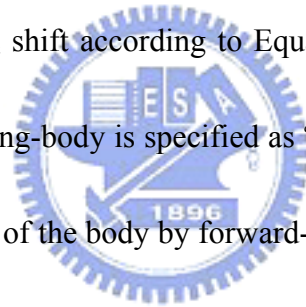
Recently, a new concept of a capacitorless 1-transistor dynamic memory (1T-DRAM) cell has been developed [2]-[12]. In this scheme, as shown in Fig. 1-1, there is no demand for a complicate storage capacitor, implying that the cell is process compatible with logic devices, in addition to a small area size. The features of these embedded memories are summarized in Table 1-1. Usually the device is built with an SOI MOSFET with its channel body floating. The memory information is stored in the floating body of the transistor. We know that a change on substrate bias (V_{BS}) may result in the threshold voltage (V_{th}) shift [13]:

$$\Delta V_{th} = \frac{\sqrt{2\varepsilon_s q N_A}}{C_o} \left(\sqrt{2\psi_B + V_{BS}} - \sqrt{2\psi_B} \right) \quad (1)$$

When the channel body is floating, excess carriers may be generated during device operation and are stored in the floating body, resulting in a change in both V_{BS} and V_{th} .

By use of the V_{th} shift, two distinct I-V conditions specifying “1” and “0” states are defined. Up to now, this concept has been demonstrated and realized on several architectures: partially-depleted (PD) SOI MOSFET [2]-[3], fully-depleted (FD) SOI MOSFET [4]-[5], MOSFET on bulk silicon [6], double-Gate SOI MOSFET [7], and FinFET [8].

Figure 1-2 schematically shows the operation principle of 1T-DRAM cell. The device is n-channel. Excess holes are generated by impact ionization [9] or GIDL [10]. These holes are stored in the floating-body and lead to an increase in substrate potential and the resultant V_{th} shift according to Equation (1). The condition when excess holes exist in the floating-body is specified as “1” state. On the other hand, the excess holes can be swept out of the body by forward-biasing the body-drain junction, and the device returns to the fresh condition denoted as “0” state.

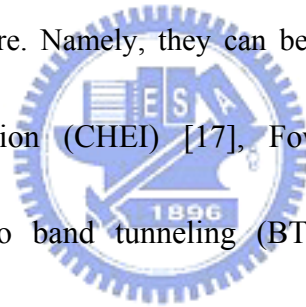


1-1-2 SONOS

High density NVMs are in high demand in the mobile and mass storage media market due to the low cost-per-bit needs. However, scaling-down of conventional NVM with a floating gate (FG) structure is expected to face a humongous barrier beyond 45-nm node due to strong interference effect and scaling limit in tunneling oxide thickness [14]. Silicon/oxide/nitride/oxide/silicon (SONOS) multi-layer

structure has been widely developed aiming to overcome the disadvantages of FG structure [15]. It differs from the floating gate device in that the injected electrons are trapped and stored in the thin insulating silicon nitride (Si_3N_4) layer rather than the thick conducting layer of polysilicon. Figure 1-3 shows a comparison of the cross-section between the two structures. One of the advantages of the charge trapping at Si_3N_4 layer in SONOS over the FG device is its better immunity to charge loss via defects (oxide traps, pin holes) in the surrounding oxide [16].

The operation principles for both program and erase operations in SONOS structure are similar to those in FG structure. Namely, they can be divided into three mechanisms, channel hot electron injection (CHEI) [17], Fowler-Nordheim tunneling (FN tunneling) [18] and band to band tunneling (BTBT) [19]. The goal of these mechanisms is to inject charged carriers to or extract those stored charges out from the insulator layer of Si_3N_4 . Because of the difference in the number of charges stored in the Si_3N_4 layer, distinctly different electrical transfer characteristics can be defined to represent the logic states.



1-2 Overview of Nanowire Technology

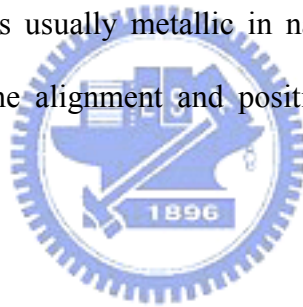
Moore's Law states that the number of transistors on a chip doubles every 18 months, and it has been a guiding principle for the semiconductor industry for over 30

years. To sustain Moore's Law, the device size must be continuously scaled down. However, as the devices reach 32 nm and beyond, traditional scaling methods performed on planar structure face more technological and fundamental challenges. Increasing demand on the resolution of the exposure system and extremely high process/tool cost push the conventional approach towards its practical limit [20]. The use of alternative 3D device structures is a possible approach to relax the aforementioned issues. Among the proposed nano-scale structures, Si nanowires (NWs) possess great potential as the active components in future nanoscale devices. Owing to its inherently high surface-to-volume ratio, tunable electronic transport properties [21]-[22], good thermal conductivity [23] and optical properties [24], NWs are attractive for many applications, including nano CMOS [25], memory devices [26], NW TFTs [27], biosensors [28]-[29], photodetectors [30], lasers [31], etc.

The preparation of NWs can be classified into two approaches, namely, "top-down" and "bottom-up". The top-down approach usually requires advanced lithographic tools such as deep UV steppers [25] and e-beam writers [32] to generate NW patterns. Although top-down methods are well developed and mature for mass production, these methods become increasingly costly and/or slow as feature size moves deeper and deeper into the nanometer regime. A great deal of works has been invested in developing alternative top-down patterning methods such as probe

microscopy methods [33], nanoimprint lithography [34], and transfer lithography [35].

Unlike the top-down methods which define key nanometer-scale metrics by lithography, the bottom-up approach employs chemical reactions to synthesize NWs. Several techniques to synthesize NWs have been developed, including thermal evaporation [36], vapor-liquid-solid (VLS) method [37], laser ablation technique [38], solid-liquid-solid (SLS) method [39], and so on. The most popular one is catalyzed VLS methods. The VLS process is shown in Fig. 1-4 [40]. NWs can be synthesized in single-crystalline form with precisely controlled structures, diameters and lengths, chemical compositions and doping properties using this catalyzed VLS method. However, since the catalyst is usually metallic in nature, metal contamination is a great concern. In addition, the alignment and positioning of NWs represent other issues for mass-production.



1-3 Motivation

As mentioned earlier in this chapter, the charges are stored in the insulator layer of Si_3N_4 for SONOS, or in the floating body for 1T-DRAM to define the logic states. Recently, while studying the electrical transfer characteristics of polycrystalline silicon (Poly-Si) thin-film transistors (TFTs), we discovered accidentally a phenomenon similar to that described for the operation of SONOS and 1T-DRAM, albeit obviously of different mechanism. The purpose of this study is to investigate and understand the root cause and physical meanings of such phenomenon, and

evaluate its potential in practical memory applications.

1-4 Thesis Organization

The organization of this thesis is divided into five chapters. A brief overview of semiconductor memory and nanowire technology is given in this chapter. In Chapter 2, we briefly describe the device structures and process flow for fabricating the test structures, including planar and nanowire devices. In Chapter 3, the hysteresis mechanism is described and discussed, and the electrical characteristics as well as program/erase characteristics of planar structure with poly-Si channels are investigated. Effects of several factors including channel thickness, bias conditions, temperature, and plasma treatment, are also investigated and analyzed. In Chapter 4, a comparison of device characteristics between planar and nanowire structure, characteristics of double-gate configurations, and influences on hysteresis windows including channel materials, channel length and width, geometry effect, bias condition effect, are investigated and discussed. Finally, a summary of conclusions achieved in this thesis and some suggestions for future work are given in Chapter 5.

Chapter 2

Device Fabrication

2-1 Device Structure and Process Flow

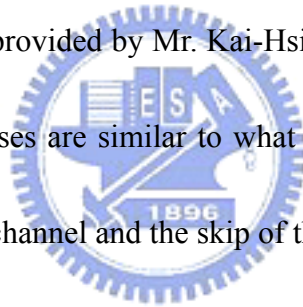
In this chapter, we will introduce two kinds of TFTs structures, i.e., planar and nanowire inverse-T gate structure, and describe the fabrication process in details.

2-1-1 Planar Structure



The process steps for the fabrication of planar poly-Si TFTs are illustrated in Fig. 2-1. First, a 100nm-thick in-situ n^+ poly-Si layer was deposited on 6-inch silicon wafers to serve as the bottom electrode (Fig. 2-1(a)). Then, a SiO_2 layer of 20 nm, 50 nm, or 200 nm, respectively, was deposited by low-pressure chemical vapor deposition (LPCVD) to serve as the bottom dielectric according to the split conditions shown in Table 2-1. Next, a 30 nm, 50nm and 1000 nm-thick amorphous-Si (a-Si) layer was individually deposited by LPCVD, followed by a solid phase crystallization (SPC) step performed at 600°C in N_2 ambient for 24-hours to transform the a-Si into poly-Si (Fig. 2-1(b)). Afterwards, the active region was formed by I-line lithography

and subsequent etching. This is followed by the deposition of an LPCVD oxide layer of 20nm to serve as the top gate dielectric. A 100nm-thick in-situ-doped n^+ poly-Si was then deposited and patterned to serve as the top gate electrode (Fig. 2-1(c)). Phosphorous S/D doping was implanted with a dosage of $5 \times 10^{15} \text{ cm}^{-2}$ at the energy of 35KeV (Fig. 2-1(d)) with the top gate serving as the self-aligned mask. A 350 nm-thick TEOS oxide by LPCVD was then deposited to serve as the passivation layer (Fig. 2-1(e)). Finally, the fabrication was completed after the formation of test pads using standard metallization steps (Fig. 2-1(f)). In this work we also characterized poly-Si_{1-x}Ge_x TFT which are provided by Mr. Kai-Hsiang Chang, also of our research group. The fabrication processes are similar to what we had described above except for the use of a poly-Si_{1-x}Ge_x channel and the skip of the bottom electrode formation.



2-1-2 Inverse-T double-gate NWTFT

These devices are provided by Mr. Hsin-Hwei Hsu of our group. Top views and cross-section views of inverse-T double gate NWTFT are shown in Fig. 2-2. The process steps for the fabrication of these devices are illustrated in Fig. 2-3. First, 6-inch silicon wafers capped with 100 nm-thick SiO₂ were used as the starting substrates (Fig. 2-3(a)). Then, a 150 nm-thick in-situ n^+ poly-Si layer was deposited. The poly-Si layer was subsequently patterned by twice applying G-line lithography

and subsequent dry etching steps to fabricate the inverse-T gate structure (Figs. 2-3(b), (c)). Note that the first etch step removed 100nm-thick poly-Si layer to form the upper part of the inverse-T gate. While the remaining 50nm-thick poly-Si outside the inverse-T gate regions was completely removed in the second etch step. A 20nm-thick SiO₂ was grown on the inverse-T gate by LPCVD to serve as the gate dielectric. Then, a 100 nm-thick a-Si layer was deposited by LPCVD. Next, an annealing step was performed at 600°C in N₂ ambient for 24-hours to transform a-Si into poly-Si (Fig. 2-3(d)). Subsequently, S/D implant was performed by P⁺ implantation at 15 keV and $5 \times 10^{15} \text{ cm}^{-2}$ (Fig. 2-3(e)). Note that the implant energy was sufficiently low so that the implanted dopants were mostly distributed near the top surface of the poly-Si layer. Then, S/D photoresist patterns were formed on the substrate by a standard lithography step. A reactive plasma etch step was subsequently employed to remove poly-Si layer except at the S/D regions and the nanowire channels were simultaneously formed in this step in a self-aligned manner (Fig. 2-3(f)). Note that the etching time was carefully controlled to make sure that only two NW channels remained on the corner of inverse-T structure and the channel size was determined by various over-etching time. Afterwards, the S/D dopants were activated by an annealing treatment. Then, a 20nm-thick SiO₂ was deposited by LPCVD to serve as the top gate oxide. Subsequently, a 100nm-thick in-situ-doped n⁺ poly-Si was deposited and patterned to

serve as the top gate electrode (Fig. 2-3(g)). A 200nm-thick TEOS oxide was then deposited by LPCVD to serve as the passivation layer. Finally, the fabrication was completed after the formation of test pads using standard metallization steps.

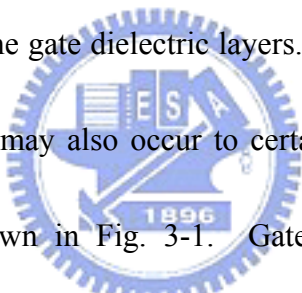


Chapter 3

Electrical Characteristics of Planar TFTs

3-1 Observation of Hysteresis in polycrystalline silicon

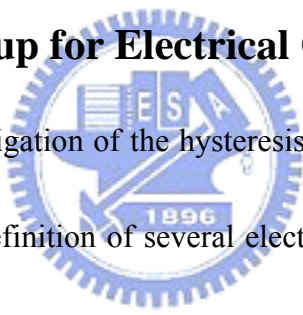
Several hysteresis phenomena have been reported and discussed in many reports [41-44]. In these reports, the origin of the hysteresis could be due to the trapping of charges inside at the gate dielectric layers. In our study, accidentally, we found that such phenomenon may also occur to certain poly-Si thin-film transistors (TFTs). An example is shown in Fig. 3-1. Gate voltage (V_g) follows A-B-A sequence, as indicated by the arrows shown in the figure. The loop is anti-clockwise.



In the following, “forward sweeping (FS)” refers to the sweeping of gate voltage from negative to positive, and “reverse sweeping (RS)” means an opposite direction sweeping. But such hysteresis phenomenon can not be explained by the mechanisms identified in the previous reports [41-44]. As mentioned previously, the hysteresis caused in the operation of SONOS is due to electron trapping and de-trapping events in the Si_3N_4 trapping layer. But the electron trapping causes a positive shift in V_{th} , resulting in a clockwise loop, which is different than that shown in Fig.3-1. On the

other hand, the operation of 1T-DRAM shows similar anti-clockwise loop. However, its mechanism is also not suitable for explaining our experimental results. The reason is that the operation principle of 1T-DRAM relies on impact ionization [9] or GIDL [10] to generate the excess holes accumulating in the floating body of the SOI channel. This is obviously not the case in Fig.3-1, in which a small drain voltage of 0.1V is applied. In this situation, the excess holes generated by impact ionization [9] or GIDL [10] are not feasible.

3-2 Measurement Setup for Electrical Characterization



Before go into the investigation of the hysteresis phenomenon, we first describe the measurement setup and definition of several electrical parameters in this section. Electrical characteristics of fabricated devices are mainly characterized by the automated measurement setup constructed by an Agilent™ 4156A semiconductor parameter analyzer and Interactive Characterization Software (ICS) software. During all measurements, the temperature was controlled at a stable value by temperature-regulated hot chuck.

Based on the transfer characteristics measured at $V_d=0.1V$, the parameters of the TFTs including field-effect mobility (μ_{FE}), threshold voltage (V_{th}), subthreshold swing ($S.S.$), were extracted according to their definition.

Here, the field-effect mobility (μ_{FE}) is determined by

$$\mu_{FE} = \frac{Lg_m}{WC_{OX}V_d}$$

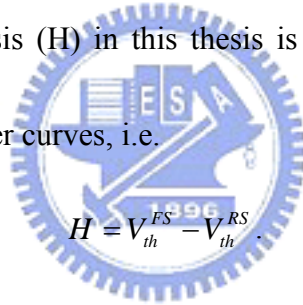
where g_m is the maximum transconductance and C_{OX} is the gate oxide capacitance per unit area.

The threshold voltage (V_{th}), calculated by the constant current method, is defined as the gate voltage (V_g) need to achieve a drain current (I_d) of $(W/L) \times 100nA$, i.e.

$$V_{th} = V_g @ I_d = \frac{W}{L} \times 100nA,$$

where L and W are the channel length and width respectively.

The width of the hysteresis (H) in this thesis is defined as that the shift of V_{th} between the FS and RS transfer curves, i.e.



$$H = V_{th}^{FS} - V_{th}^{RS}$$

Finally, the subthreshold swing ($S.S.$) can be calculated from the subthreshold current in the weak inversion region by

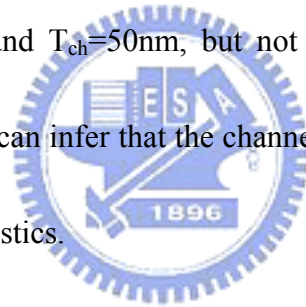
$$S.S. = \frac{\partial V_g}{\partial (\log I_d)}$$

3-3 Effects of Poly-Si Channel Thickness

To more deeply understand the origin of the hysteresis in the poly-Si TFTs, we designed several splits of experiments and investigated the effect of various device structural parameters and/or process treatments. Note that most of the devices we

characterized in the thesis received no post-metal treatment such as NH_3 -plasma or H_2/N_2 mixed gas anneal. So in the following sections, unless we particularly investigated the effect of post-metal treatment, the presented results were obtained from devices without post-metal treatment.

First, we changed the channel thickness (T_{ch}) to understand whether the different T_{ch} would influence the hysteresis formation or not. Figure 3-1, 3-2, and 3-3 show the transfer characteristics of poly-Si TFTs with T_{ch} of 50nm, 100nm, and 30nm, respectively. It can be obviously seen that the hysteresis phenomenon exists in the poly-Si TFTs of $T_{\text{ch}}=30\text{nm}$ and $T_{\text{ch}}=50\text{nm}$, but not in the device with $T_{\text{ch}}=100\text{nm}$ devices. From the results, we can infer that the channel thickness plays a key factor to affect the hysteresis characteristics.



3-4 Effects of Plasma Treatments

The electrical properties of poly-Si TFTs are very different from the traditional MOSFETs. This is due to the presence of abundant grain boundaries and defects in TFTs, which result in large threshold voltage (V_{th}), poor subthreshold swing, low mobility, and high leakage current. It is well known that post-metal treatments effectively reduce both the tail and deep defect states inside the bandgap [45-51], and hence enhance the performance of TFTs. There are many kinds of post metal

treatment methods such as forming gas annealing (denoted as FG) in furnace [45], plasma treatments in hydrogen or other ambient such as NH_3 [46], N_2/H_2 [47], H_2/O_2 [48] or deuterium [49], high pressure of water vapor treatment [50], or water passivation [51]. By these approaches, defects like dangling bonds at grain boundaries and oxide/channel interface are passivated. As a result, the performance of TFTs improves, including lower subthreshold swing, higher on/off ratio, and higher carrier mobility.

In this work, we employed FG annealing in furnace and NH_3 plasma treatment in PECVD to investigate the effect of post-metal treatment, respectively. For the case of FG annealing, the devices were exposed in N_2/H_2 (5%) ambient at 400°C for 30 minutes. For the case of plasma treatment, the devices were put in a diode NH_3 plasma with RF power of 200 watt at 300°C for 1 hour.

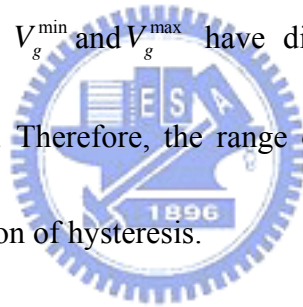
Figure 3-4 shows the transfer characteristics of the devices which received FG annealing. After annealing the device exhibits improved performance in terms of better subthreshold swing, larger on-state current. Moreover, the hysteresis window disappears after FG annealing, as observed in Fig.3-4 (a). The major performance parameters for the FG annealing are listed in Table3-1. When applying larger gate voltage range (in this case $V_g = -3\text{V} \sim 8\text{V}$), we found that the hysteresis appears again but with a much smaller hysteresis window as compared with the windows before FG

annealing, as observed in Fig.3-4 (b). This is ascribed to the relatively low concentration of H₂ species in FG annealing, or the strength of certain passivated Si-H bonds is not strong and will be broken when the gate voltage is sufficiently high. To demonstrate this conjecture, we also employed NH₃ plasma treatment in PECVD for 1hr. The diffused hydrogen and nitrogen species have been reported to more effectively passivate the trap states of poly-Si film. The transfer characteristics of the devices, which received NH₃ plasma treatment in PECVD for 1hr, are shown in Fig.3-5. Similar to the case of FG annealing, improvement in device performance is observed after the plasma treatment. The device parameters are list in Table3-2. However, the hysteresis window disappears even under high gate voltage bias (see Fig.3-4(b)). From the results, we can infer that the post-metal treatments and different bias condition may play another role in influencing the hysteresis phenomenon.

3-5 Effects of Bias Conditions

From the results we observed in previous sections, we got some hints that different gate sweeping range (V_g^{\min}, V_g^{\max}) may have influences on the hysteresis window. Figure 3-6 shows the results for the gate voltage dependence of hysteresis window in poly-Si TFTs. Transfer curves were measured by applying gate voltages from -3V to a various maximum voltage ranging from 3 to 9V. It appears that, with

increasing gate voltage, the transfer characteristics of RS are shift negatively and widen the hysteresis window. Moreover, in Fig.3-6 (b), it is seen that when the applied gate voltage reaches or larger than a specific value (around 8V in the case), the window width of hysteresis would saturate. Figure 3-7 also shows the measured transfer curves by applying the gate voltages starts at a value ranging from -5, -3, to -1V, and stop at 6V. It is seen that the transfer characteristics of RS curves of the three conditions coincide together; however, the transfer characteristics of the three FS curves are diverse. The outcomes are very dissimilar to the results as observed in Fig.3-6. To briefly sum up, V_g^{\min} and V_g^{\max} have different effects on the transfer characteristics of FS and RS. Therefore, the range of the gate voltage may also a factor to influence the formation of hysteresis.



