much lesser extent on the size distribution. The relationship between the differential attenuation and the phase allows the propagation analysis to be simplified because this analysis will depend mainly on one parameter. The importance of this simplification is apparent in the adaptive cancellation of crosspolarisation. For dual linearly polarised links, in the case where differential phase shift is the dominant cause of crosspolarisation, it is proposed to perform adaptive cancellation of the cross-polarisation by means of a simplified network with only one variable parameter.

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References

- OGUCHI, T.: Radio Sci., 1980, 16, pp. 691–730
 HOLT, A. R.: Acoustic, Electromagnetic and Elastic Wave Scattering, 1980, pp. 225–268
- JOLWORTH, J. J.: IEE colloquium on cross polar cancellation techniques, Digest No. 111, December 1985
 HARDEN, W., NORBURY, J. R., and WHITE, W.: 'Radio attenuation studies'. IEE/URS1 International Conference on Antennas and Propagation, pp. 87-91, 1978
- 5 MORRISON, J. A., and CROSS, M. J.: Bell Syst. Tech. J., 1974, 53, (4), pp. 955-1019 6 GODDARD, J. W. F., and CHERRY, S.: U.R.S.I., open symposium,
- ODDARD, J. W. F. and CHERRY, S. CHERRY, SPAR Sparsen, Bournemouth, UK, pp. 161–167, 1982
 OLALERE AJAYI, G.: "Waves using a tropical raindrop size distribu-tion", Int. J. Infrared Millim, Waves, 1985, pp. 771–806

- 9 WCEWAN, N. J., GUNES, M., and MAHMOUD, M. S.: 'RF methods for
- McEWAN, N. J., GONS, M., and MAMMOUD, M. S., KP interfoots for adaptive cancellation of cross-polarisation in microwave satellite systems'. *IEE Conf. Publ. 195*, 1981, Pt. 2, pp. 188–192
 ROGERS, D. V., and ALLNUTT, J. E.: 'Some practical considerations for depolarisation compensation in 14/11 and 14/12 GHz commu-nications satellite systems', *Electron. Lett.*, 1985, **21**, pp. 1093–1094

COMPLETE DECODING ALGORITHM OF (11, 6, 5) GOLAY CODE

Indexing terms: Codes, Error-correction codes, Decoding, Algorithms

A simple complete decoding algorithm for the (11, 6, 5) perfect ternary Golay code is presented. This algorithm is based on a step-by-step method and requires only 17 shift operations for decoding one received word.

Introduction: The three known perfect codes are the (7, 4, 3) Hamming code, the (23, 12, 7) binary Golay code and the (11, 6, 5) ternary Golay code.¹ The (11, 6, 5) Golay code is a double-error-correcting perfect code since the minimum dis-tance of this code is confirmed to be $5.^2$ Therefore, a complete decoding algorithm can be easily achieved if any combination of two or fewer errors can be corrected. Since the (11, 6, 5) Golay code is a cyclic code, the step-by-step decoding method can be employed if the different weights of received error pat-terns can be distinguished by the receiver.³ In this letter, a simple algebraic step-by-step decoding algorithm of the (11, 6, 5) Golay code is presented. This method is much faster than the method presented in Reference 2 which considers the temporary correction of two errors and hence takes a large amount of shift operations to complete the correction process.

(11, 6, 5) Golay code: Consider the (11, 6, 5) Golay code with generator polynomial $g(x) = x^5 + x^4 + 2x^3 + x^2 + 2$, the encoded codeword c(x) can be simply generated in systematic form by $K(x)x^6 - Mod \{K(x)x^6/g(x)\}$, where K(x) is the information polynomial of degree 5 and Mod $\{K(x)x^6/g(x)\}$ indi-

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cates the remainder polynomial of $K(x)x^6$ divided by g(x). The roots of this generator polynomial are confirmed to be $(\beta, \beta^3, \beta^4, \beta^5, \beta^9)$, where β is the primitive 11th root of unity.^{1,2} b, β , γ , β , β , β , β is the primitive element in GF (3⁵), it is found that $\beta = \alpha^{22}$ and therefore the arithmetic computation of this code can be performed in GF (3⁵). Furthermore, since {0, 1, 2} are the elements of GF (3), the addition of any two elements in GF (3⁵) is accomplished by adding the corresponding polynomial term by term using modulo-3 addition. The multiplication of nonzero elements may be done by multiplication of the corresponding polynomials and reduction of the product modulo $\alpha^5 + 2\alpha + 1.^4$