3-6 Effects of Temperature

Up to now, we have already obtained a few clues which would cause the hysteresis phenomenon. To be worthy of noting, the former measurements were all done at room temperature. In this section, we further investigate the temperature dependence of hysteresis to acquire more information about the formation of hysteresis. Figure 3-8 depicts the transfer characteristic curves measured at temperatures of 25°C, 50°C, 75°C, and 100°C. The increase in temperature tends to

increases the off-state leakage and narrow the hysteresis window. It is also observed that the subthreshold swing of FS curves increases with increasing temperature, but opposite trend observed for the RS curves. The results clearly evidence the effectiveness of temperature in affecting the hysteresis.

3-7 Mechanism for the Occurrence of Hysteresis

Up to now, the hysteresis phenomena have been observed under some circumstances discussed in our previous sections. Four important observations are summarized as follows:

(I) The hysteresis phenomenon occurs as the poly-Si thickness is sufficiently thin.

(Sec.3-3)



(II) The hysteresis phenomenon occurs if poly-Si TFTs skip the post-metal treatments

such as FG annealing and NH_3 plasma treatment. (Sec.3-4)

(III) The window width of the hysteresis and the subthreshold characteristics of FS

and RS depend on both maximum and minimum sweeping gate voltage

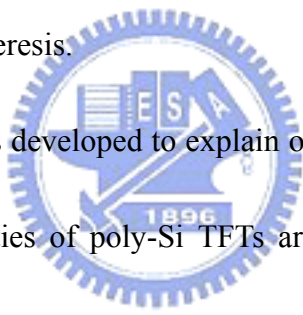
(V_g^{\min}, V_g^{\max}) . (Sec.3-5)

(IV) Window of the hysteresis shrinks as the measurement temperature increases

(Sec.3-6).

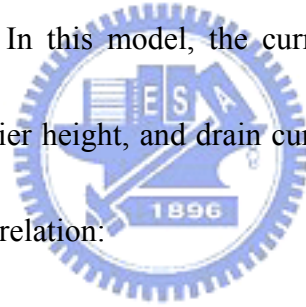
We've surveyed the literature but found no papers reported similar phenomenon in

poly-Si TFTs. However, this is reasonable based on the above observations. First most of papers reported previously studied the poly-Si TFTs with channel thickness around or thicker than 100nm [52]. In some papers the channel thickness is indeed around or thinner than 50nm, but the hysteresis phenomenon is still lacking [53, 54]. This is attributed to the process sequence of device fabrication. In order to obtain high-performance poly-Si TFTs, it is necessary to reduce the trap-states of the polysilicon films. As a result almost all previously published works [45-51] employed the plasma treatments to passive the defects in polysilicon films. Hence, it is hard to discover the formation of hysteresis.



In this section, a model is developed to explain our interesting finding in poly-Si TFTs. The electrical properties of poly-Si TFTs are different from the traditional MOSFETs. This is due to the presence of grain boundaries (GBs), which result in large threshold voltage (V_{th}), lower mobility, and high leakage current. Consider an n-channel poly-Si TFT. Assume that the GBs run perpendicular to the channel and no intentional channel doping is performed. So we limit our discussion to a device with a doping concentration below the critical value of N^* ($N^*=N_T/L_g$), in which N_T is the concentration of traps located at the GBs, and L_g is the grain size. The presence of the active traps at GBs provides deep-level trap sites for both electrons and holes and induces band bending with energy barrier height (E_B) for carriers. The energy barrier

height is proportional to the defect density (N_T) at GBs and to the inverse of the electron concentration (n) induced in the channel by the gate bias [55]. When the gate voltage is gradually applied, the surface potential Φ should vary along not only the film depth but also the channel direction. Although the electrostatic distribution is a three-dimensional problem, for the purpose of calculating its transport properties, it is sufficient to treat the problem in one dimension as the channel length and width are much larger than the grain size. The associated Poisson equation has been analyzed and calculated [55, 56]. Levinson *et al.* [57] proposed a model for carrier transport based on Seto's [55] theory. In this model, the current is governed by thermionic emission above the GBs' barrier height, and drain current (I_d) and the gate bias (V_g) are related with the following relation:



$$I_d = \frac{W}{L} \mu_o C_{ox} (V_G - V_{th}) V_{ds} \exp\left(-\frac{q^3 N_T^2 t}{8 \epsilon_s \epsilon_o k T C_{ox} (V_g - V_{th})}\right) \quad (2),$$

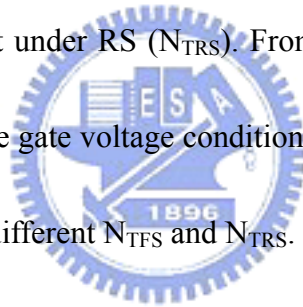
where

$$V_{th} = \phi_s + \frac{\sqrt{2qN_{eff} \epsilon_s \epsilon_o \phi_s}}{C_{ox}} + V_{FB} \quad (3),$$

$$\text{and } N_{eff} = N_D + \frac{2qN_T}{L_g} \quad (4).$$

Here, L and W are the channel length and width; C_{ox} is the oxide capacitance per unit

area; t is the thickness of the channel layer; V_{th} is the threshold voltage, and is specified in terms of the effective doping N_{eff} and the surface potential Φ_s . N_D is the concentration of shallow donors, V_g and V_{ds} are the gate voltage and drain voltage, respectively. The equation is valid in the subthreshold regime of the transistor. According our experimental results, the dissimilar transfer characteristics between FS and RS curves which form the hysteresis window can be attributed to the different trap densities of GBs (N_T) in the opposite sweeping direction. The V_{th} of FS is larger related to RS as observed in Fig.3-1. This implies that the trap densities at GBs under FS (N_{TFS}) are higher than that under RS (N_{TRS}). From equations (2) and (3), we can understand that under the same gate voltage condition, the drain current of FS (I_{dFS}) is smaller than RS (I_{dRS}) due to different N_{TFS} and N_{TRS} .



To turn on an n-channel poly-Si TFT, the traps under the quasi-Fermi level in the depletion layer of the surface channel need to be filled by the electrons. This explains why the V_{th} of poly-Si TFT is larger than traditional MOSFET. Normally, during the RS process, the trapped electrons at the GBs should be immediately de-trapped from the trap sites so that the transfer curve will follow the FS curve and hysteresis does not exist. An example is shown in Fig.3-2 for the device with 100 nm-thick poly-Si channel layer. On the other hand, hysteresis is obviously observed for the devices with channel thickness T_{ch} of 30nm and 50nm. This indicates that trapped electrons in the

FS process may retain in the trap sites during the RS process, resulting in a lowering in the V_{th} and the formation of hysteresis. According to one of our previous publications [58], the depletion width (W_{dep}) was estimated to be around 60nm for poly-Si layers formed with a scheme identical to this work (e.g., SPC at 600°C in N_2 ambient for 24-hours). Correlating this with results of Figs.3-1 ~3-3, it is clear that W_{dep} plays a key role for the occurrence of hysteresis. When the channel thickness is smaller than W_{dep} in the channel, the channel becomes fully depleted as the device is turned on; conversely, when the channel thickness is larger than W_{dep} in the channel, the channel is partially depleted, and in the channel outside the depletion region the potential is not disturbed by the gate voltage, so this region remains neutral. The energy band diagrams for fully and partially depleted conditions at flat-band and on-state modes under FS, and RS is illustrated as shown as Fig. 3-9 and 3-10.

In the case of partially-depleted case ($T_{ch} > W_{dep}$), under the process of FS, the energy bands near the semiconductor surface are bent downward (Fig.3-9 (b)). In the neutral region ($x > W_{dep}$), there are many empty trap states at the GBs with levels distributed in the energy band gap. Nevertheless, in the depletion region ($x < W_{dep}$), most of the trap sites under the Fermi level are filled with the induced electrons. Subsequently, when the RS operation is executed and the bending of the energy band in the depletion region gradually recovers and shifts back to the flat-band condition.

In the meantime, the trapped electrons tend to de-trap from the trap sites. Three possible de-trapping paths are identified in the figures. The first path (the path 1 denoted in Fig.3-9 (c) and Fig.3-10 (c)) is that the trapped electrons at the GBs are released from the trap site to the conduction band by the thermionic emission. The second path (the path 2 denoted in Fig.3-9 (c)) is that the trap-to-trap conduction via the traps located in the neutral region of GBs, and then the electrons move to the neutral region would recombine with holes therein. Consider the energy barrier for paths 1 and 2, we find that energy needed for path 1 ($\sim 1/2 E_g$ for deep levels) is bigger than that for the path 2, thus path 2 is more feasible. The third path (as denoted in Fig.3-10 (d) as path 3) is that the de-trapping of electrons to the intra-grain region. Similarly, it requires a sufficiently high energy for the electrons trapped in the deep levels to leave. Since the path 2 is feasible and efficient as discussed above, we believed it is the major de-trapping path, and the transfer curve under RS operation can promptly follow the FS curve and eliminate the hysteresis, as observed in the devices with a thick poly-Si channel (e.g., Fig.3-2).

In the case of fully-depleted mode ($T_{ch} < W_{dep}$), the path 2 is hindered by the buried oxide, however. So the available de-trapping paths are paths 1 and 3, both of which require a sufficient energy (or a period of time to acquire the energy) to occur for electrons trapped in a deep level. Hysteresis phenomenon is thus resulted. This

well explains why the hysteresis we observed in the devices with $T_{ch}=30\text{nm}$ (Fig.3-3) and $T_{ch}=50\text{nm}$ (Fig.3-1).

We can examine the above model by extracting the trap density. If the model is correct, portion of the traps (i.e., those deep states trapped with electrons) become inactive during the RS process. Assume the amount of charges in those inactive traps is Q_T (Coul/per unit area), the effective trap density in RS case, N_{TRS} , is smaller than that in FS mode, N_{TFS} , with the following relation:

$$N_{TFS} = N_{TRS} + \frac{Q_T}{q} \quad (5).$$

For poly-Si TFT, the subthreshold swing (SS) is affected by the effective trap density, N_T , with the following equation [59]:

$$\begin{aligned} SS &= \frac{kT}{q} \times \ln 10 \times \left(1 + \frac{C_T + C_S}{C_{ox}} \right) \\ &\approx \frac{kT}{q} \times \ln 10 \times \left(1 + \frac{C_T}{C_{ox}} \right) \quad \text{where } C_T = q \times N_t \end{aligned} \quad (6).$$

where C_S is the capacitance of interface states at the channel/oxide interface, which is usually negligible in poly-Si TFT. For the results shown in Fig.3-1, the SS are around 460mV/dec and 370mV/dec for the FS and RS I-V curves, respectively. Such difference becomes reasonable with the inference of (5) to (6): SS is reduced owing to the existence of inactive trap states. According Eq.6, an inactive trap density, ΔN_T (equal to $\frac{Q_T}{q}$), is extracted to be $1.61 \times 10^{12} / \text{cm}^2$.

The trapped charges could not only alter the SS, but also the threshold voltage

according to the following relation:

$$V_{th}^R - V_{th}^F = \Delta V_{th} = -\frac{(Q_T)}{C_{ox}} \quad (7).$$

With this relation, we can approximately estimate the amount of the trapped charges exist during the RS measurement. We used two ways to define the V_{th} , one is the constant-current method described in Sec.3-2; the other is the intercept of gate voltage of the linear drain current-Vg plots. For the results shown in Fig.3-1, the threshold voltage defined by constant-current method ($V_{th} = V_g @ I_d = \frac{W}{L} \times 100nA$) for the FS and RS curves are $V_{th}^F = 1.41V$ and $V_{th}^R = -0.21V$, respectively. The ΔN_T is thus estimated to be around $1.46 \times 10^{12} / cm^2$. The linear-Id vs. Vg plot is shown in Fig.3-11. The extracted threshold voltage is $V_{th}^F = 1.83V$ for FS mode and $V_{th}^R = 0.29V$ for RS mode, and ΔN_T is calculated to be around $1.66 \times 10^{12} / cm^2$. The results based on the threshold voltage difference are very well consistent with that obtained with SS difference $1.61 \times 10^{12} / cm^2$. These estimated values are also close to that reported in the literature [55-57].

In Sec.3-4, we have discovered that the post-metal treatments have strong influence on the hysteresis. When the devices received post-metal treatments such as FG annealing and NH_3 -plasma treatment, the deep-level trap density N_T is expected to be reduced due to the defect passivation mechanism. The hysteresis window is thus largely shrunk, as shown in Figs.3-4 and 3-5.

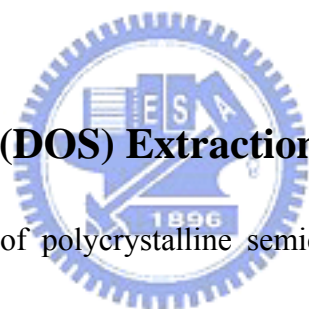
In Sec.3-5 section, we found that variation in the gate sweeping range (V_g^{\min} and V_g^{\max}) may have effects on the transfer characteristics of FS and RS, as well as the hysteresis window. By means of our model, we can further realize the effects of V_g^{\min} and V_g^{\max} . According our measurement sequence, the sweeping follows the A-B-A sequence as shown in Fig.3-1. FS (A to B) “programs” the device by filling the deep-level traps with electrons and the amount of the stored charges is determined by the V_g^{\max} . From this argument, it becomes reasonable to understand the results illustrated in Figures 3-6 (a) and (b). The increase in hysteresis width with increasing V_g^{\max} is attributed to the increase in the number of electrons trapped in deep levels, since a higher V_g^{\max} may result in a higher band bending and the amount of deep levels under the Fermi level in the channel would increase accordingly.

On the other hand, to erase the programmed device, the gate voltage applied in the RS event must be sufficiently low to render an upward band-bending in the channel, so that the stored electrons can be effectively expelled out from the stored sites. Thus the extent of erasing is determined by V_g^{\min} . In Fig.3-7, transfer curves of a device with fixed V_g^{\max} but various V_g^{\min} are measured. We can see that the RS curves coincide due to the fixed V_g^{\max} . However, the V_g^{\min} used in the second sweeping (-3V) is not sufficiently low to complete wipe off the stored charges in the channel, so the third sweeping starting at -1V shows a smaller threshold voltage in the

FS curve as compared with the former two sweeping.

Figure 3-12 shows the transfer characteristics of devices under consecutive 20 times of measurements following the A-B-A cycles. In Fig.3-12 (a) V_g^{\max} and V_g^{\min} are 8V and -3V, respectively. It can be seen that there exists a positive shift in the transfer curves with increasing cycle (indicated by the arrow). This is attributed to the degradation caused by the high V_g^{\max} . This issue could be solved by reducing the V_g^{\max} . An example is shown in Fig.3-12 (b), the shift in transfer becomes negligible as the V_g^{\max} is reduced to 6V.

3-8 Density-of-States (DOS) Extraction



The electrical properties of polycrystalline semiconductor devices are strongly affected by defects such as dangling bonds and strained bonds located at the GBs or inside the grains. A comprehending of the intrinsic, energy distribution, density, and behavior of these defects is important for development of polycrystalline TFTs. Therefore, characterization and analysis of the density of gap states (DOS) in the poly-Si is essential for understanding the devices characteristics. Moreover, accurate DOS is required for the device modeling and simulation of poly-Si TFTs. In this regard, there are several methods to determine DOS in polycrystalline devices such as capacitance-voltage (CV) method [60], optical absorption [61], doping dependence of

conductivity [62], and field effect conductance (FEC) [63] method, etc. Owing to the fact that the FEC method could be directly applied to the TFT structure characterized in this study, we adopt the technique to determine DOS. First the transfer characteristics of both FS and RS are measured as a function of temperature, as shown in Fig.3-8. The flat-band voltages then can be obtained using temperature method with equation 8 [64].

$$\frac{d \log G}{dV_g} \cong \frac{\epsilon_{ox}}{t_{ox}} \cdot \frac{1}{qkTN_o} \left[1 + \frac{1}{2} \left(\frac{q\psi_s}{kT} \right) + O \left(\frac{q\psi_s}{kT} \right)^2 \dots \right] \quad (8).$$

From the above equation, the flat-band voltages can be determined as the V_G when $T \cdot \left(\frac{d \log G}{dV_g} \right)$ is temperature independent. Once the flat-band voltages are obtained, the relationship between gate voltage V_g and surface potential, ψ_s , could be established by incremental method [62] with the following relation

$$\psi_{s,i+1} = \psi_{s,i} + \frac{G_{i+1} - G_i}{G_o} \cdot \frac{d}{t_{ox}} \cdot \frac{\epsilon_{ox}}{\epsilon_{Si}} \cdot \frac{V_{g,i} - V_{FB} - \psi_{s,i}}{\exp \left(\frac{q\psi_{s,i}}{KT} \right) - 1} \quad (9).$$

Finally, the DOS of a given energy inside the band gap related to the surface potential ψ_s can be expressed as follows:

$$DOS(E_F + \psi_s) = \frac{\epsilon_{Si}}{2q} \frac{\partial^2}{\partial \psi_s^2} \left(\frac{d\psi}{dx} \Big|_{x=0} \right)^2 \quad (10)$$

The extracted DOS results of FS and RS are shown in Fig.3-13. It is observed that the DOS of FS and RS remains the same near the band edge, but have a significant

difference in the mid-gap regime. As the previous literatures reported, the deep-level states are primarily contributed by the silicon dangling bonds locating at the GBs [45,61]; while the tail states are ascribed to strained bonds or structure disorder near the grain boundaries or interface of semiconductor and oxide [65, 66]. However, the tail and deep states have different effects on the performance of transfer characteristics. Deep-level states dominate the property of the subthreshold regime such as subthreshold swing and V_{th} , while tail states control the above-threshold on current regime such as on current (I_{ON}) and mobility. This is because the Fermi level at the channel is in the proximity of the mid-gap in the subthreshold region, whereas it is closer to the edge of the conduction band above the threshold. The results of DOS are shown in Fig.3-13, it is clearly seen that under the operation of FS and RS, the tail trap states are almost equal, but the deep states have significant difference. This can reasonably explain the hysteresis phenomenon observed in this thesis. In Fig.3-10, the electrons trapped in the tail states are easy to escape via path 1 or path 3 since their energy level is close to the conduction band. Based on the results shown in Fig3-13, we conclude the difference in trap density between FS and RS, i.e. N_{TFS} and N_{TRS} , mainly originates from the deep states.

3-9 Program/Erase Characteristics

Further, the shift of the V_{th} for V_g increase and decrease, i.e., a hysteresis in the transfer characteristics has been observed and can be translated into memory effects for the application of the memory devices. To investigate its feasibility in this aspect, we also study the characteristics of programming/erasing operation, and identify the retention property, which is a measure of the ability that a memory cell can retain the stored information (or the charges in the present case).

In this section, we employed two approaches for the purpose of programming and erasing operations. One is applied with a pulse of high voltage to the gate, and the other is to sweep the gate voltage from a low value to a sufficiently high value for programming, and *vice versa* for erasing, similar to the procedure shown in Fig.3-7.

First we investigate the P/E characteristics using the gate-pulse method. The transfer characteristics of a test device under FS and RS modes are shown in Fig.14(a). Here we select the reading gate voltage at 0V, and the two corresponding points in the FS and RS curves as denoted as Q and R, respectively, in Fig.3-14 (a). We also set the gate pulse with high level, V_{gH} (=charging voltage), to 6V, and low level (V_{gL}) to 0, as shown in the inset of Fig.3-14 (a). Figure 3-14 (b) shows the time dependences of drain current (I_D) (in linear scale) of the device applied with pulse width $t_w=1, 0.1, 0.01,$ and 0.001 sec. Note that the current at P in Fig.3-14 (a) corresponds to the

programming current under static case. Since it is not easy to tell the major difference among the cases having different pulse-width, we further illustrate results of the first 9 seconds Fig.3-14 (c) by expressing the drain current in log-scale. The current difference before and after the gate pulse programming operation becomes clear, and such difference depends on the pulse width applied. When the pulse width reduces, the discrepancy between two states lowers. It is even negligible as the gate pulse shrinks to about $t_w=1\text{ms}$. This is owing to the fact that the device structure has not been optimized, and thus the programming is actually limited by the high RC time delay due to the significant parasitic resistance and capacitance components contained in the device (and, likely, the measurement instruments).



Figure 3-15 shows the results of retention by monitoring the drain current at reading voltage of 0V at room temperature and 85 °C. The programming pulse gate voltage is $V_{gL}=0\text{V}$, $V_{gH}=6\text{V}$, and pulse width is $t_w=1$ sec. Obviously, the retention is significantly degraded as the temperature is raised. According to our model shown in Fig.3-10, the trapped electrons at GBs are easy to escape via path 1 or path 3 at high temperature, owing to the increase in thermal energy. Figure 3-16 shows the time dependence of drain current under the alternate programming and erasing operation. Two distinct logic states with different reading current are clearly illustrated in the figure.

The next we investigate the P/E characteristics with sweeping gate voltage method. As shown in Fig.3-7, portion of the stored electrons would remain in the trap sites if the gate voltage of the RS curves stopped at a value larger than the specific voltage for completely discharging the traps (around -3V in the present case). This hints that the stored electrons could be retained in the channel for a period of time unless a discharging at $V_g^{\min} = -3V$ is applied for effective discharging (erasing). By this mechanism, we can simply program our devices to “1” state (with a low threshold voltage) by means of single sweeping of gate voltage along the forward direction to a high gate voltage (e.g., 6V in the present case), and erase the device with RS operation to a low voltage (-3V in the present case)). Such P/R cycles form complete anti-clockwise loops in I-V characteristics in the range between V_g^{\min} and V_g^{\max} .

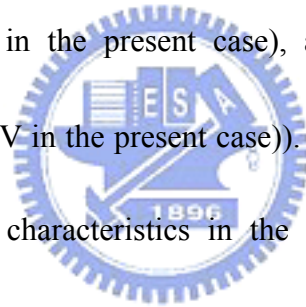


Figure 3.17 (a) depicts the transfer characteristics of a programmed device. It is seen that the transfer curve is different from the fresh state, owing to the sweeping voltage range starting at -1V. Figure.3-17 (a) also shows the time dependence of transfer characteristics. We can see that the transfer curves are gradually approaching the fresh stat. This can be explained that the charges loss with time. The time dependences of drain current (I_d)@ $V_g=0$ shown in Fig.3-17(a) is plotted in Fig.3-17 (b). Distinct difference between the programmed state and fresh state is still observed after 2400 sec.

Chapter 4

Effects of Channel Materials and Device Structures on Hysteresis

In Chapter 3, we put our focus on the hysteresis phenomenon of planar poly-Si TFTs. In order to further examine our model for the hysteresis and understand the potential of such mechanism for practical applications, we now consider the effects of channel materials and device structures on hysteresis in this chapter.



4-1 Characteristics of Poly-SiGe TFTs

Several previous reports have addressed the effect of Ge incorporation on device performance [45, 67, 68]. In this section, we will not discuss the details of basic electrical characteristics of poly-Si_{1-x}Ge_x devices. Instead, we will compare the width of hysteresis window between the poly-Si and poly-Si_{1-x}Ge_x TFTs. As shown in Figure 4-1, in which the width of hysteresis (H) of the two splits of device is shown as a function of channel length (L). Both splits have the same channel thickness (50 nm) and gate oxide thickness (20 nm). Channel width of the devices characterized in this

figure is $20\ \mu\text{m}$ and $5\ \mu\text{m}$ for poly-Si_{1-x}Ge_x and poly-Si devices, respectively. As can be seen in the figures, the variation in channel length has no influence on H. This is an indication of uniform trap density (N_T) along the channel. However, we discovered that the poly-Si_{1-x}Ge_x TFTs exhibit larger H over the poly-Si ones. This is attributed to the fact that the incorporation of Ge atoms would introduce a considerable amount of trap density [45, 67, 68]. Actually this would dramatically degrade the device performance, including worse subthreshold swing, higher threshold voltage, larger leakage current, and lower field-effect mobility, as observed in Fig.4-2. As a result, the numbers of electron trapped in the deep-levels of a poly-SiGe channel in the FS operation would be far more than that of poly-Si devices, resulting in a larger hysteresis window.

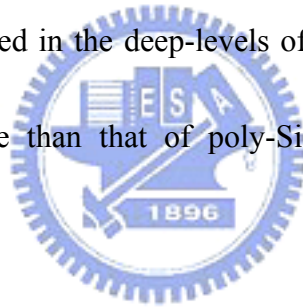


Figure 4-3 depicts the H of the devices as function of channel width. Channel length of the devices characterized in this figure is $2\ \mu\text{m}$ and $0.4\ \mu\text{m}$ for poly-Si_{1-x}Ge_x and poly-Si channel layers, respectively. The effect of channel width on the hysteresis window depicts a similar behavior to the channel length effect.

4-2 Effects of Double-gated Configuration

The transfer characteristics of the double-gated devices with split conditions listed in Table 2-1 are shown in Fig. 4-4. In the measurements the bottom-gate is grounded.

It can be seen that when the bottom oxide thickness decreases, the drain-induced barrier lowering (DIBL) becomes less severe. This is ascribed to the use of a thinner bottom oxide thickness that more effectively suppresses the field penetration from the drain to the channel, resulting in a better control over the short-channel-effect, as has been demonstrated in previous papers [69-71]. Figure 4-5 describes the effects of bottom-gate bias, V_{Bg} , on the transfer characteristics of the DG devices. It is obviously seen that, when T_{BOX} is thinner, V_{Bg} has a stronger influence on the transfer characteristics. This can be understood by the band diagrams of the DG configurations shown in Fig. 4-6. When the bottom oxide is thinner (Fig. 4-16 (a)), the variation in V_{Bg} would change the field strength in the bottom oxide more significantly and draw a more significant effect on the channel potential as well as the performance parameters such as V_{th} and SS. On the other hand, if the bottom oxide thickness is thicker (Fig.4-6 (b)), the variation of bottom oxide field by the V_{Bg} is smaller, and results in negligible change in the device characteristics.

Figure 4-7 displays the hysteresis characteristics of a DG device by executing the FS and RS operation to the top gate while a specific voltage ranging from -3 to 3 V is applied to the bottom-gate. The bottom oxide is 20 nm for the device. It is seen that the hysteresis window shrinks as the V_{Bg} is positive. From the transfer curves the leakage is high, indicating a leakage path from source to drain via the bottom channel

interface due to the positive V_{Bg} . In the RS mode, the bottom channel also serves as a path for the de-trapping of the electrons trapped in the channel. As a consequence, the window becomes smaller.

4-3 Basic Characteristics of Inversed-T-gated NW TFTs

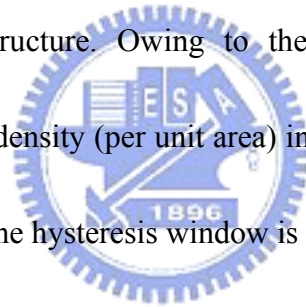
The fabrication of the NW devices is described in Chapter 2. The cross-sectional transmission electron microscopic (TEM) image of a device with inverse-T gate is shown in Fig.4-8. In the figure, it can be clearly seen that the location of the NW channels is precisely on the upper-step corners of the inverse-T gate. The perimeter of the triangular NW channel can be characterized in the enlarged view of the NW shown in Fig. 4-8 (b). Lengths of the three edges are 20nm, 30nm and 40nm, respectively. Fig.4-9 compares the transfer characteristics of NWTFTs with those of a conventional planar self-aligned device. Note that the drain current has been normalized to the channel width. The results indicate that the shrinkage of channel dimensions could help improve device performance in terms of higher on-current and steeper subthreshold swing and better control of short-channel effect, which can be ascribed to the increase in surface-to-volume ratio. The inverse-T double-gated (ITDG) -NWTFT can be operated in several operation modes, including single-gated (SG) modes and double-gated (DG) modes. SG modes can be further divided into two types:

MSG and SSG. MSG is to apply the sweeping gate bias to the inverse-T gate electrode while the top-gate is grounded, and vice versa for the SSG mode. In DG mode, both the inverse-T gate and top-gate electrodes are connected together and applied with the sweeping gate bias. The transfer characteristics of ITDG-NWTFT operated in these modes are shown in Fig. 4-10. In the figure, it can be clearly seen that the DG mode shows better electrical characteristics, including steeper subthreshold swing as low as 165 mV/dec. (@ $V_d=0.1V$) and higher on-state current, as compared with SG modes. This could be ascribed to its better gate controllability and larger conducting width. The subthreshold swing for MSG and SSG modes are 253 mV/dec and 313 mV/dec, respectively. This indicates that the inverse-T gate has a better gate controllability. Plasma damage induced during the formation process on the outer surface of the NW next to the top-gate is presumably another factor responsible for the worse subthreshold swing of the SSG mode.

To investigate the hysteresis characteristics of the ITDG-NWTFTs, we apply the gate voltage respectively to the different gate electrodes, i.e., top gate and inverse-T gate. Figure 4-11 shows the results of an ITDG-NWTFT under SSG and MSG modes of operation. It can be seen that the hysteresis phenomena are observed. Because the channel thickness of ITDG-NW is sufficiently thin, the NW channel is fully depleted, and it conforms to the condition that the hysteresis form according our model. The

width of hysteresis for MSG mode is smaller than that for SSG mode. This is ascribed to a higher density of defects on the outer surface of NW channels as we mentioned in the last paragraph.

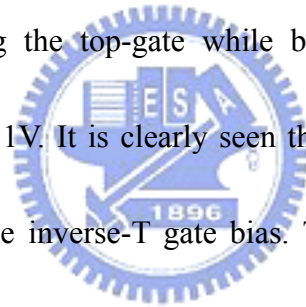
Figure 4-12 shows the results for the width of hysteresis (H) of the planar and ITDG-NW structure as a function of channel length (L). The sweeping voltage ranges from -3 to 6V in the measurements. As can be seen in the figure, the ITDG-NW structure depicts similar behavior trend as the planar structure, i.e., the channel length exhibits only very minor influence on H. In addition, the H is larger in planar structure than ITDG-NW structure. Owing to the thinner channel thickness in ITDG-NW, the effective trap density (per unit area) in the channel is smaller than that in planar structure, therefore the hysteresis window is also smaller.



We also vary the gate sweeping range to study its effect on the hysteresis window. Figure 4-13 shows the results of a device operated under MSG mode. In Figs. 4-13(a) and (b) the gate voltage starts at -3V but the highest voltage is varied from 3 to 9V. Basically the outcome is similar to that of planar structure presented in last chapter. It appears that, with increasing gate voltage, the transfer characteristics of RS mode shifts toward more negative direction and results in an increase of the hysteresis window, as observed in Fig.4-13 (a). The window becomes saturated as the maximum gate voltage reaches a specific voltage (about 8V in the case), as observed in Fig. 4-13

(b). In addition, in Fig. 4-13 (c), the maximum gate voltage is fixed at 6V, while the minimum gate voltage is set at -3 or 1 V, respectively. It is seen that the transfer characteristics of FS mode shifts as the starting voltage is changed, while the curves of the RS mode coincides.

Because of the small volume and thin body of silicon NW, the channel potential is sensitive to both gates, and thus strong gate-to-gate coupling can be observed and would have the capability of tuning the threshold voltage by adjusting the other gate voltage [72]. Figure 4-14 displays the hysteresis characteristics of ITDG-NWTFTs by sweeping the top-gate while biasing the inverse-T gate at a constant value, $V_{int}=-1, 0, \text{ or } 1\text{V}$. It is clearly seen that the threshold voltage can be successfully modulated by the inverse-T gate bias. This technique of adjusting the threshold voltage by controlling the bias of the second gate could be applied for low standby power circuits [72]. For the results shown in Fig. 4-14, not only the threshold voltage shifts with V_{int} , but also the hysteresis window does. With such feature, it is feasible to enlarge the hysteresis window by applying different V_{int} in FS and RS modes.



Chapter 5

Conclusions and Future Works

5-1 Conclusions

In this study, we report an interesting hysteresis phenomenon in poly-Si TFTs. To observe such phenomenon, the poly-Si channel must be thinner than a characteristic depletion width and receive no passivation treatment. We present a model to explain the finding, based on the electron trapping and de-trapping events occurring in the grain boundaries of the poly-Si channel. The mechanism is new and distinctly different from the origins of hysteresis phenomena reported in the literature. The proposed mechanism can be verified very well by our measurement results. The effects of several factors such as channel materials, channel length, channel width, temperature, and multiple-gate configuration are also investigated.

Since the hysteresis window exhibiting in the transfer current-voltage characteristics is significant, we also exploit its feasibility for memory device applications. We found that the retention property is better than that of 1-T DRAM, although a long programming time (> 1 ms) is needed.

5-2 Future Works

The preliminary investigation of hysteresis in poly-Si TFTs have been performed in this thesis. For the purpose of further enhancing the device performance, some valuable suggestions for future work are listed as follows:

- (1) Reducing the programming time of the proposed device is essential for practical application. Since the external resistance of source and drain in our devices is large, it may lead to additional RC delay in operation. To greatly reduce the resistance, we can employ salicide process to the source/drain.
- (2) Our results show the use of NW channel can improve subthreshold swing and on/off ratio, while polycrystalline $\text{Si}_x\text{Ge}_{1-x}$ films can enlarge the hysteresis window and improve the retention characteristics, we can combine the two features to fabricate the poly-Si/ poly- $\text{Si}_x\text{Ge}_{1-x}$ / poly-Si NW-TFTs. In the NW channel structure, a thin poly- $\text{Si}_x\text{Ge}_{1-x}$ serving as the major charge storage medium is sandwiched between two poly-Si layers. We expect the proposed structure to effectively enhance the performance of the memory device.

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Table 1-1

Features of EMBEDDED MEMORIES

	DRAM	SRAM	1T-DRAM
Cell Size	8~12F ²	140~150F ²	4F ²
Cell Formation	1T1C	6T	1T
Store Mesh	Capacitor	Flip flop	Floating body
New material	High-k	None	None
Speed	Fast	Ultra fast	Fast

* F: minimum feature size

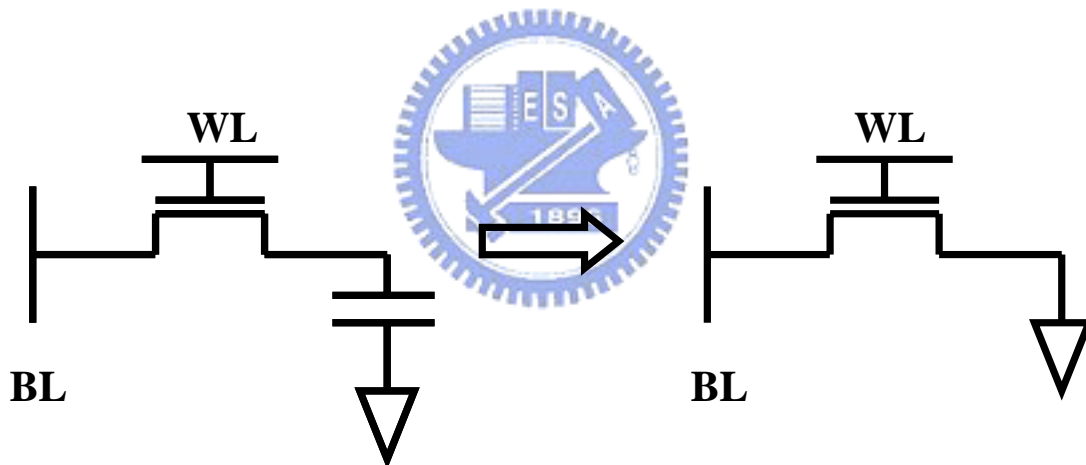


Fig. 1-1 Comparison between conventional one transistor/one capacitor (1T/1C) DRAM and capacitor-less one transistor (1T) DRAM.

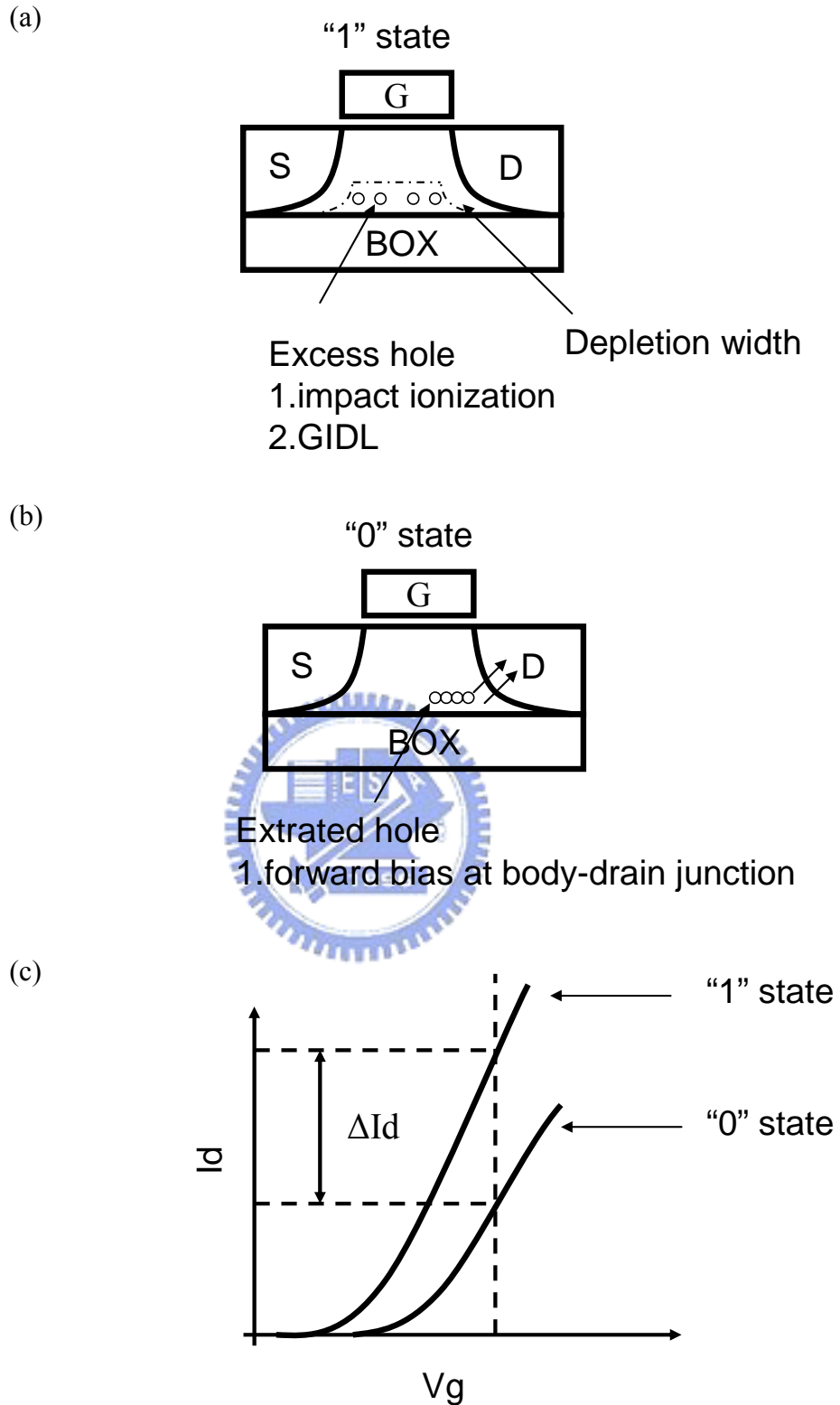


Fig. 1-2 Operation principle of 1T-DRAM cell. (a) Excess holes accumulate in the body region at "1" state (low V_{th}). (b) Excess holes are swept out of the body region at "0" state (high V_{th}). (c) I_d - V_g characteristics to represent (a) and (b) (not experiment data)

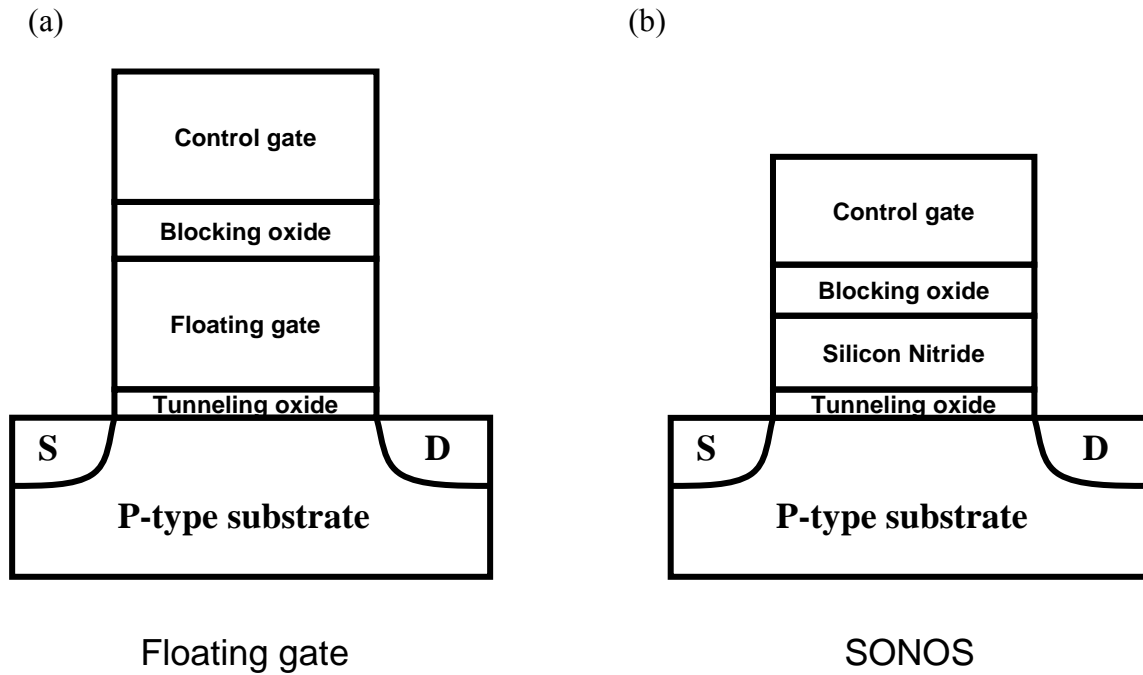


Fig. 1-3 Comparison of memory cell cross-section between the floating gate structure and SONOS.

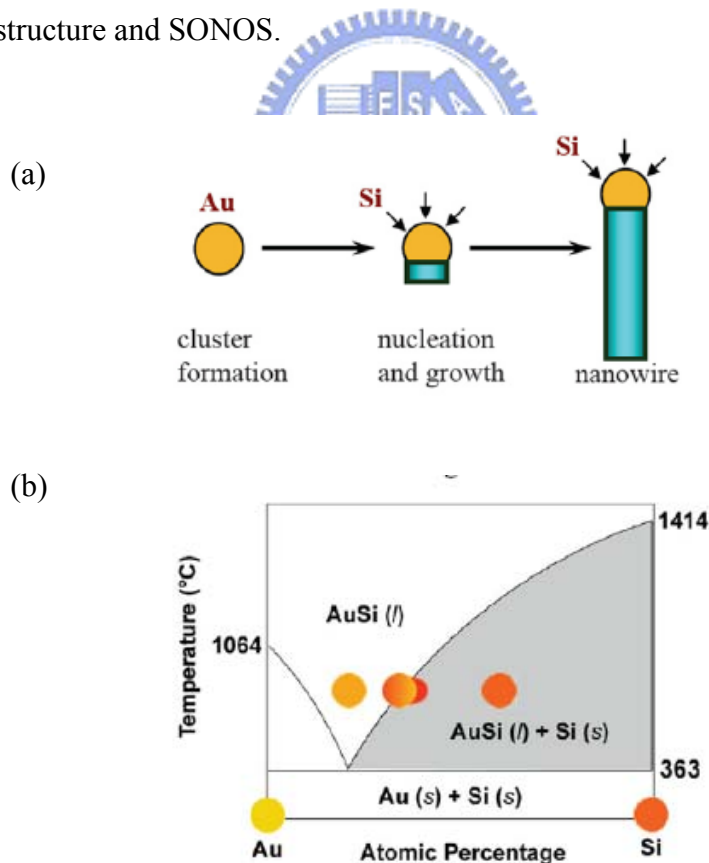


Fig. 1-4 (a) Schematic representation of VLS growth of Si nanowires. (b) Binary phase diagram for Au and Si illustrating the thermodynamics of VLS growth. [40].