Complete decoding algorithm: If the received polynomial is expressed as $\tau(x) = \tau_0 + \tau_1 x + ... + \tau_{10} x^{10}$, by choosing two roots β and β^5 , the syndrome values S_1 and S_5 of $\tau(x)$ can be obtained from

$$S_i = \tau(\beta^i) \qquad i = 1, 5$$

= Mod { $\tau(x)/g(x)$ }_{|x=a²²ⁱ} $i = 1, 5$
= $\sum_{i=1}^{2} Y_i X_i^i$ $i = 1, 5$ (1)

where X_j is the error locator of *j*th error symbol and Y_j is the corresponding error value. Since Y_j can only be 0, 1 or 2, it is found that if only one error has occurred, then

$$(S_1)^5 = (Y_1X_1)^5 = Y_1X_1^5 = S_5$$
⁽²⁾

It implies that $T_5 = (S_1)^5 + 2S_5 = 0$ if one error has occurred. Moreover, as confirmed by computer simulation, we found that $T_5 \neq 0$ if two or three errors have occurred. Clearly, $S_1 = T_5 = 0$, if no error occurs. Thus, the relationships of the syndrome values for various weights of error patterns can be concluded as follows:

(i) if there is no error, then $S_1 = T_5 = 0$

(ii) if there is one error, then $S_1 \neq 0$ and $T_5 = 0$

(iii) if there are two or three errors, then $S_1 \neq 0$ and $T_2 \neq 0$.

Furthermore, the syndrome values of a cyclic shift of $\tau(x)$, $S_1^{(1)}$ and $S_5^{(1)}$, can be simply obtained by shifting the syndrome generator of g(x) once with initial contents S_i (Reference 5, theorem 8.7). Thus,

$S_i^{(j)} = \text{Mod } \{ \tau^{(j)}(x) / g(x) \}_{|x = a^{22i}}$

 $0 < j \le 11, i = 1, 5$ (3)

are the syndrome values of $\tau^{(j)}(x) = \tau_{11-j} + \tau_{12-j} x + \ldots + \tau_0 x^j + \ldots + \tau_{10-j} x^{10}$. Since the relationships between $S_1^{(j)}$ and $S_5^{(j)}$ change only when the weight of the error pattern changes and are independent of the cyclic shift of $\tau(x)$, a complete decoding algorithm of the (11, 6, 5) Golay code is then presented as follows:

(a) Calculate the syndrome values S_1 and S_5 . If $S_1 = 0$ then read out $\tau(x)$ and end this algorithm; otherwise, calculate T_5 .

(c) Find
$$S_i^{(j)}$$
, $i = 1, 5$.

(b) i = 1.

(d) If $T_5 = 0$ then calculate $Z_1^1 = S_1^{(j)} + 1$ and $Z_1^2 = S_1^{(j)} + 2$. If $Z_1^1 = 0$ then replace τ_{11-j} with $\tau_{11-j} + 1$; if $Z_1^2 = 0$ then replace τ_{11-j} with $\tau_{11-j} + 2$. Go to step (f).

(e) If $T_5 \neq 0$ then calculate $Z_5^1 = (S_1^{(j)} + 1)^5 + 2(S_5^{(j)} + 1)$ and $Z_5^2 = (S_1^{(j)} + 2)^5 + 2(S_5^{(j)} + 2)$. If $Z_5^1 = 0$ then replace τ_{11-j} by $\tau_{11-j} + 1$; if $Z_5^2 = 0$ then replace τ_{11-j} by $\tau_{11-j} + 2$.

(f) If j = 6 then the decoding algorithm is completed; otherwise, j = j + 1 and go to step (c).

This step-by-step decoding algorithm needs only 17 shift operations to decode one received word, where 11 shift operations are used for calculating the syndrome values S_1 and S_5 in step (a) and the other 6 shift operations are used for correcting the errors in the information part.

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Conclusions: A simple algebraic step-by-step decoding algo-rithm of (11, 6, 5) Golay code in systematic form has been presented. This decoding algorithm requires only 17 shift operations to decode one received word. Since the decoding algorithm makes use of the cyclic properties of the code and requires only the calculation of the value of $(S_1)^5 + 2S_5$ in GF (3⁵), the algorithm can be easily implemented in hardware by employing ternary-state logic gates.

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References

- 1
- 2
- MACWILLIAMS, F. J., and SLONE, N. J. A.: 'The theory of error correct-ing code' (North-Holland Mathematical Library, 1978) CHIEN, R. T., and LUM, V.: 'On Golay's perfect codes and step-by-step decoding', *IEEE Trans.*, 1966, **TI-12**, pp. 403–404 HARTMANN, C. R. P.: 'A note on the decoding of double-error-correcting binary BCH codes of primitive length', *IEEE Trans.*, 1971, IT.17, pp. 765–766
- correcting binary BCH codes of primitive length', *IEEE Trans.*, 1971, IT-17, pp. 765-766 MICHELSON, A. M., and LEVESQUE, A. H.: 'Error-control techniques for digital communication' (John Wiley & Sons, Inc., 1985) PETERSON, w. W., and WELDON, E. J., R.: 'Error-correcting codes' (2nd ed., MIT press, Cambridge, 1972)

JUSTESEN CONSTRUCTION FOR POLYNOMIAL REDUNDANT RESIDUE CODES

Indexing terms: Codes, Information theory, Error-correction codes, Polynomials

The Justesen construction for asymptotically good codes is described in terms of polynomial redundant residue codes by constructive methods using polynomial operations. Irreduc-ible polynomials of higher degree give asymptotic limits equivalent to RS codes. A different construction gives higher rate codes with asymptotic results similar to the punctured Justesen codes.

Redundant residue $codes^2$ are a general class of linear maximum distance separable codes. In a certain sense, they can be said to contain Reed-Solomon (RS), Goppa and alternant codes. The Justesen codes1 were constructed with RS codes as the outer codes in a concatenated scheme. In this letter it is shown how redundant residue codes, instead of RS codes, can be used in a similar construction. The construction and proofs involve operations on polynomials modulo irreducible polynomials.

Let u(x) be a polynomial of degree less than mK with coefficients over a field F. Let $m_0(x)$, $m_1(x)$, $m_2(x)$, ..., $m_{N-1}(x)$ be N relatively prime polynomials (called the moduli) each of degree *m*, where $N \ge [\deg u(x)]/m$ and N > K. Then the word or vector $c = [r_0(x), r_1(x), r_2(x), \dots, r_{N-1}(x)]$ is a code word of a (N, K, N - K + 1) linear maximum distance separable code, where $r(x) = u(x) \mod m(x)$. This code is called a redundant residue code since $r_i(x)$ is the residue of u(x) modulo $m_i(x)$. The term redundant comes from the fact that u(x) can be recovered from c by K residues $r_0(x)$, $r_1(x)$, ..., $r_{K-1}(x)$ if $mK > \deg u(x)$ by means of the Chinese remainder theorem. The residues $r_{\mathbf{K}}(x), r_{\mathbf{K}+1}(x), \dots, r_{N-1}(x)$ are redundant residues which lead to the distance $N - \mathbf{K} + 1$ for this code. If $m(x) = x - \alpha^{i}$ where α^i are the elements of a finite field $GF(q^h)$, the code is a Reed-Solomon (RS) code where u(x) is a polynomial over $GF(q^h)$ which carries the information.

Let C be a (N, K, N - K + 1) redundant residue code over Let C be a $(n, K, N \to K + 1)$ redundant residue code over a field F where $a = [a_0(x), a_1(x), \dots, a_{N-1}(x)]$ is a codeword of C, and $a_i(x) = u(x) \mod m_i(x)$ for deg u(x) < mK. Let b be the vector

 $b = [a_0(x), v_0(x); a_1(x), v_1(x); \dots; a_{N-1}(x), v_{N-1}(x)] \quad (1)$

where $v_i(x) = a_i(x)v(x) \mod m_i(x)$ for $0 \le i \le N - 1$, and where v(x) is a polynomial of degree *m* over *F* so that the members of the set $m_0(x)$, $m_1(x)$, ..., $m_{N-1}(x)$, v(x) are relatively prime. It is now shown that every nonzero pair of polynomials $[a_i(x),$ $v_i(x)$ is distinct, i.e. $[a_i(x), v_i(x)] \neq [a_i(x), v_i(x)]$ for $i \neq j$ and $a_i(x)$ and $a_j(x)$ both nonzero. For the obverse, $[a_i(x), v_i(x)] =$ $a_i(x)$ and $a_i(x)$ been divided in the order of $a_i(x)$, $a_i(x) = a_i(x)$ and $v_i(x) = v_i(x)$, which means that $a_i(x)v(x) \mod m_i(x) \equiv a_i(x)v(x) \mod m_j(x)$. Since deg $a_i(x)v(x) \ge m$, then

$$a_{i}(x)v(x) - c_{i}(x)m_{i}(x) = a_{i}(x)v(x) - c_{i}(x)m_{i}(x)$$
(2)