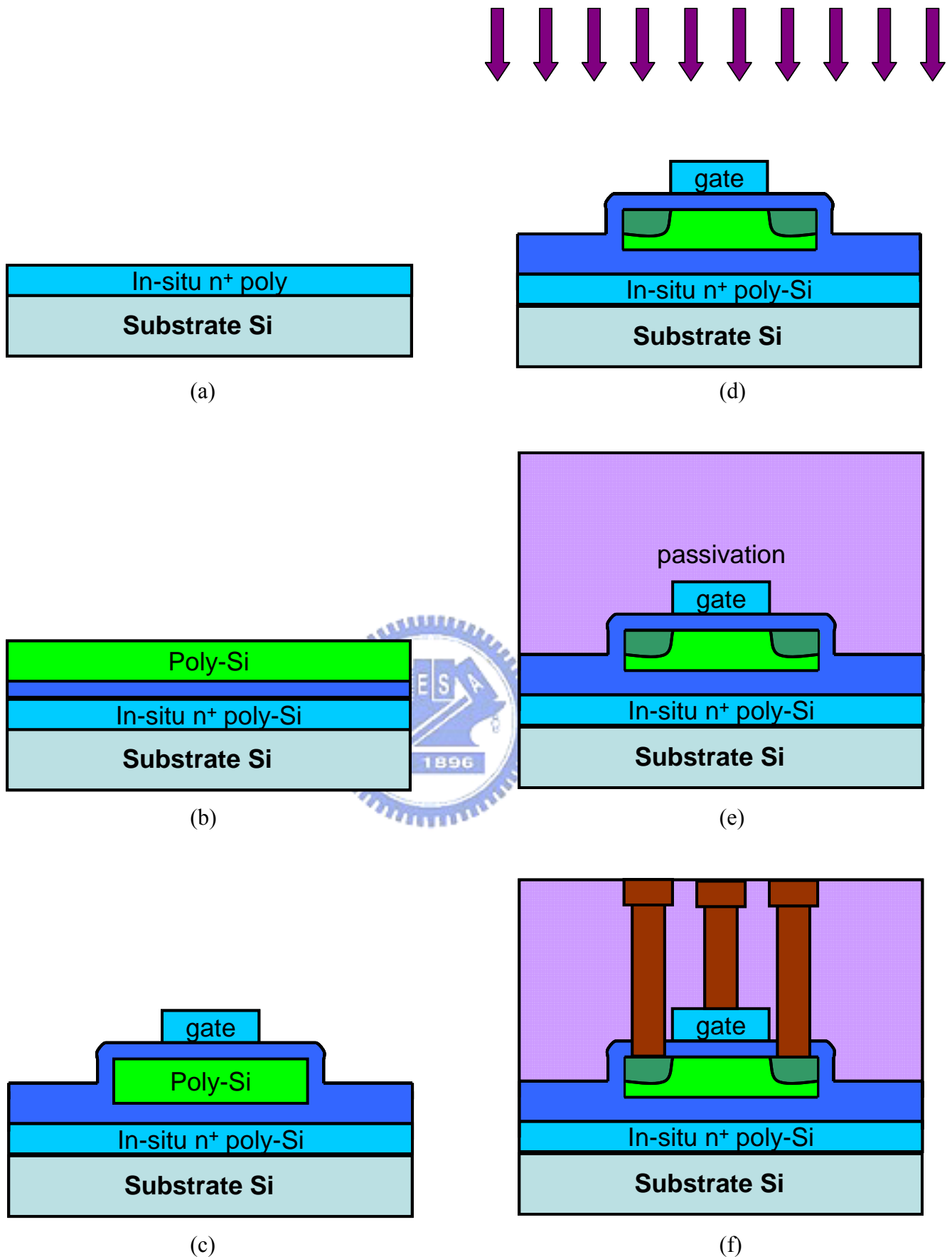
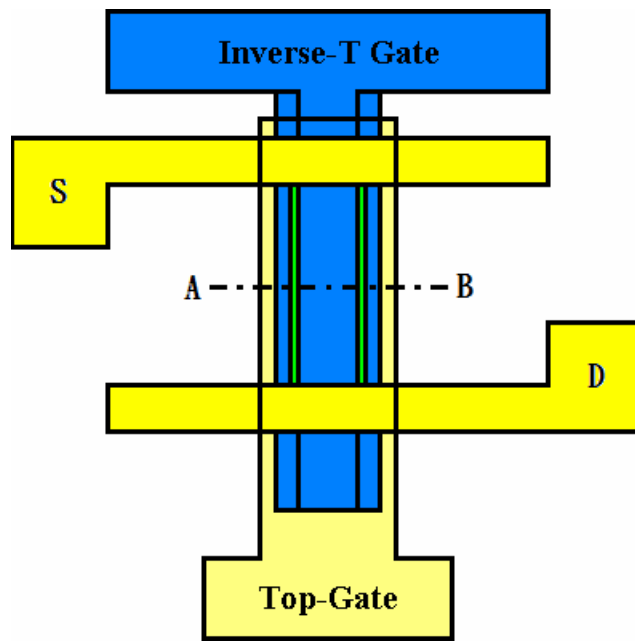
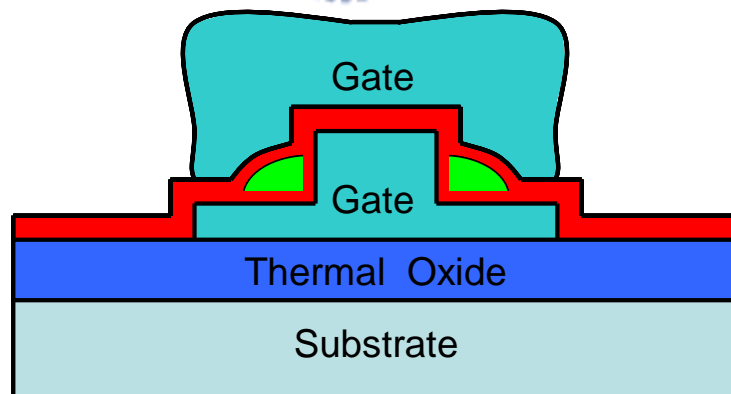


Fig. 2-1 Process flow of the conventional poly-Si TFT fabrication



(a)



(b)

Fig. 2-2 (a) Top view and (b) cross-sectional view of the ITDG-NWTFT

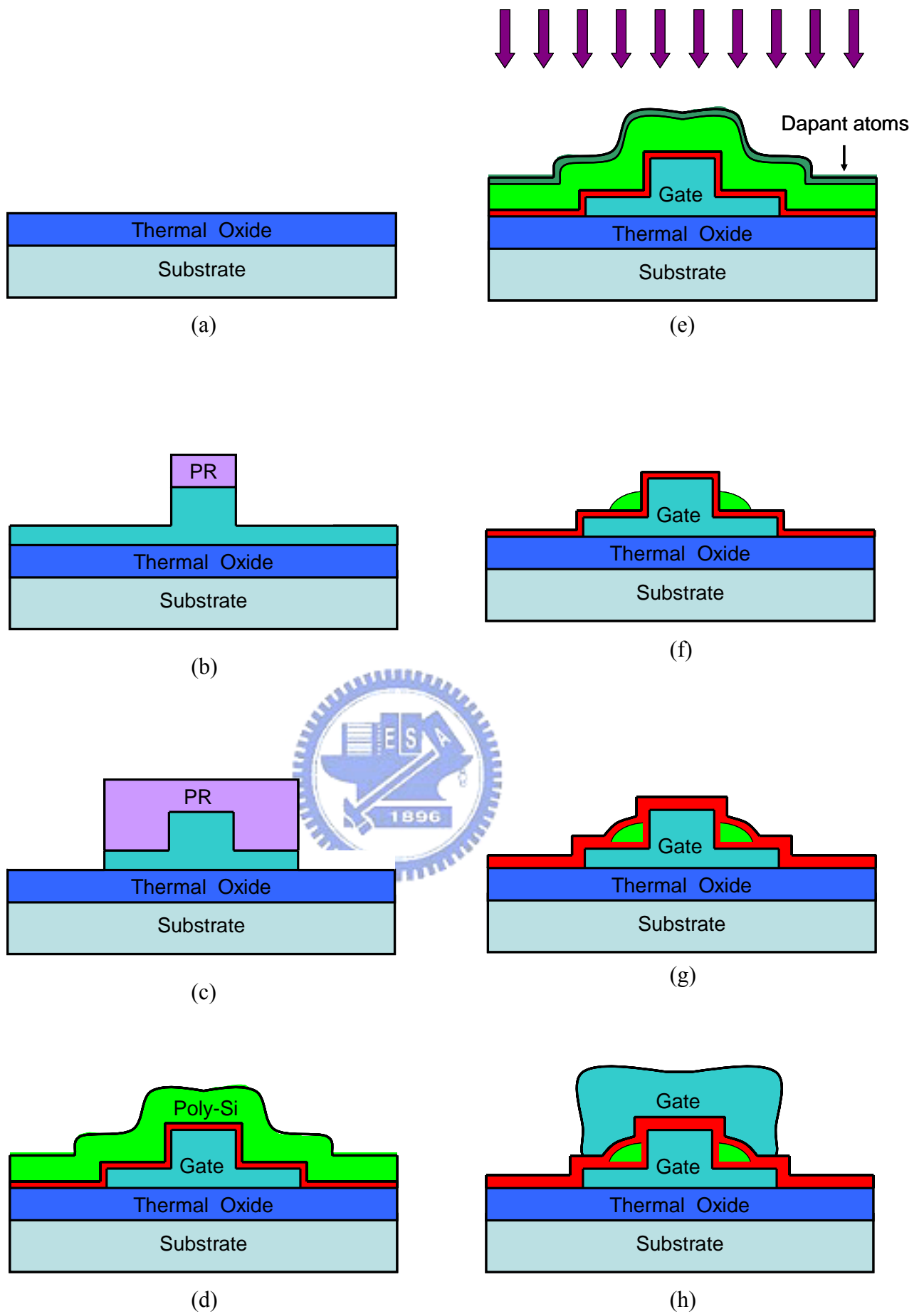


Fig. 2-3 Schematic diagram of Inverse-T double-gate NWTFT process

Table 2-1

Split conditions of bottom oxide thickness

	01	02	03
Bottom oxide thickness	20nm	50nm	200nm

*Channel thickness=50nm



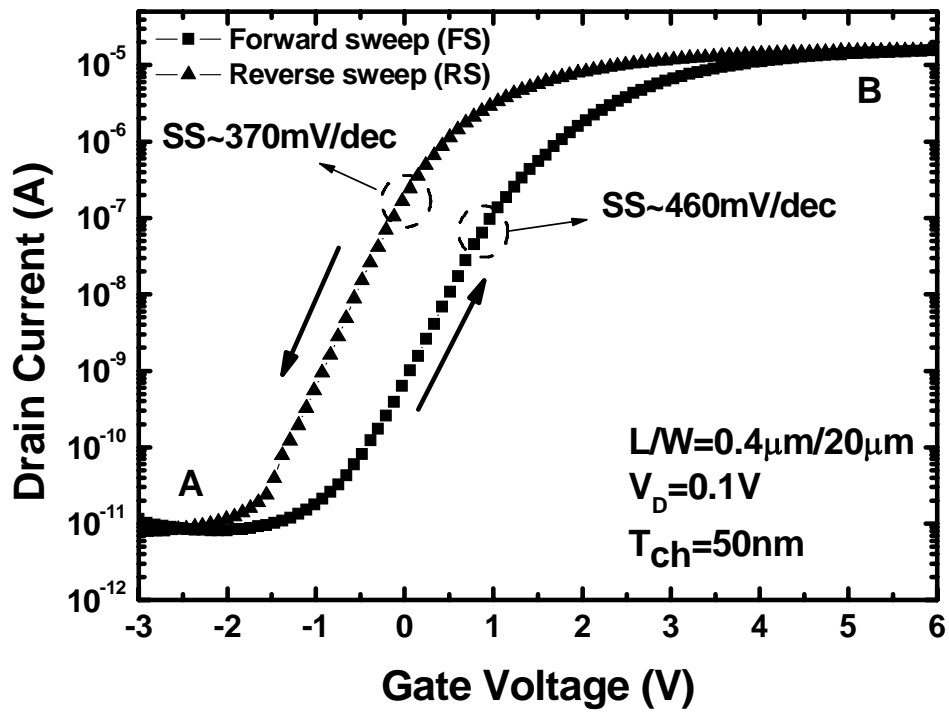


Fig.3-1 Transfer characteristics of TFTs with channel thickness 50nm under FS and RS modes.

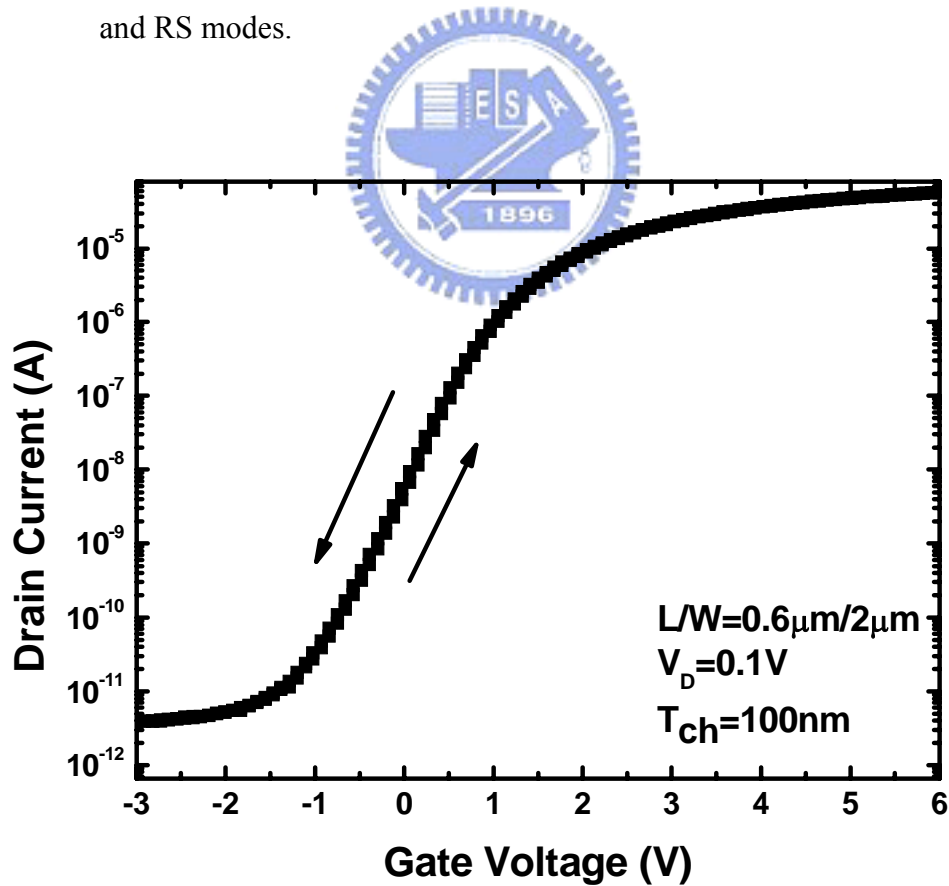


Fig.3-2 Transfer characteristics of TFTs with channel thickness 100nm under FS and RS modes.

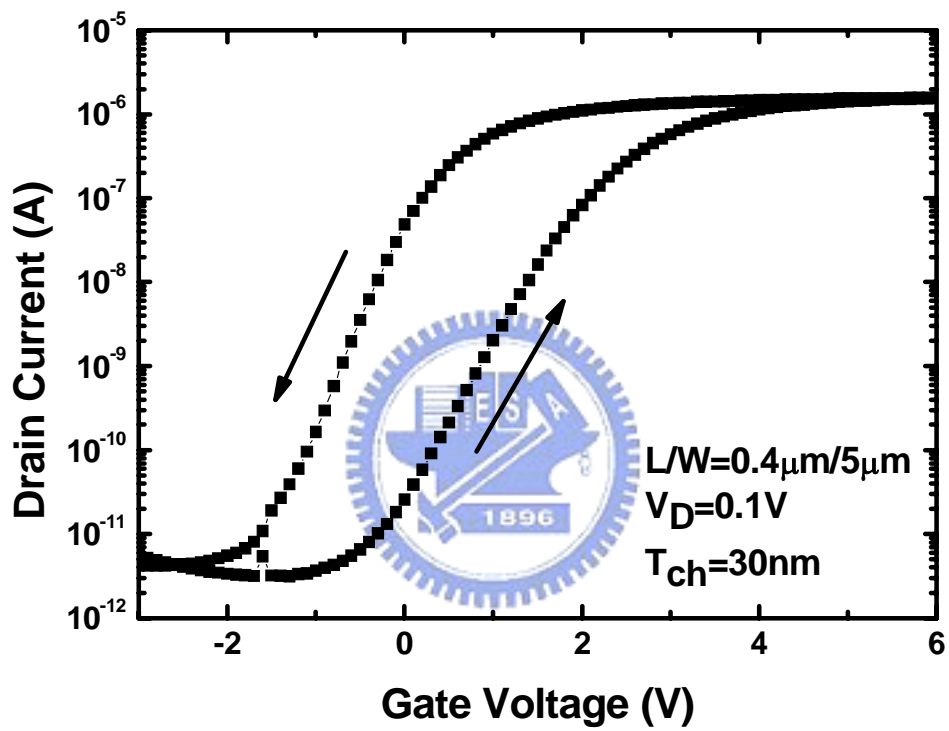
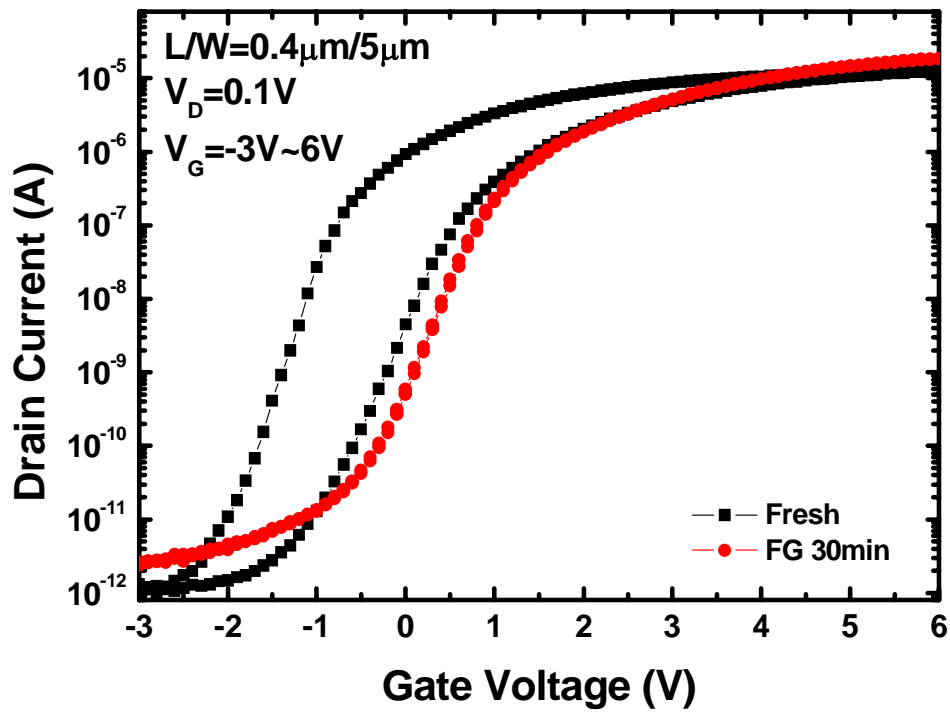
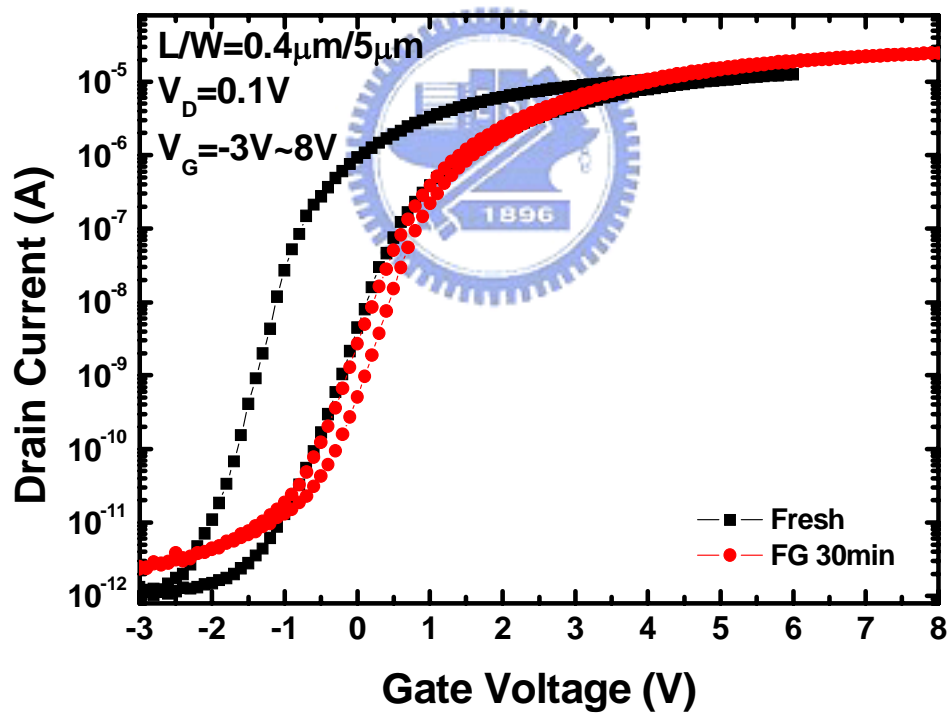


Fig.3-3 Transfer characteristics of TFTs with channel thickness 30nm under FS and RS modes.