where $c_i(x)$, $c_i(x) \neq 0$. Therefore $c_i(x)m_i(x) = c_i(x)m_i(x)$. Since $m_1(x)$, $m_1(x)$, $m_1(x)$ are relatively prime, then it is necessary that $c_1(x) = e_1(x)m_1(x)$, $c_2(x) = e_1(x)m_1(x)$, where $e_1(x)$ and $e_2(x)$ are both nonzero. As a result, deg $c_1(x) \ge m$ and deg $c_1(x) \ge m$, whereas deg $a_i(x)(x) \le 2m - 1$, so that eqn. 2 is not a polynomial of degree less than m. This gives a contradiction since $v_i(x)$ must have degree less than m. If in eqn. 1, $a_i(x) \ne 0$, then $v_i(x) \neq 0$ since $m_i(x)$ and v(x) are relatively prime. Since the code C has distance N - K + 1, then there are at least N - K + 1 distinct pairs $[a_i(x), v_i(x)]$ in eqn. 1.

The redundant residue code B, composed of words such as eqn. 1, is linear, if we assume two code words generated by the information polynomials u(x) and u'(x). Then the sum of coordinate pairs $[a_i(x), v_i(x)]$ and $[a'_i(x), v'_i(x)]$ is $[a_i(x) + a'_i(x), v_i(x) + v'_i(x)] = [[(u(x) + u'(x)] \mod m_i(x), [a_i(x) + a'_i(x)]v(x) \mod m_i(x), [a_i(x) + a'_i(x)]v(x) \mod m_i(x)$ $m_1(x)$]. This code can be considered a concatenated code for the same reasons as for the Justesen code using RS codes as the outer codes, and has length 2mN symbols from F and mK information symbols, giving a rate of K/2N. If $m_i(x) = x - \alpha^i$, $0 \le i \le q^h - 1$, where α is a primitive element of $GF(q^h)$ and $0 \le i \le d^{2} - 1$, where α is a primitive element of $G'(q^{2})$ and y(x) = x, then this construction yields the original Justesen construction.¹ This is because $a_{i}(x) = u(x) \mod x - \alpha^{i} = u(\alpha^{i})$ and $x \mod x - \alpha^{i} = \alpha^{i}$, so that $a_{i}(x)v(x) \mod m_{i}(x) \equiv [a(x)][(x) \mod m(x)] = \alpha^{i}(\alpha^{i})$. Now assume the m(x) are irreducible polynomials of degree

m over GF(2). The length of the resulting code (eqn. 1) will be determined by the number I_m of such polynomials where lim $I_m/(2^m/m) = 1$, as $m \to \infty$. Now let $y_0, y_1, y_2, ..., y_{M-1}$ be a set of M distinct binary vectors, each of length 2m, and let w_1 be the Hamming weight of y_1 . Let $p = (w_0 + ... + w_{M-1})/2mM$. Then³ log $M \le 2mH_2(p)$ where H_2 is the binary entropy function. For the above redundant residue code B, M = N - K + 1 where $N = I_m$, and each pair $y_j = [a_j(x), v_j(x)]$ has components of degree less than m and length 2*m*. The polynomial v(x) can be of the form $t_1(x)t_2(x)$, where $t_1(x)$ and $t_2(x)$ are irreducible and deg $t_1(x) + \text{deg } t_2(x) = m$. Thus all I_m irreducible binary polynomials of degree m can be used for the moduli $m_i(x)$. The rate R of B is K/2N, and since the distance of *C* is D = N - K + 1, then D/N = 1 - K/N + 1/N > 1 - K/N = 1 - 2R. Then the ratio of distance *d* to length 2mN of code, in bits, is

d/2mN[d/2m(N-K+1)][(N-K+1)/N]

 $\geq [(N - K + 1)/N]H_2^{-1}[(1/2m)\log(N - K + 1)]$

 $\geq (1-2R)H_2^{-1}[(1/2m)\log N(1-2R)]$

 $\geq (1-2R)H_2^{-1}\{(1/2m) \log \left[(2^m/m)(1-2^{-m/2+1})(1-2R)\right]\}$

 $= (1 - 2R)H_2^{-1}\{(m/2m) + (1/2m)\log(1 - 2^{-m/2 + 1})\}$

 $+ \log (1 - 2R) - \log m]$

 $\rightarrow (1 - 2R)H_2^{-1}\{(1/2) + (1/2m)[\log (1 - 2R) - \log m]\}$

 $\rightarrow (1 - 2R)H_2^{-1}(1/2)$

as $m \to \infty$, since $\log (1 - 2^{-m/2 + 1}) = -2^{-m/2 + 1} + O(2^{-m+2})$ which tends to zero as $m \to \infty$. Also, $(\log m)/m \to 0$ as $m \to \infty$. Thus we get the same asymptotic result as the Justesen codes using RS codes.

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