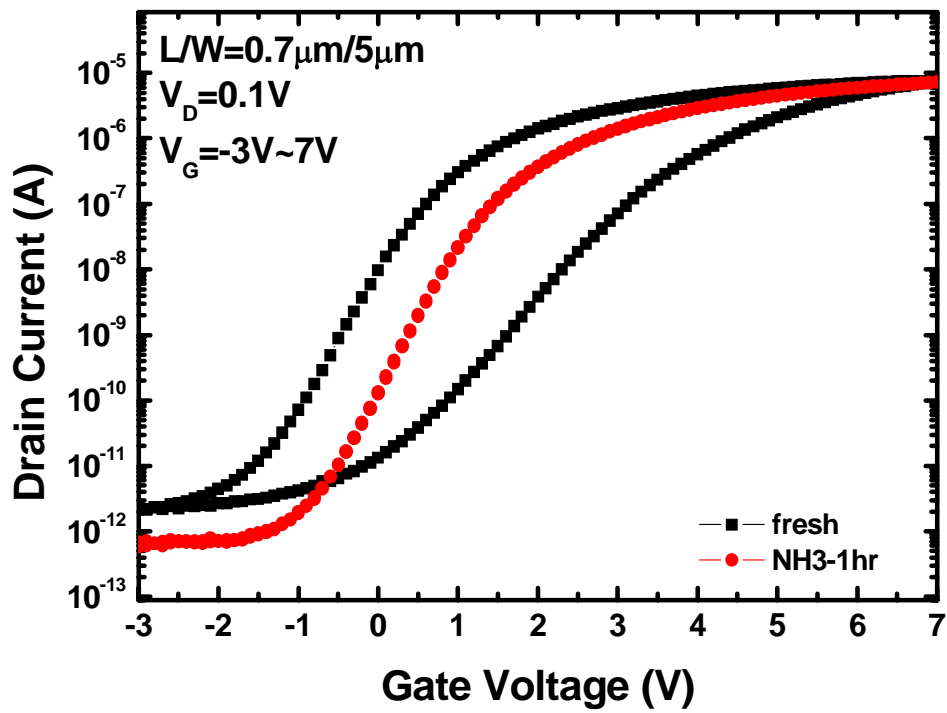


(a)

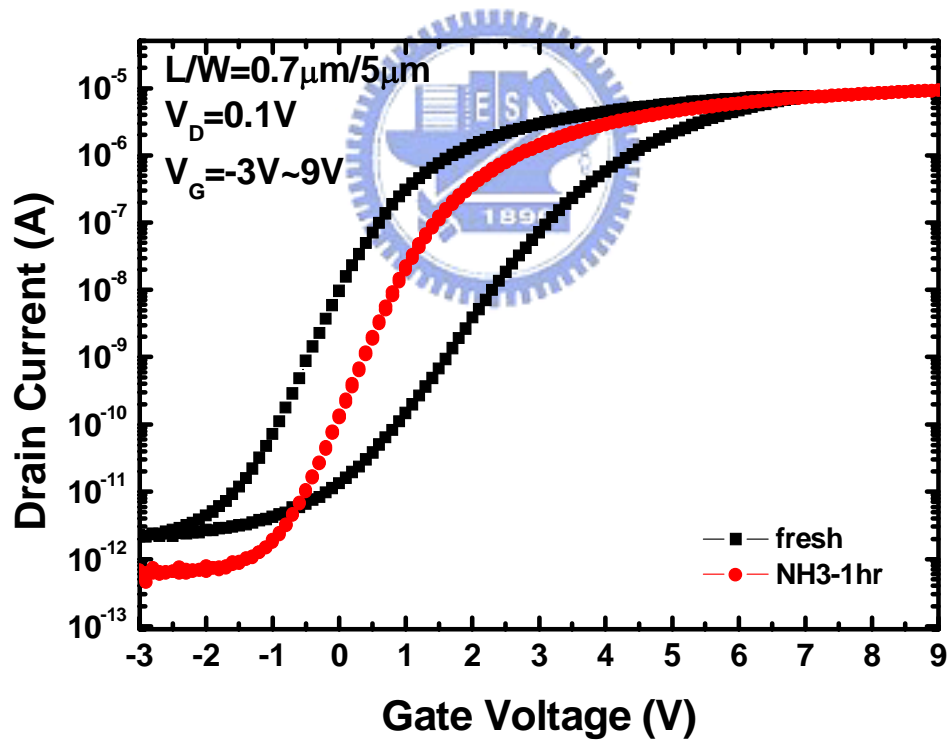


(b)

Fig.3-4 Transfer characteristics of TFTs which received FG annealing 30min under FS and RS modes with different gate sweeping range: (a) $V_G=-3\text{V}$ to 6V, and (b) $V_G=-3\text{V}$ to 8V. Characteristics of fresh device are also shown for comparison.

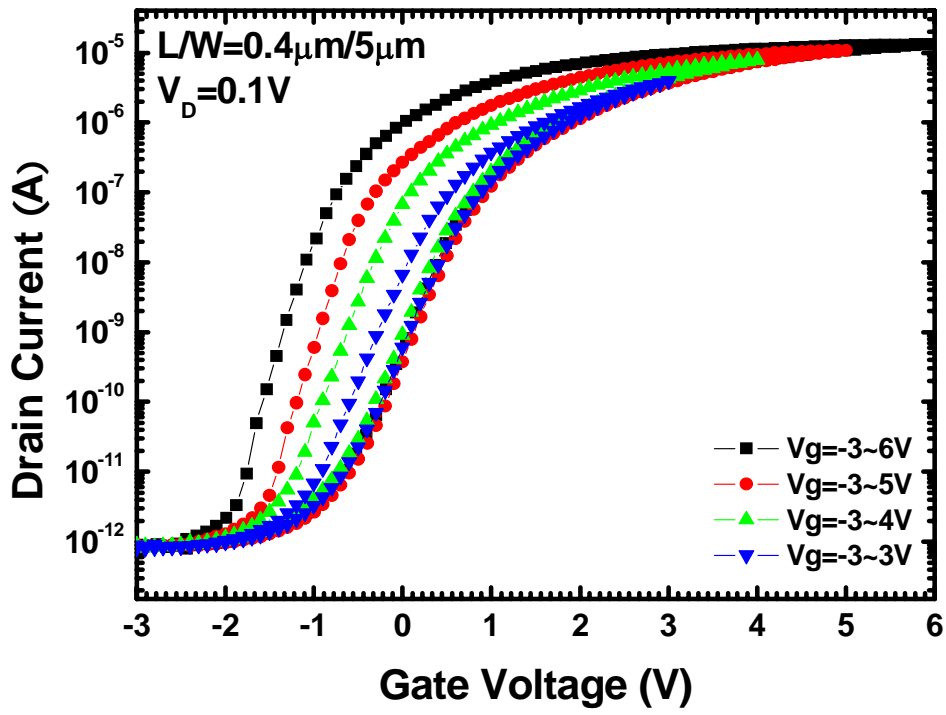


(a)

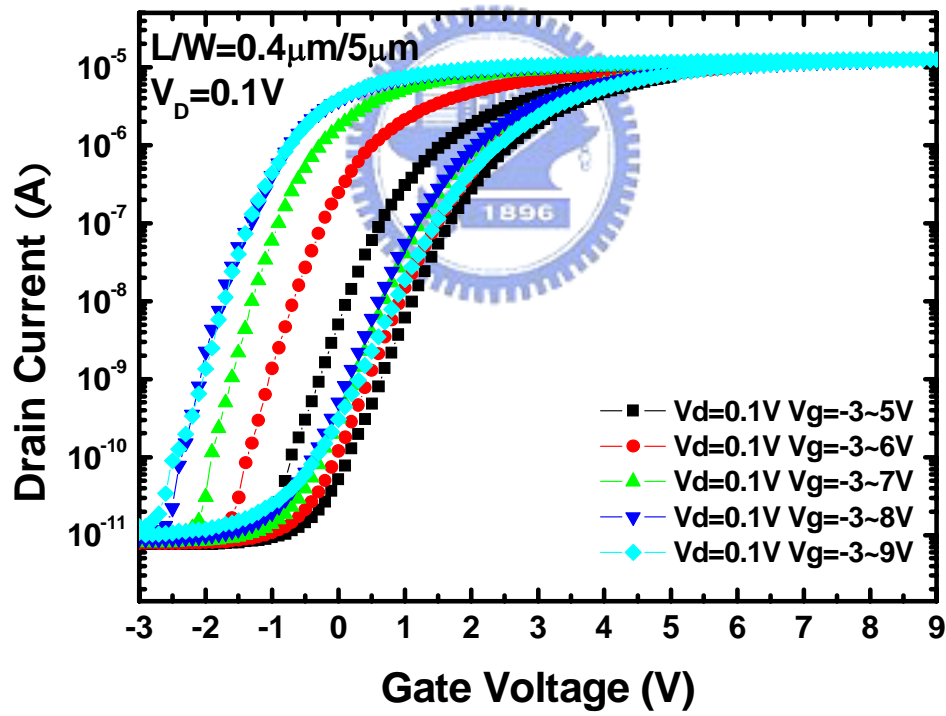


(b)

Fig.3-5 Transfer characteristics of TFTs which received NH_3 -plasma treatment under FS and RS modes with different gate sweeping range: (a) $V_G=-3\text{V}$ to 6V , and (b) $V_G=-3\text{V}$ to 8V . Characteristics of fresh device are also shown for comparison.



(a)



(b)

Fig.3-6 Transfer characteristics with fixed V_G^{\min} and different V_G^{\max} under FS and RS modes.

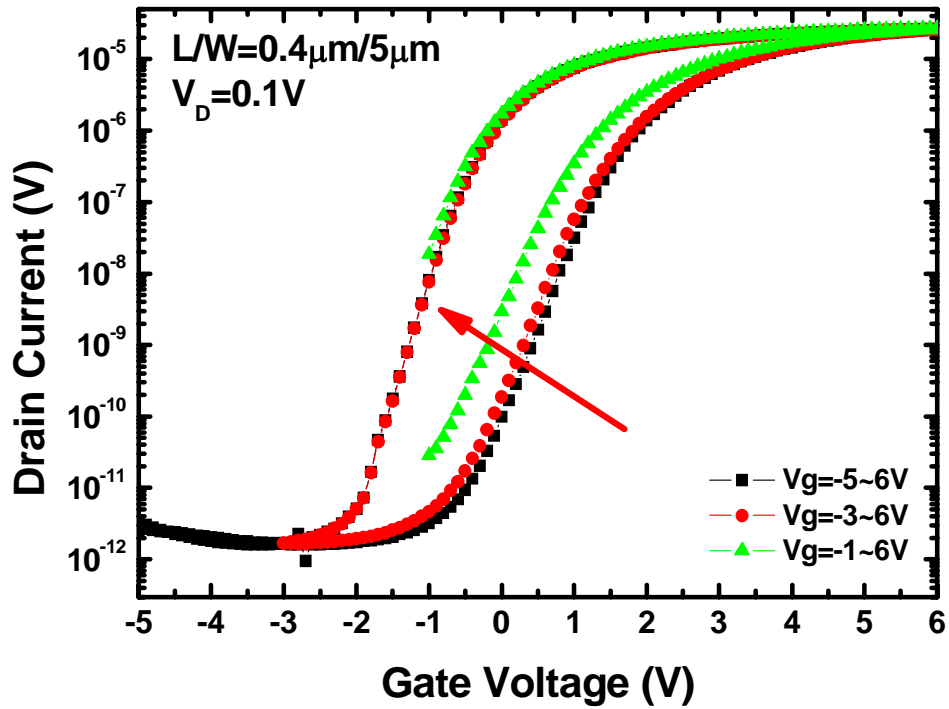


Fig.3-7 Transfer characteristics with fixed V_G^{\max} and different V_G^{\min} under FS and RS modes.

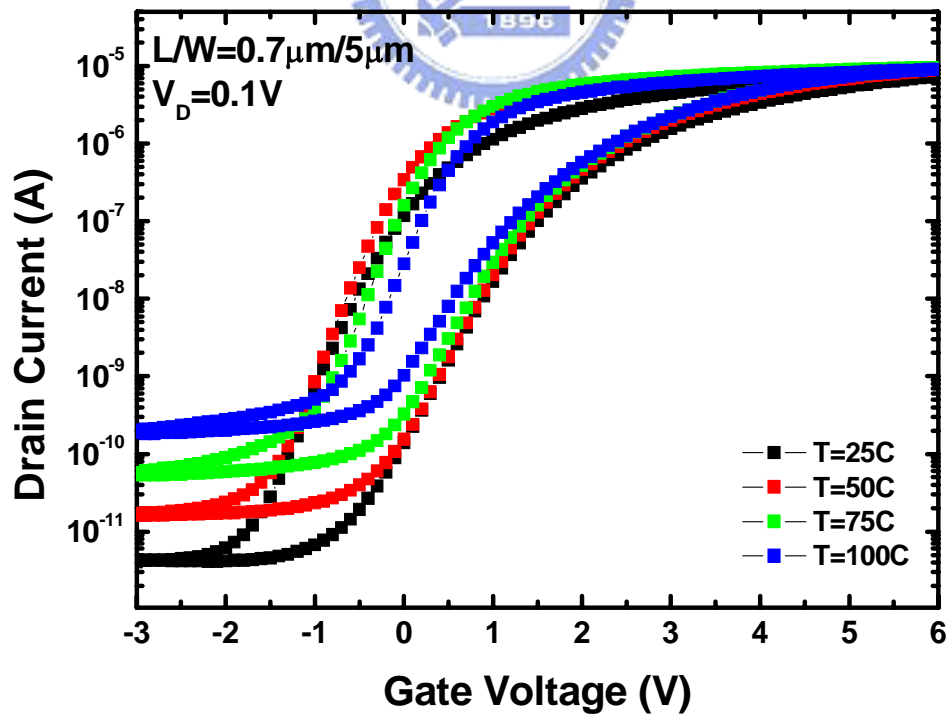


Fig.3-8 Transfer characteristics of TFTs measured at different temperature under FS and RS modes.

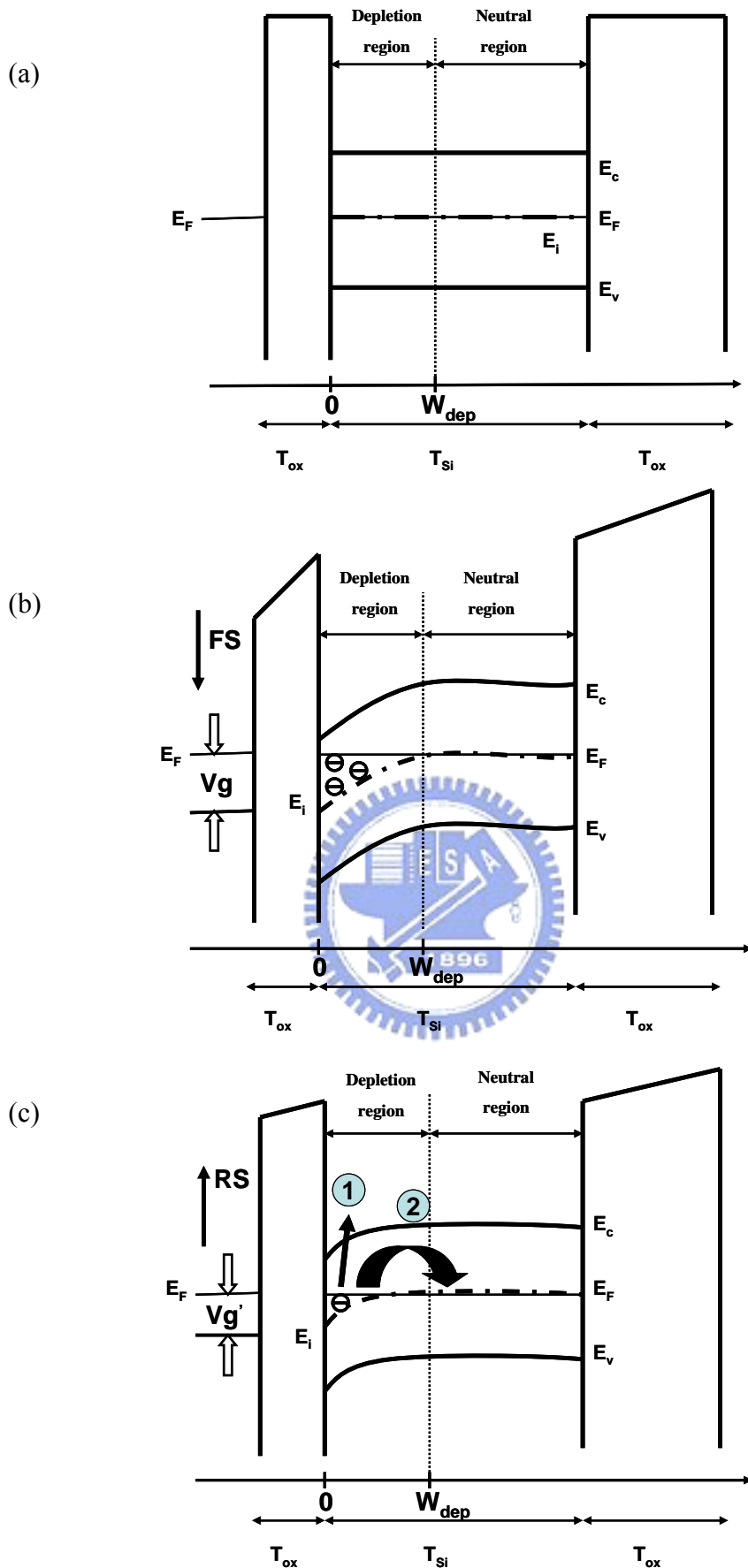


Fig.3-9 The energy band diagram of a partially depleted poly-Si channel under (a) flat-band (b) on-state (FS), (c) and on-state (RS). The two possible de-trapping paths for trapped electrons during RS period are also depicted and denoted as 1 and 2.

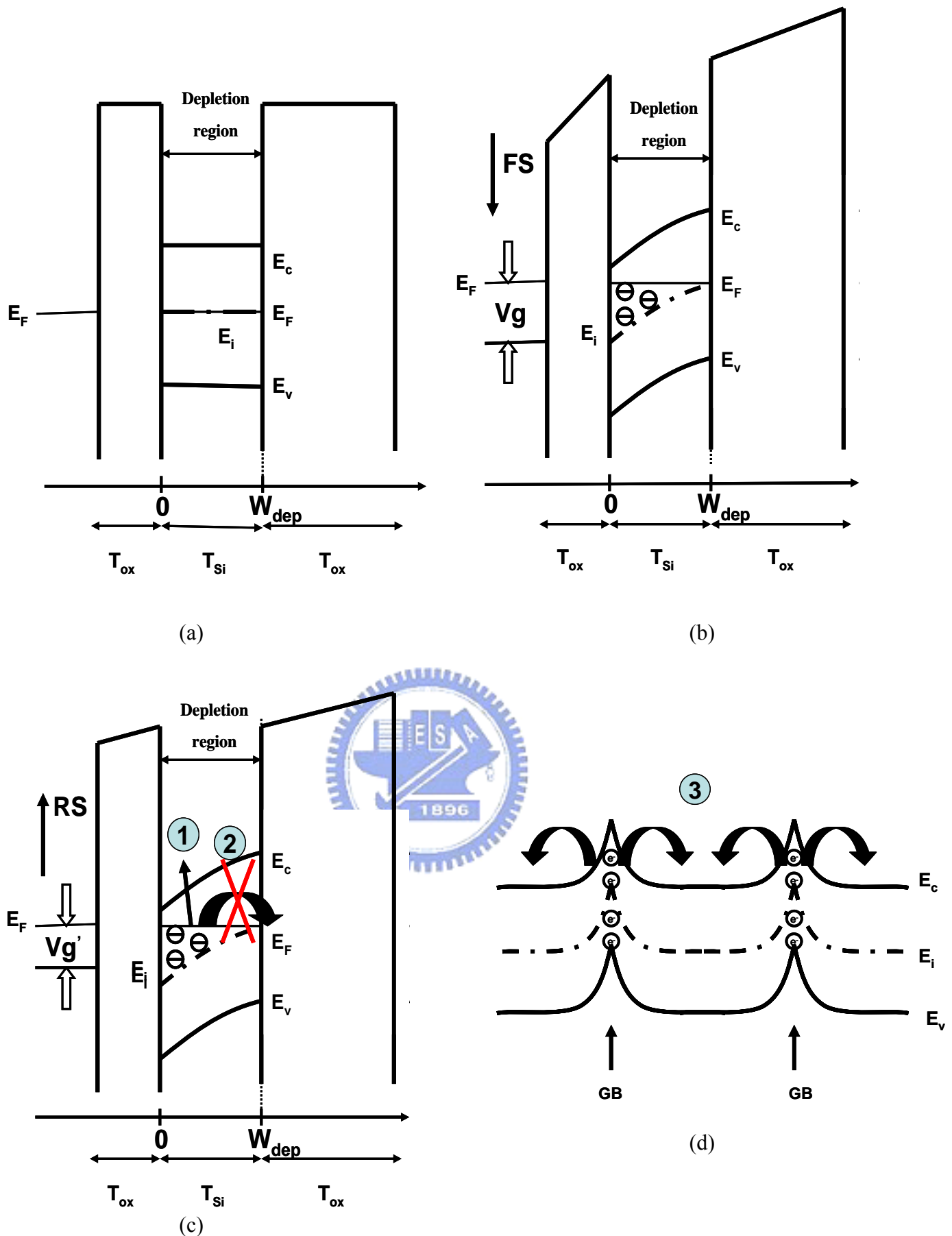


Fig.3-10 The energy band diagram of a fully depleted poly-Si channel under (a) flat-band (b) on-state (FS), (c) and on-state (RS). The energy band structure along the channel direction is also illustrated in (d). The three de-trapping paths for trapped electrons during RS period are also depicted and denoted as 1, 2 and 3, although the path 2 is prohibited by the buried oxide.

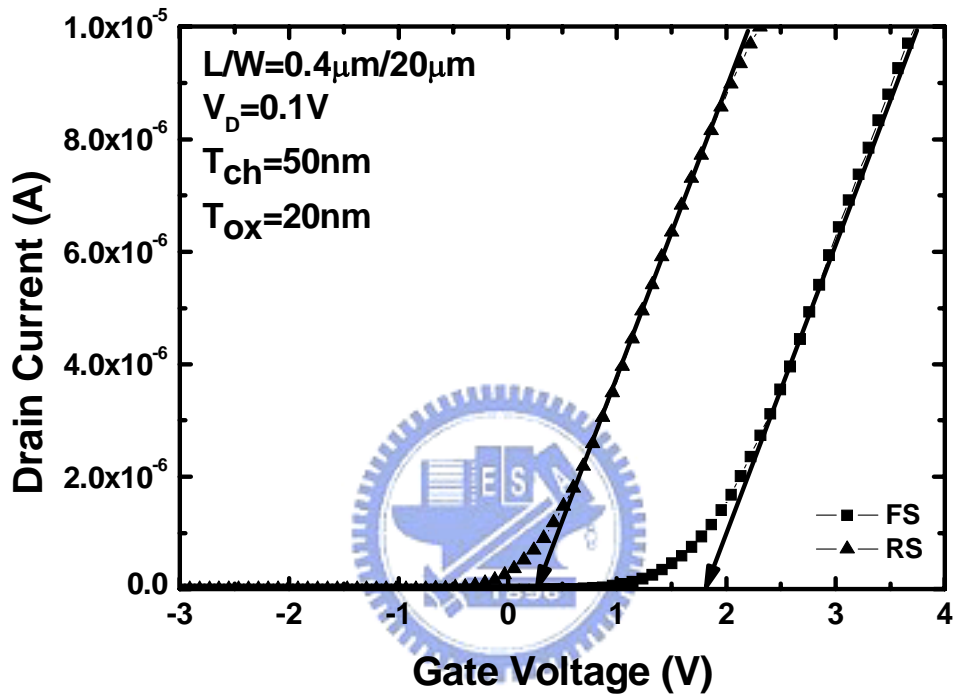
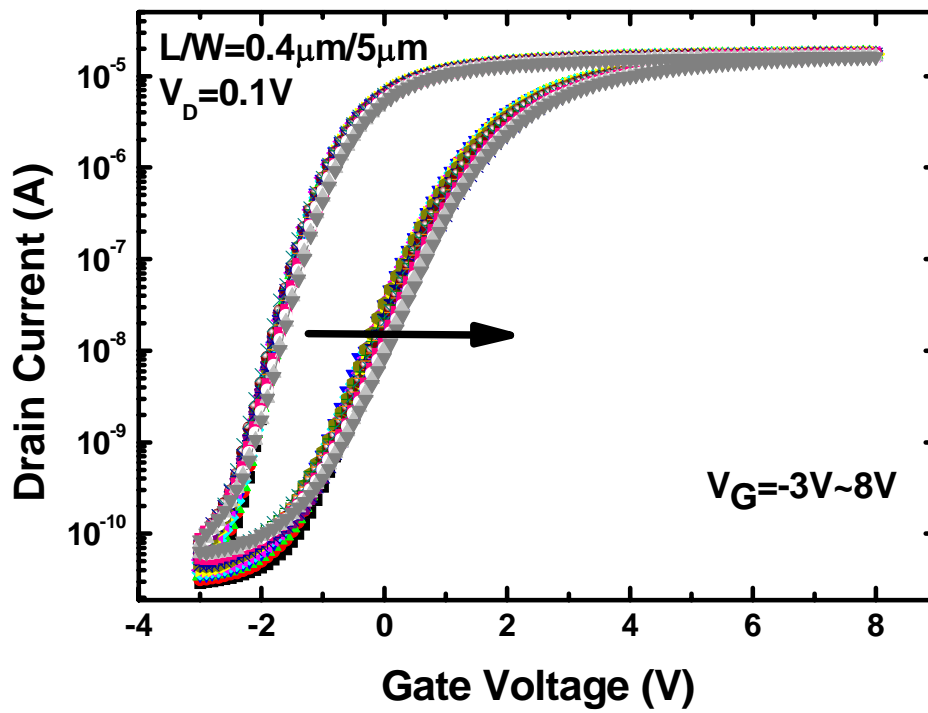
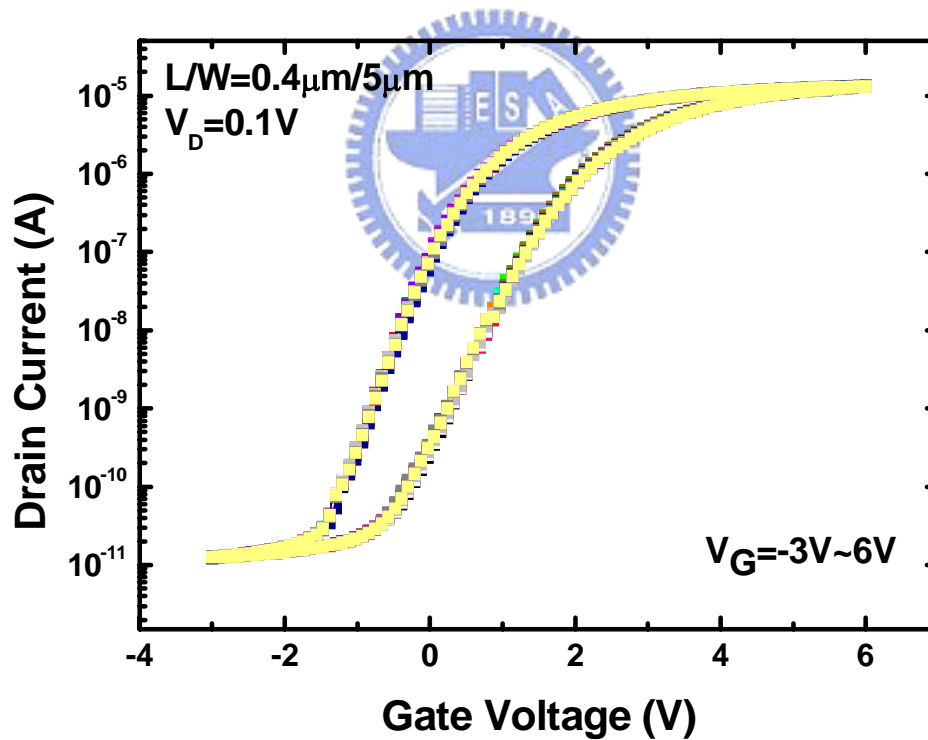


Fig.3-11 Transfer characteristics of TFTs in linear scale for determining the threshold voltage.



(a)



(b)

Fig.3-12 Transfer characteristics of 20 consecutive RS and FS measurements of a TFT with sweeping gate voltage range of (a) $V_G=-3\text{V}$ to 8V and (b) $V_G=-3\text{V}$ to 6V .

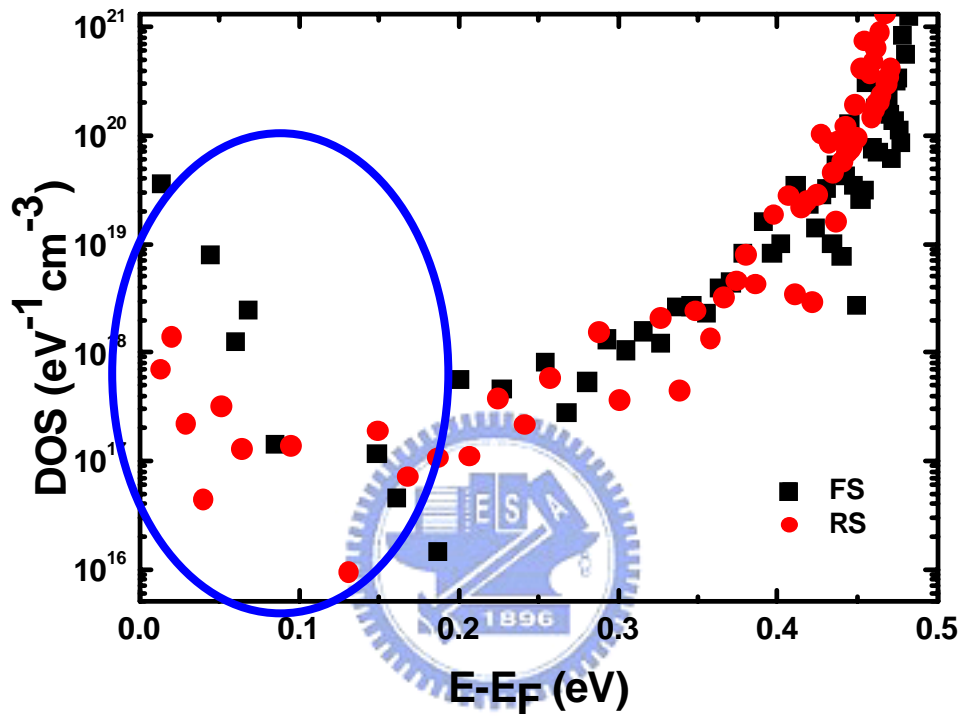


Fig.3-13 Calculated density-of-states as a function of surface band-bending.

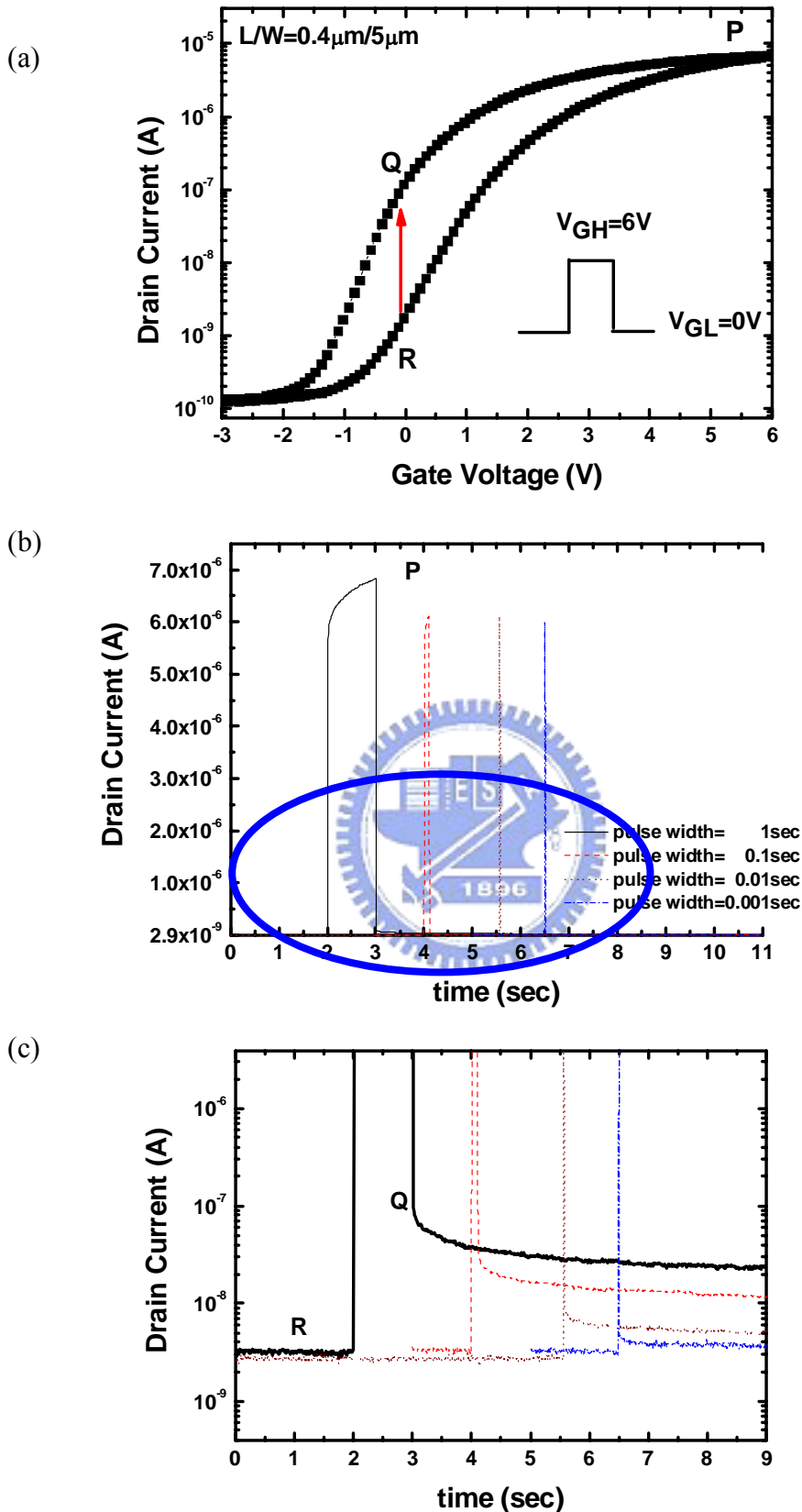


Fig.3-14 (a) Transfer characteristics of a TFT under FS/RS mode. A pulse with gate voltage of 6 V is applied to the gate. The inset is the pulse form. R and G points represent the conditions corresponding to the reading of the two logic states if we set the read voltage at $V_g = 0\text{V}$, and P point represents the condition programming. (b) Evolution of drain current of a device before, during, and after the programming with different pulse width. The circled area in the figure is further illustrated in (c) in which the drain current is illustrated in log-scale.

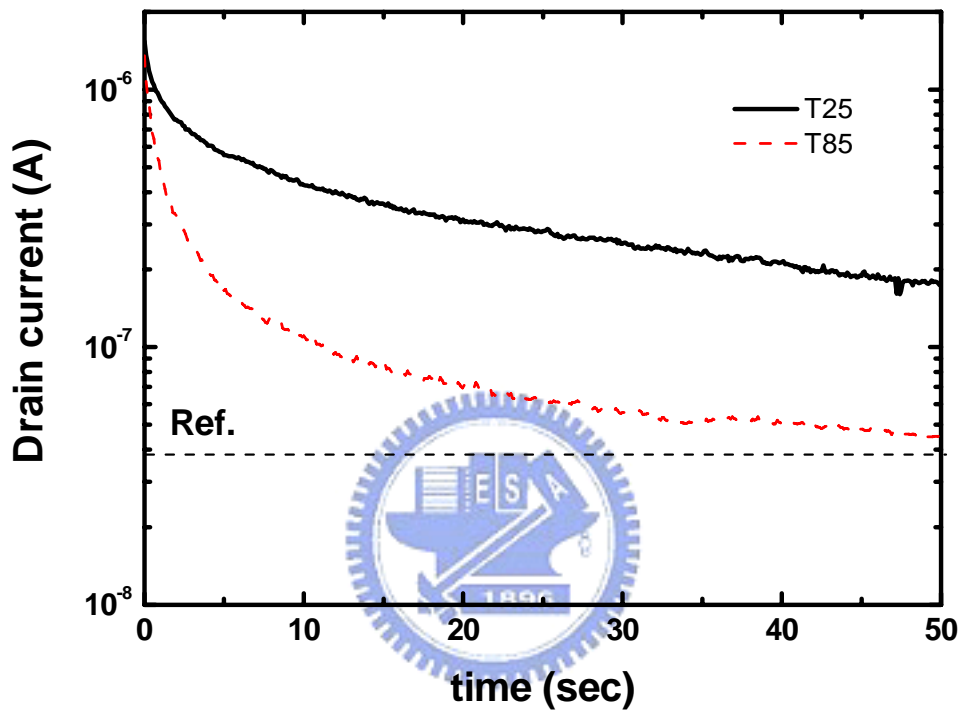


Fig.3-15 Retention characteristics of a programmed device characterized at room temperature and 85 °C.

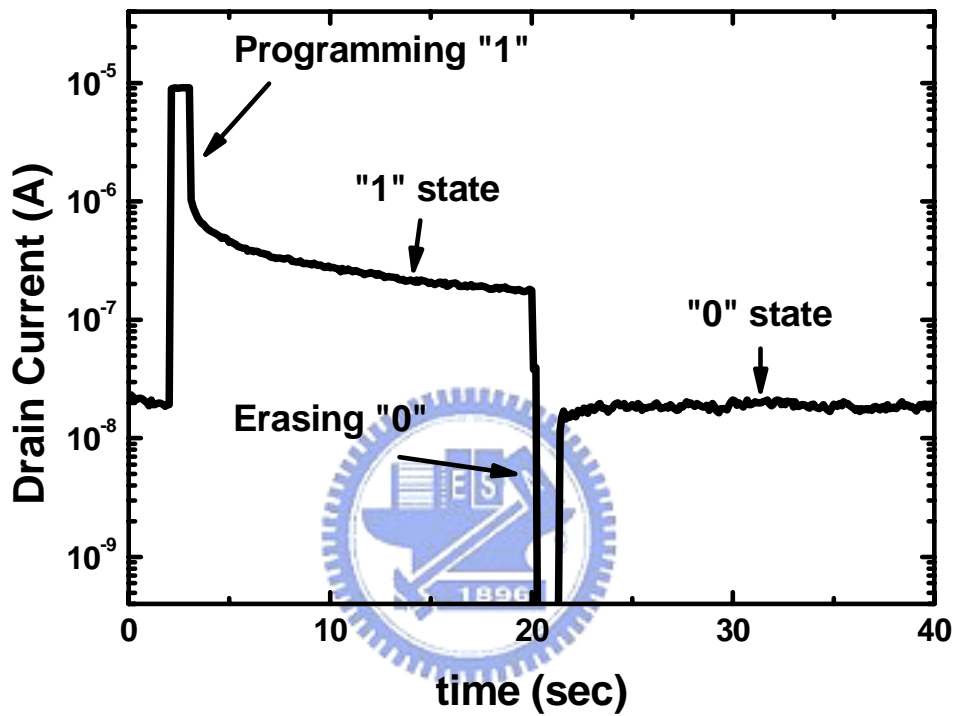


Fig.3-16 Evolution of drain current for a poly-Si TFTs with $L/W=0.4\mu\text{m}/20\mu\text{m}$ at $V_G=0\text{V}$, $V_D=0.1\text{V}$. "1" state is a programmed state with a gate pulse of 6V and pulse width 1 sec, "0" state is obtained by erasing the device with a gate pulse of -3V and pulse width 1 sec.

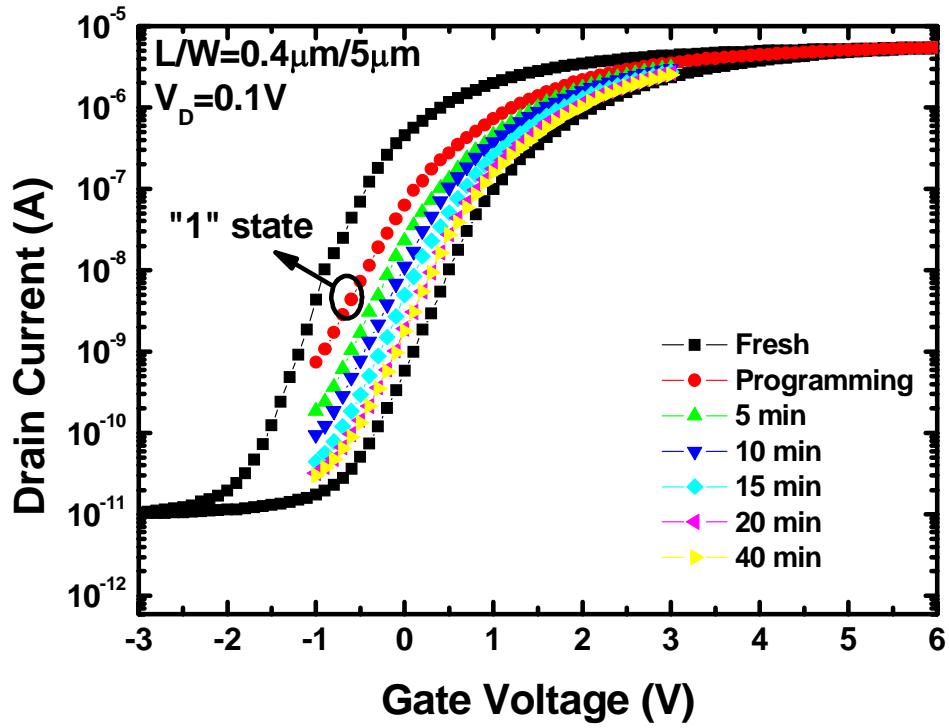


Fig.3-17 (a) Time dependence of transfer characteristics.

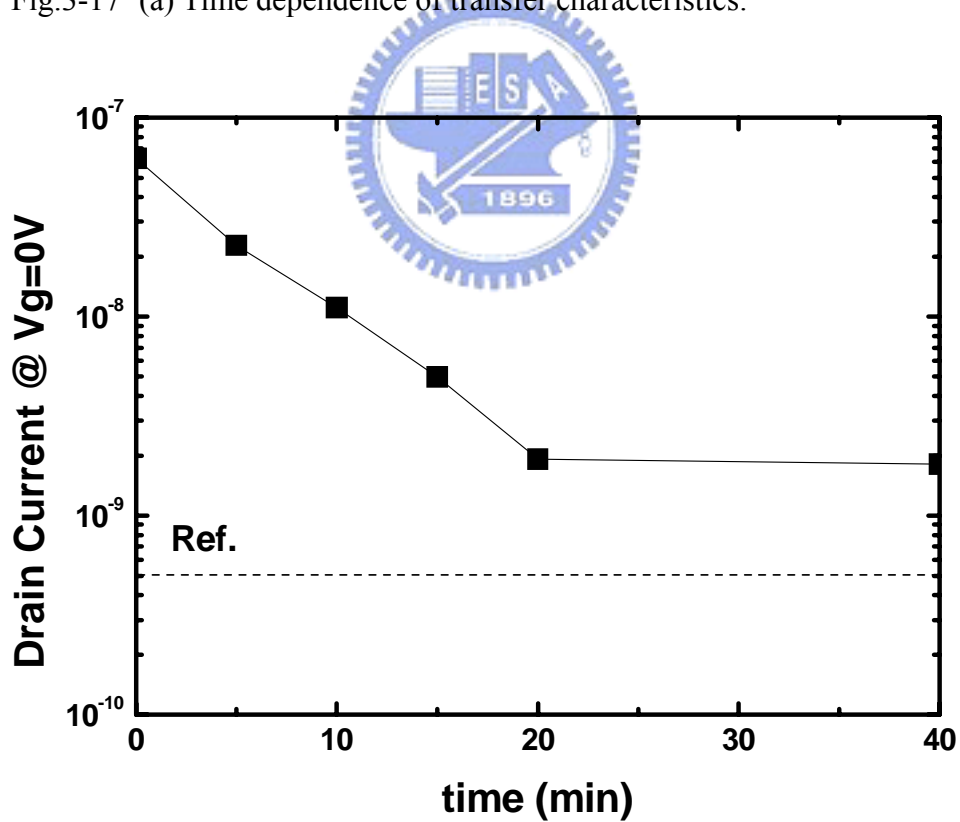


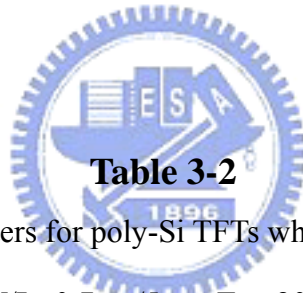
Fig.3-17 (b) Retention time of TFTs at $V_G=0V$.

Table 3-1

Major performance parameters for poly-Si TFTs which received FG annealing 30min.

W/L=0.4um/5um, T_{ox} =20nm

	V_{th} (V)	SS (mV/dec)	Mobility (cm²V⁻¹S⁻¹)	On/off current ratio
Fresh	1.61	367	16.69	9.69x10 ⁶
FG 30min	1.72	365	25.05	7.09x10 ⁶

**Table 3-2**

Major performance parameters for poly-Si TFTs which received NH₃-plasma 1hr.

W/L=0.7um/5um, T_{ox} =20nm

	V_{th} (V)	SS (mV/dec)	Mobility (cm²V⁻¹S⁻¹)	On/off current ratio
Fresh	4.14	0.693	13.79	3.69x10 ⁶
NH₃-plasma 1h	2.39	0.432	24.34	1.06x10 ⁷

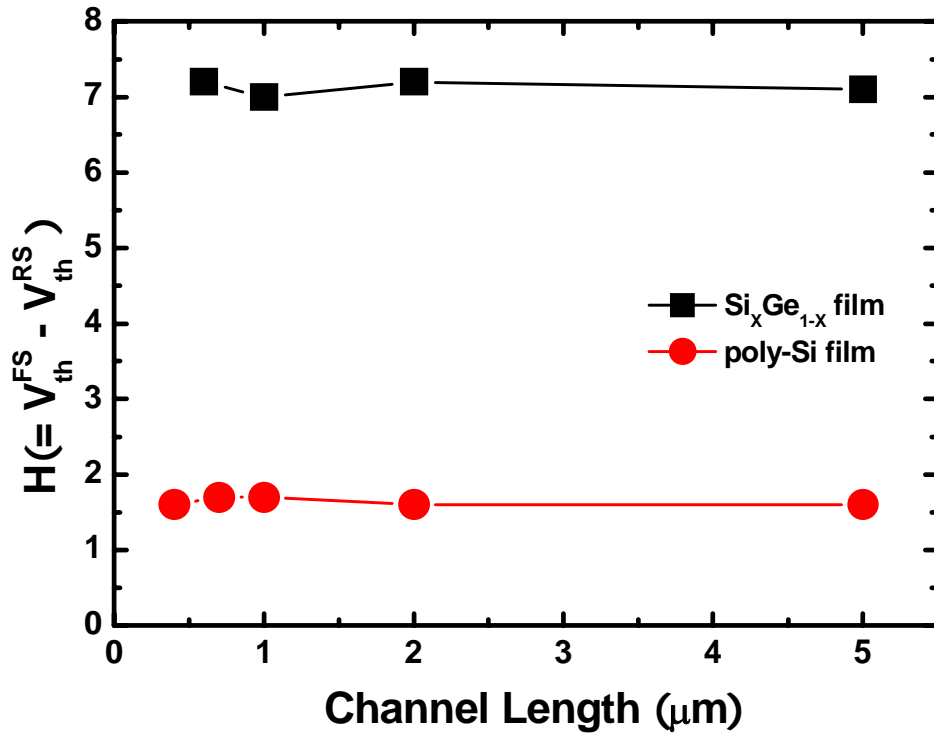


Fig. 4-1 The width of hysteresis (H) as a function of channel length for different channel poly-Si and poly-Si_xGe_{1-x} TFT devices.

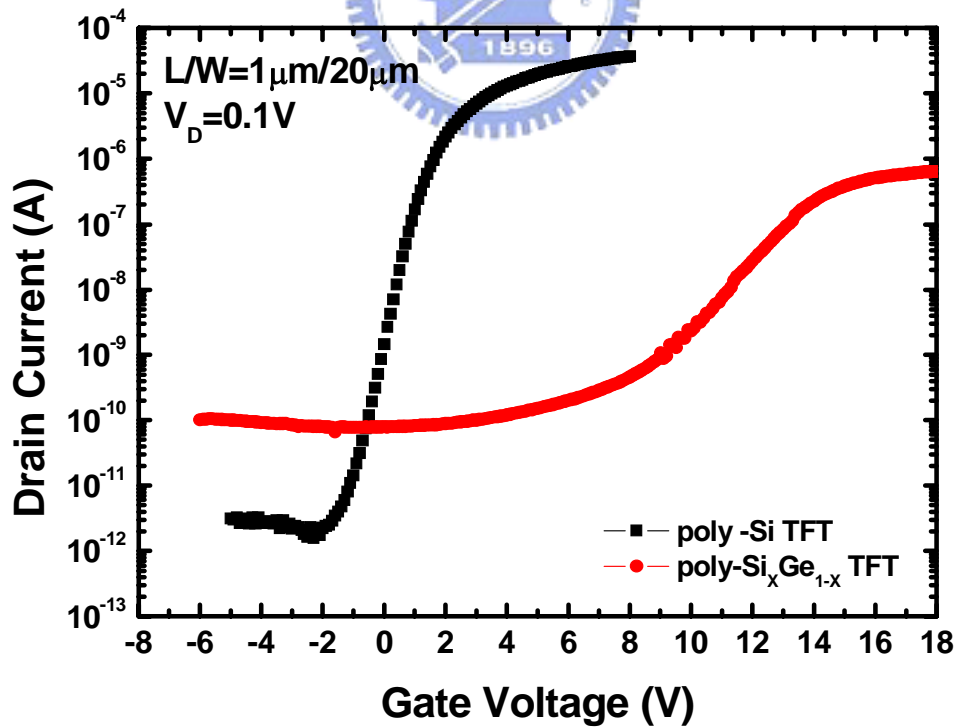


Fig. 4-2 Comparisons between transfer characteristics of poly-Si and poly-Si_xGe_{1-x} TFTs.

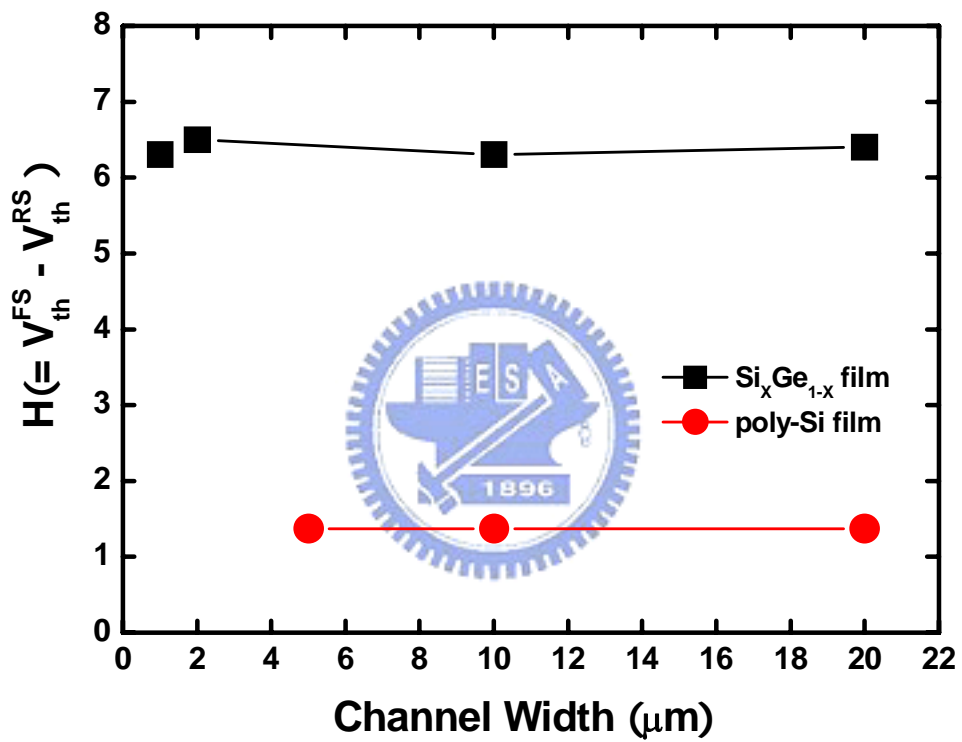


Fig. 4-3 The width of hysteresis (H) as a function of channel width for poly-Si film and poly-SixGe1-x TFTs.

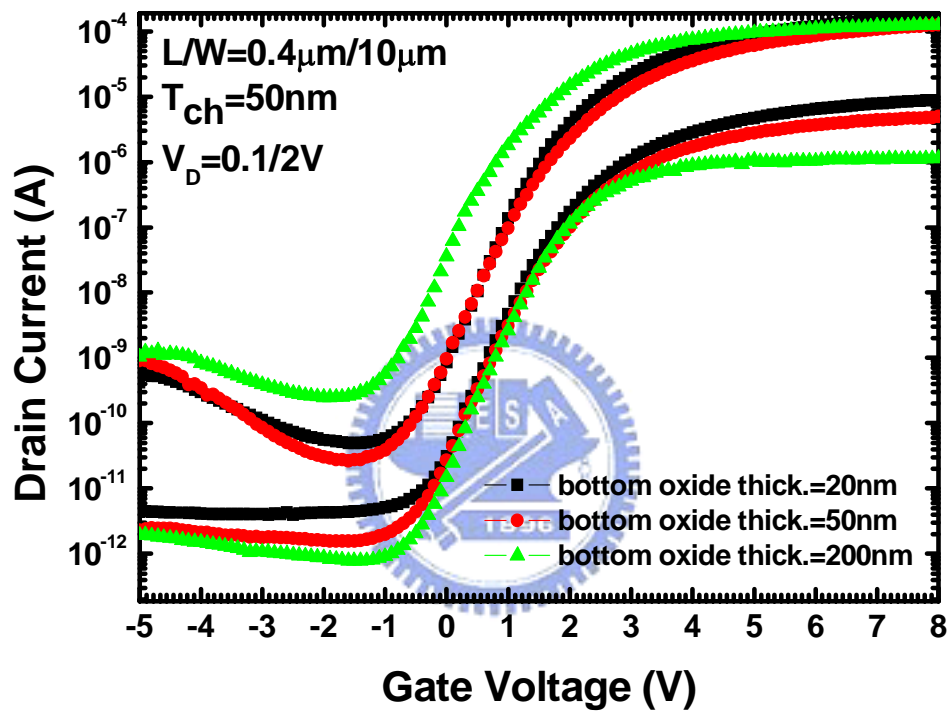


Fig. 4-4 Transfer characteristics of TFTs with the same channel thickness (50 nm) but different bottom oxide thickness. Note that all bottom gates are grounded in the measurements.

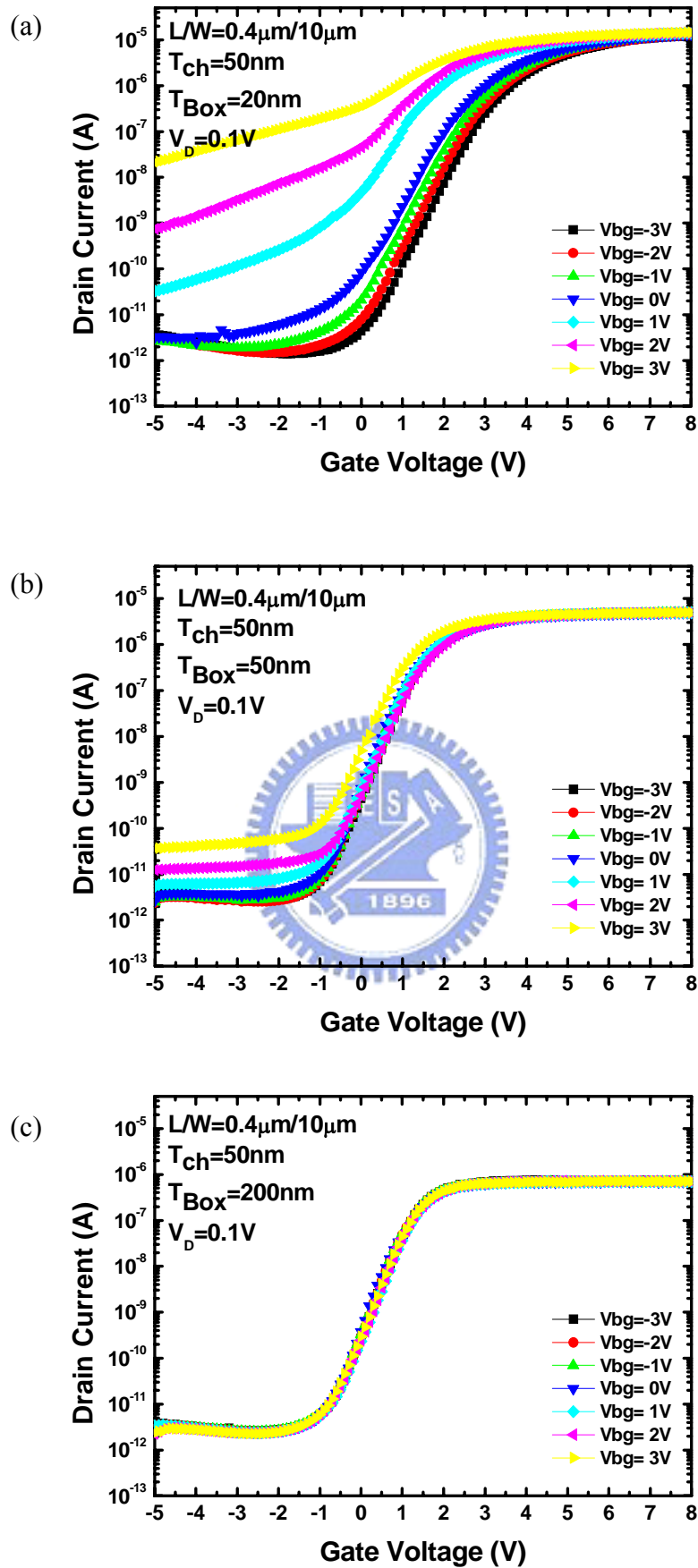


Fig. 4-5 Comparisons between transfer characteristics of TFTs with different bottom oxide thickness (a) $T_{\text{Box}} = 20\text{nm}$ (b) $T_{\text{Box}} = 50\text{nm}$ (c) $T_{\text{Box}} = 200\text{nm}$ under different V_{bg} .

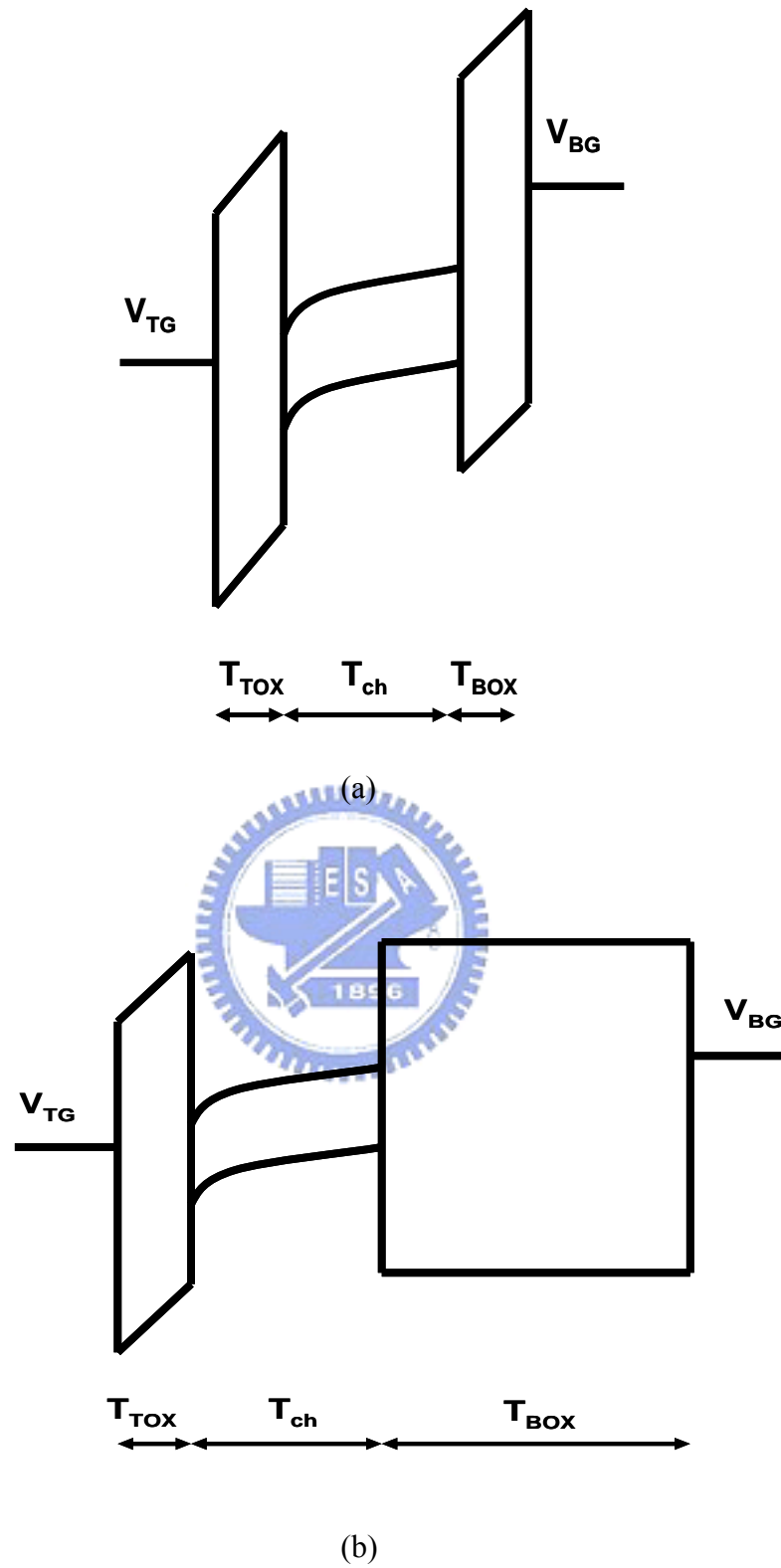


Fig. 4-6 Energy band diagram of the devices with (a) a thin and (b) a thick bottom oxide.

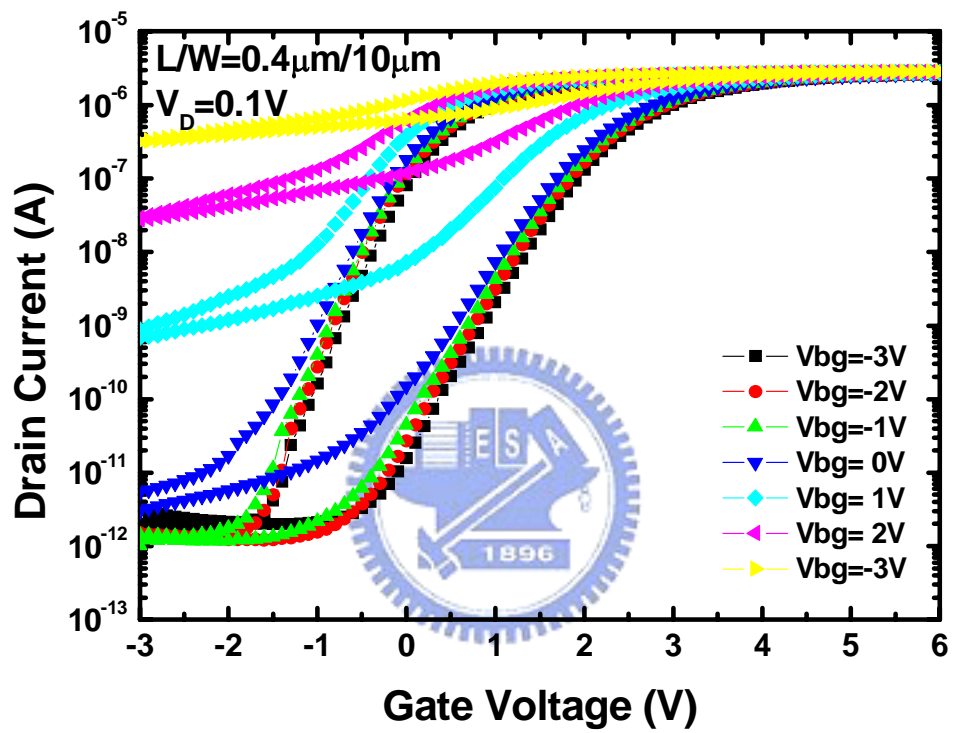
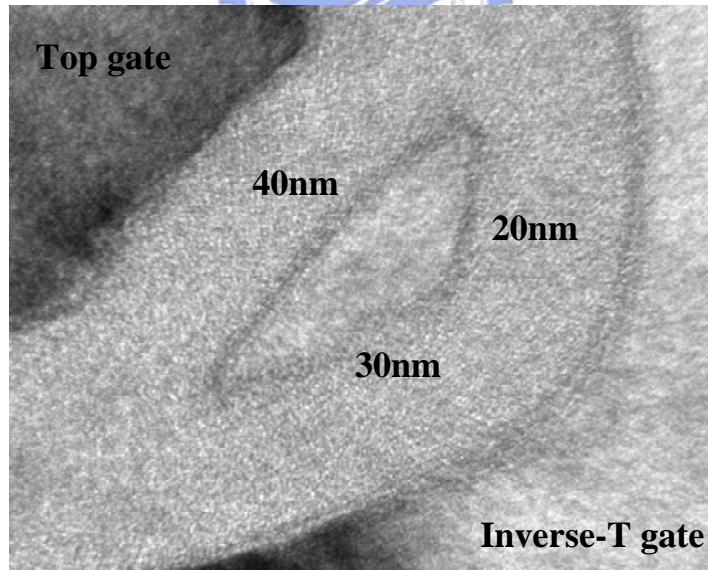
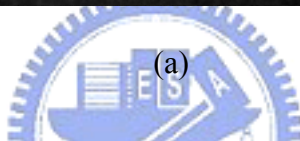
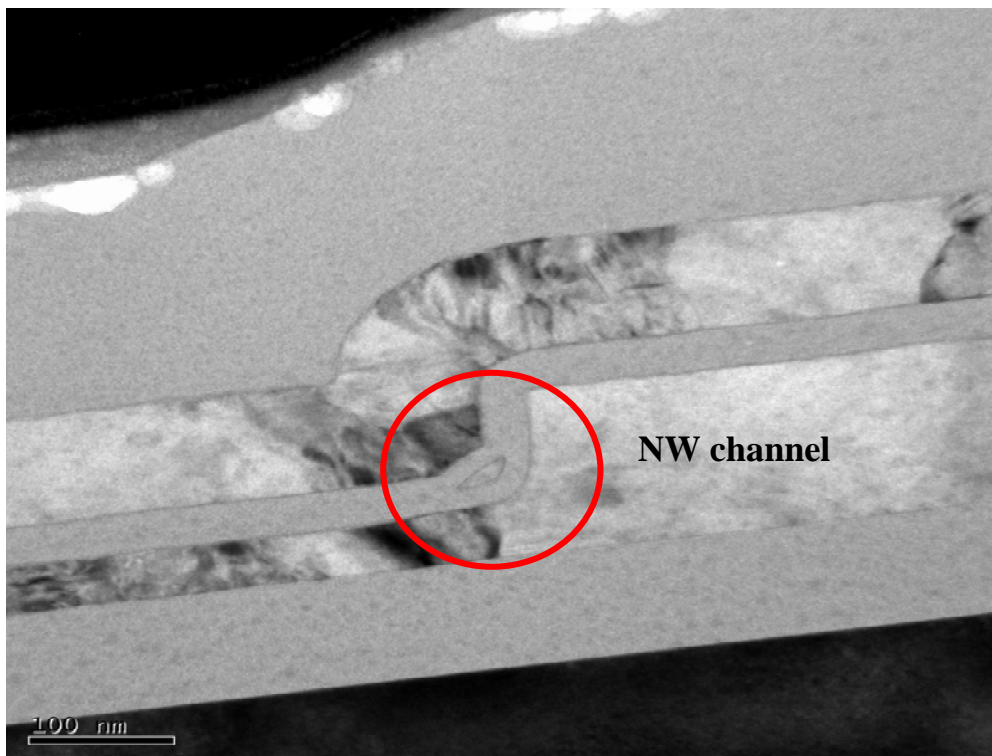


Fig. 4-7 Transfer characteristics of planar TFTs with different constant bottom gate bias under FS/RS mode.



(b)

Fig. 4-8 The cross-sectional view of ITDG-NWTFTs taken by TEM. The circled area in (a) is expanded as shown in (b).

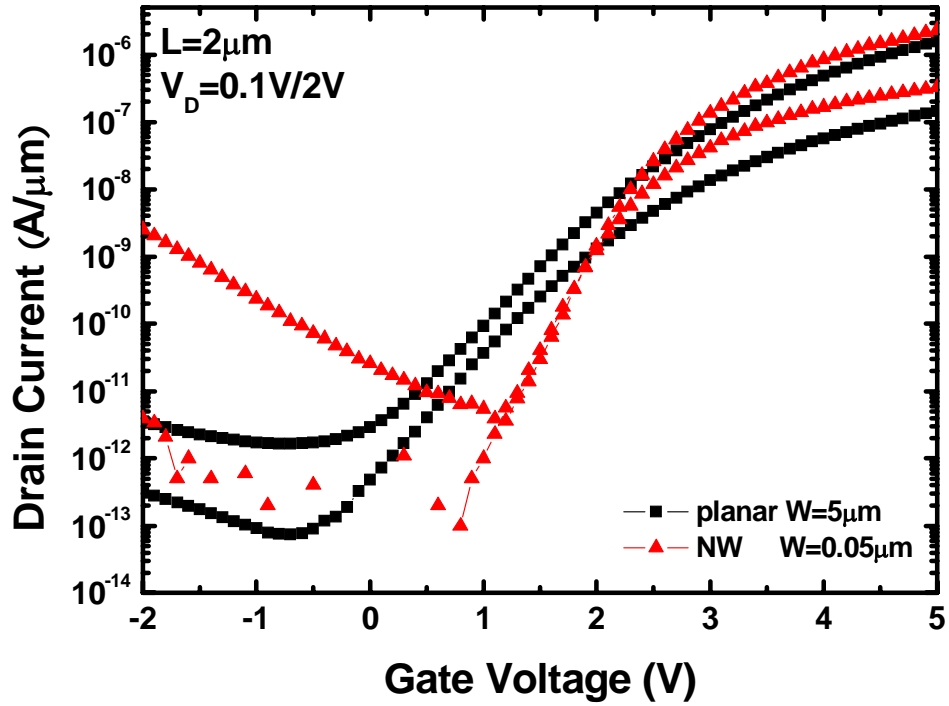


Fig. 4-9 Transfer characteristics of a planar TFT and an ITDG-NWTFT.

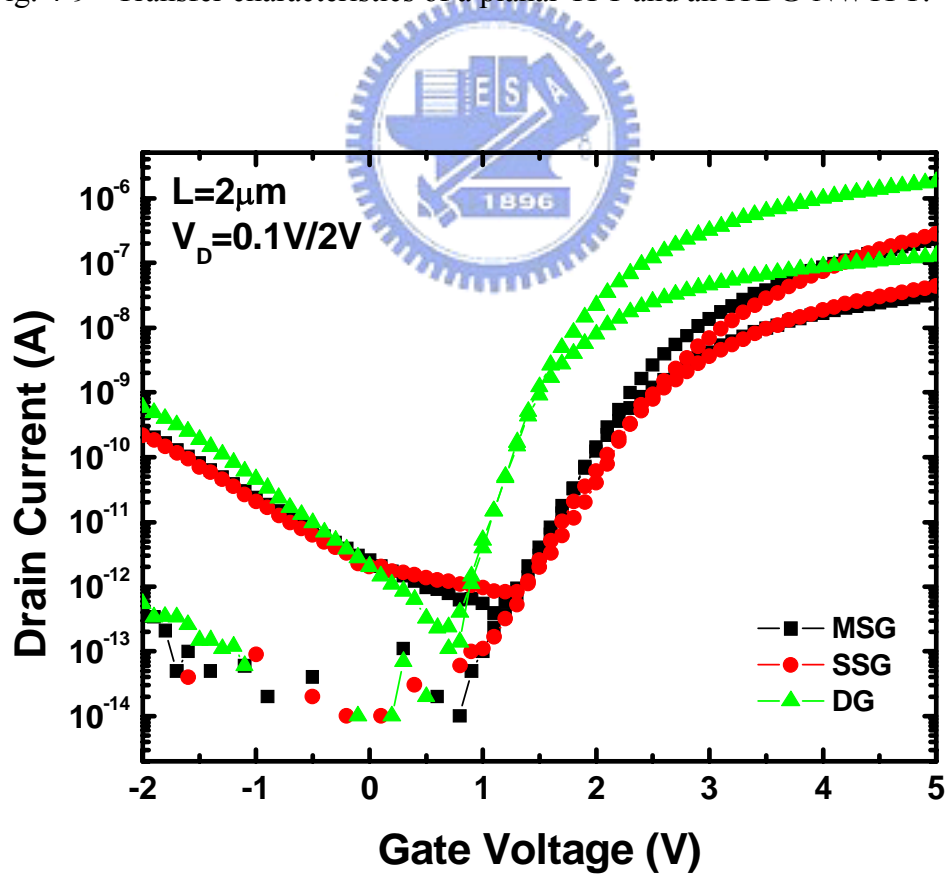


Fig. 4-10 Transfer characteristics of ITDG-NWTFTs under different operation modes.

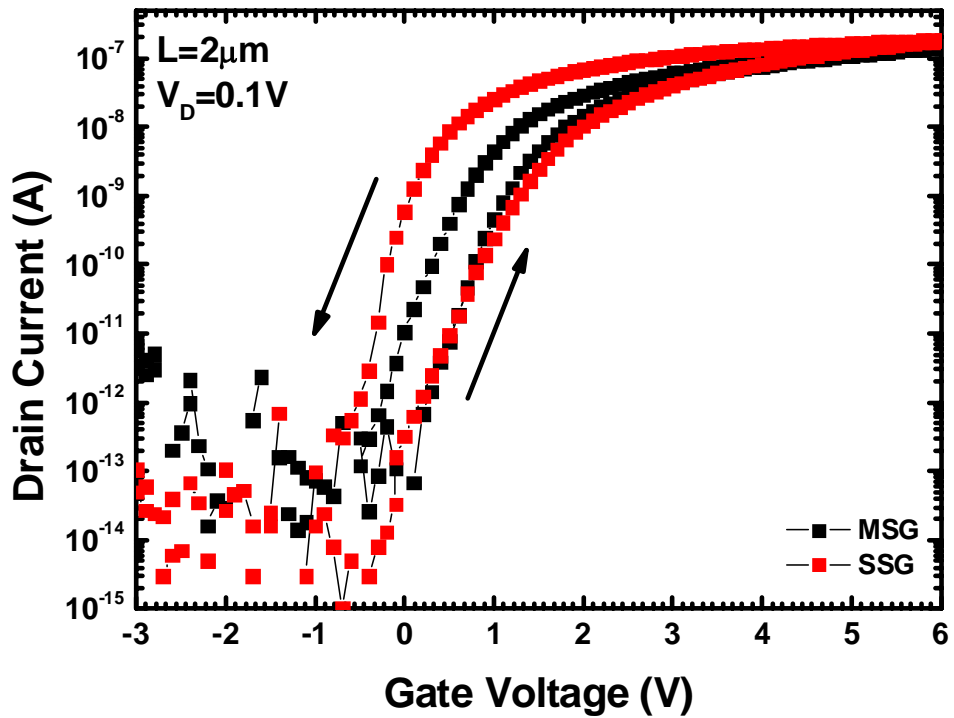


Fig. 4-11 Hysteresis characteristics of ITDG-NWTFTs under MSG and SSG modes.

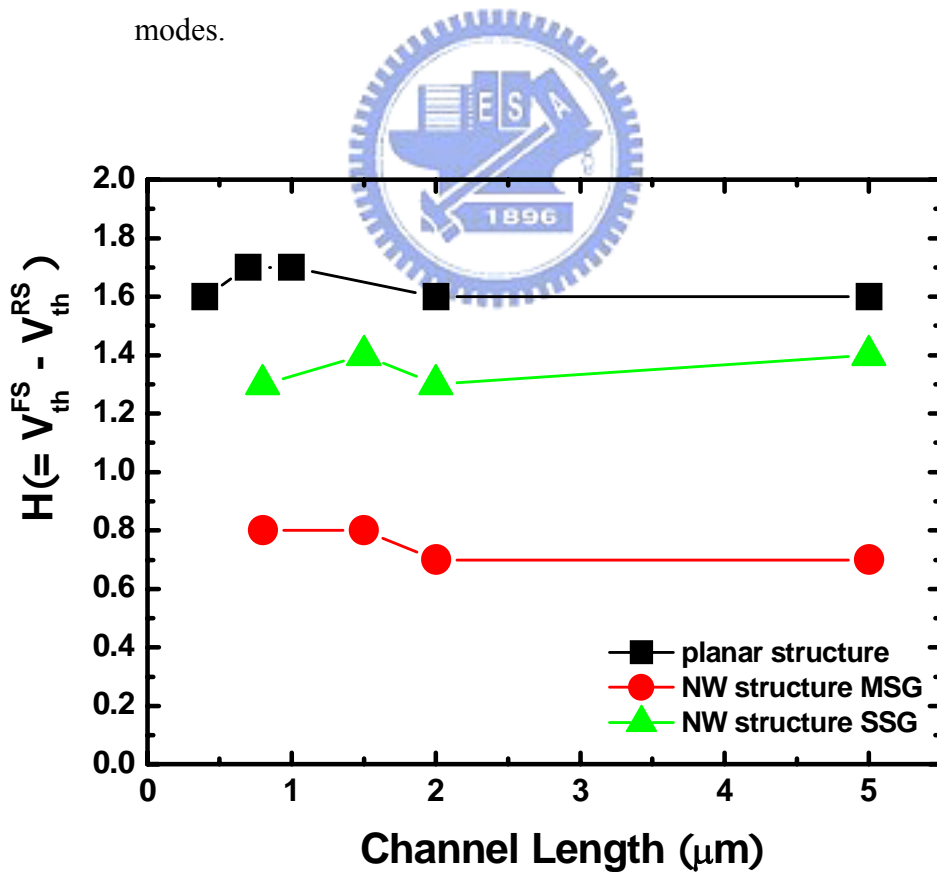


Fig. 4-12 The width of hysteresis (H) as a function of channel length for planar TFTs and ITDG-NWTFTs.

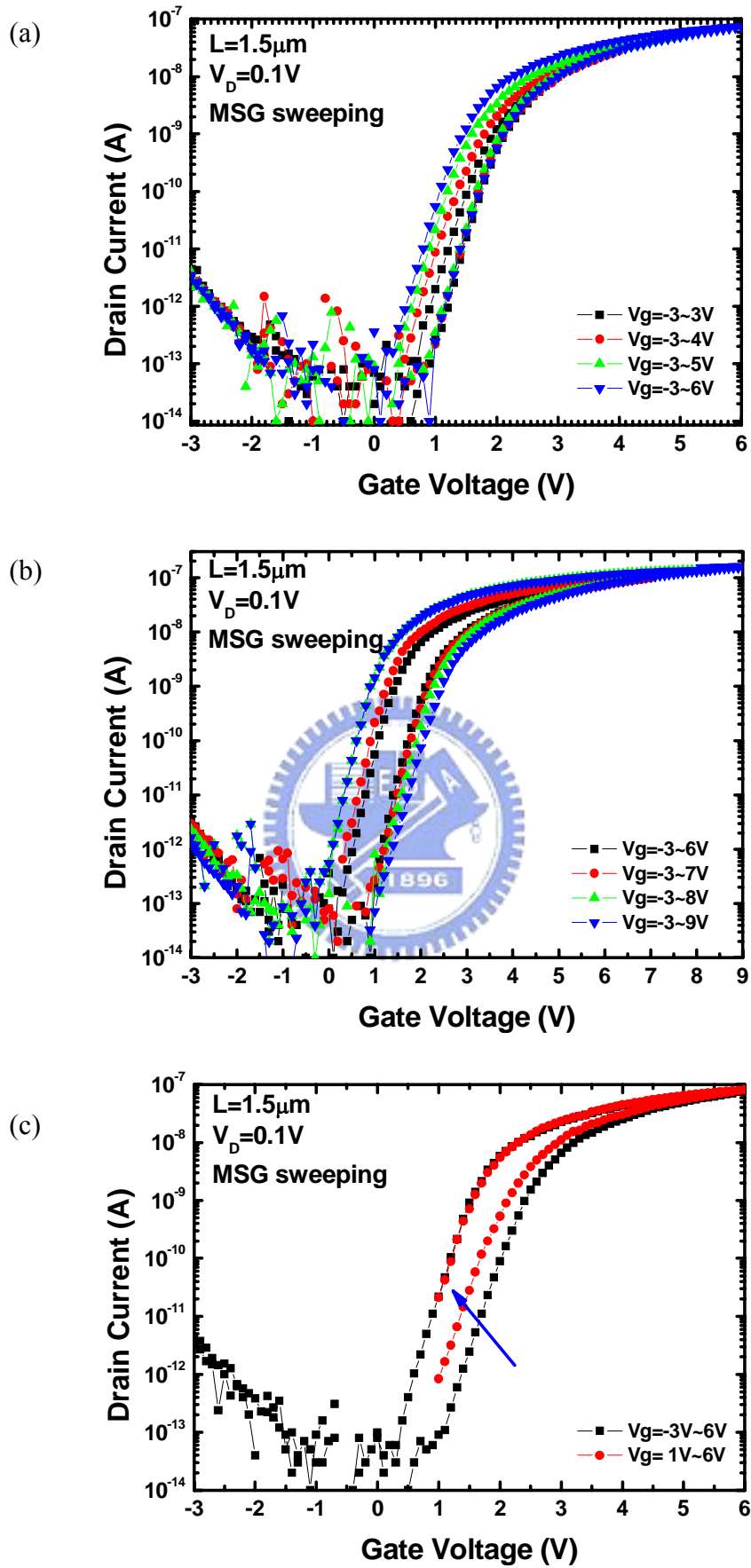


Fig. 4-13 Transfer characteristics with gate voltage variation of ITDG-NWTFTs.

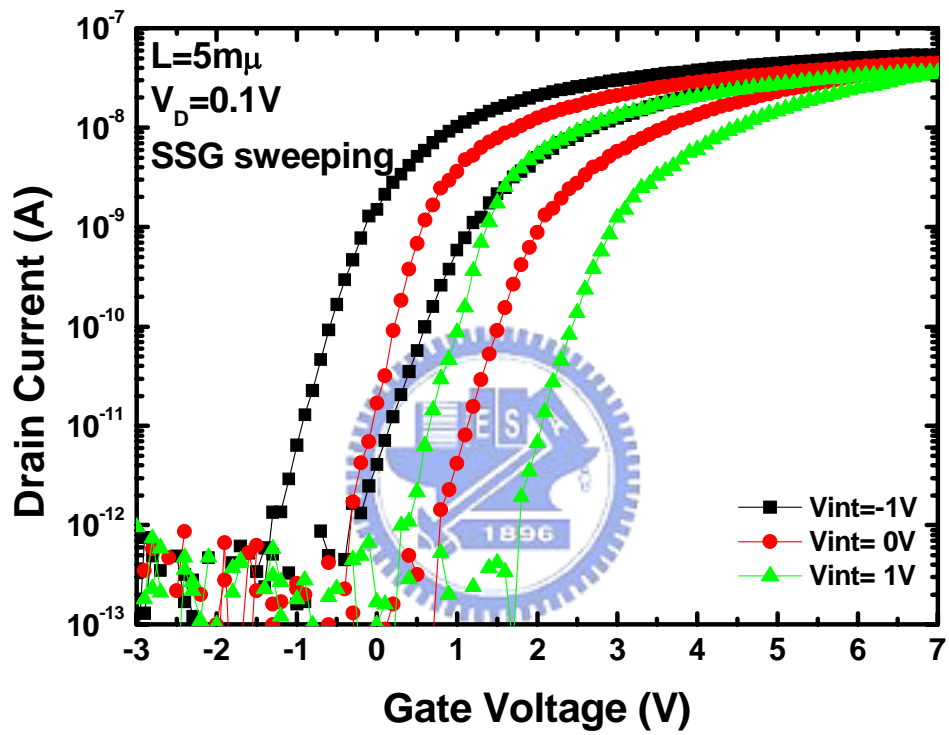


Fig. 4-14 Hysteresis characteristics of ITDG-NWTFTs under SSG mode of operation with different inverse-T gate bias.

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論文題目：複晶矽薄膜電晶體的充放電機制用於記憶體應用之研究

A Study of Trapping/De-trapping Mechanisms in Poly-Si Thin Film Transistors for Memory Applications.

Publication List

A. International Conference :

H. C. Lin, **C. H. Hung**, and T. Y. Huang, "Origin of Hysteresis phenomenon in Poly-Si Thin-Film Transistors," submitted to the IEDM'2008.

B. Patent :

1. 林鴻志，徐行徽，陳威臣，洪政雄，黃調元，“一種新式多晶矽記憶體元件”，中華民國專利(申請中)

