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Simulation of characteristic variation in 16 nm gate FinFET devices due to intrinsic parameter fluctuations

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Abstract

High- κ /metal-gate and vertical channel transistors are well-known solutions to continue the device scaling. This work extensively explores the physics and mechanism of the intrinsic parameter fluctuations in nanoscale fin-type field-effect transistors by using an experimentally validated three-dimensional quantum-corrected device simulation. The dominance fluctuation sources in threshold voltage, gate capacitance and cutoff frequency have been found. The emerging fluctuation source, workfunction fluctuation, shows significant impacts on DC characteristics; however, its impact is reduced in AC characteristics due to the screening effect of the inversion layer. Additionally, the channel discrete dopant may enhance the electric field and therefore make the averaged cutoff frequency of fluctuated devices larger than the nominal value of cutoff frequency.

(Some figures in this article are in colour only in the electronic version)

1. Introduction

As the dimensions of semiconductor devices continue to be reduced, device variability is pronounced [1–8]. Fin-type field-effect transistors (FinFETs) [7–9] and high- κ /metal-gate technology [10] are promising for semiconductor manufacturing. However, the use of metal as a gate material may introduce additional random fluctuations, workfunction fluctuation (WKF), due to the uncontrollable grain orientations of metal during the growth period. The WKF-induced threshold voltage (V_{th}) fluctuation on planar MOSFETs has been reported [10]. The different orientations of crystal structure may have different surface densities and workfunctions [11]. Since the device gate area is thus composed of a number of grains with different workfunctions, the workfunction and threshold voltage of the device are probabilistic distributions rather than deterministic values.

In studying characteristic fluctuations of FinFETs, diverse approaches have recently been presented [7, 8]; however, they only focused on the random-dopant fluctuation (RDF) and

process-variation effect (PVE) on device DC characteristics, such as V_{th} and current fluctuation. The effects of WKF on FinFET and these intrinsic parameter fluctuations on AC characteristics are not clear. Thus, a comprehensive understanding of the intrinsic parameter fluctuations on FinFETs is required. This study extensively explores the intrinsic device parameter fluctuations (WKF, PVE and RDF) on 16 nm gate silicon-on-insulator (SOI) FinFETs by an experimentally validated three-dimensional quantum-corrected device simulation [4]. Both the current–voltage characteristics and the physical models of FinFET have been calibrated with experimentally fabricated FinFETs [9]. The major variability sources in DC and AC characteristics of FinFETs are provided for the first time. This study explores the fluctuations on FinFET devices, which can in turn be used to optimize the reliability of nanoscale systems.

This paper is organized as follows. Section 2 introduces the simulation technique for studying the effect of intrinsic parameter fluctuations in nanoscale devices and circuits. Section 3 studies the characteristic fluctuations in 16 nm gate

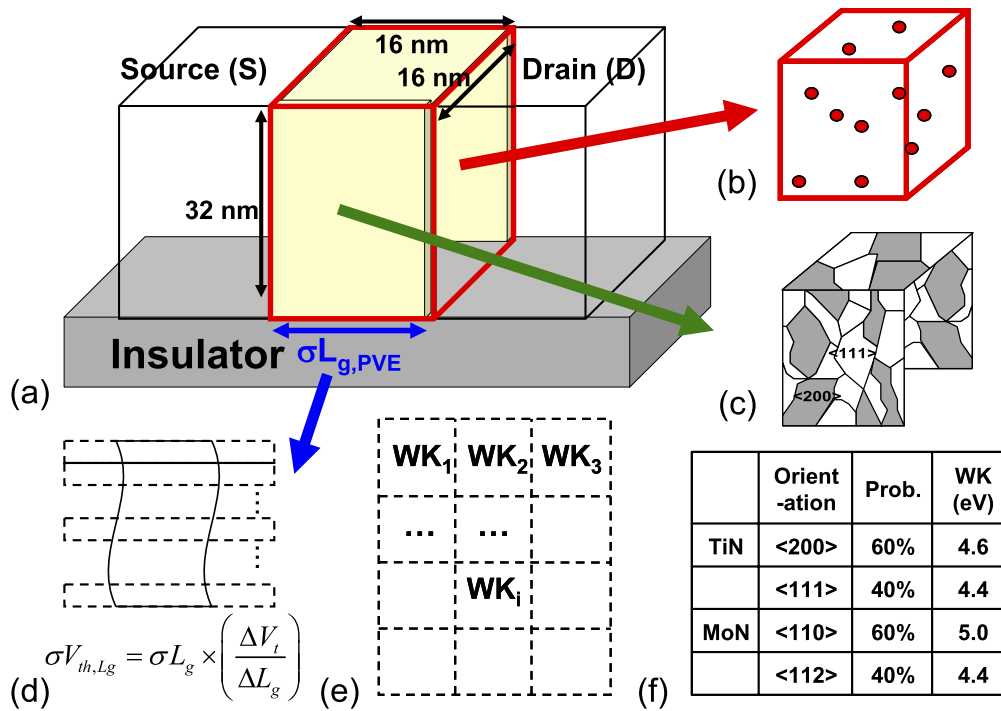


Figure 1. (a) An illustration of the explored SOI FinFETs with (b) random-dopant fluctuation, (d) process-variation effect and (c) workfunction fluctuation. The number of channel dopants in the device may vary from 2 to 22, and the average number is 13. (d) The discretization scheme for estimating PVE-induced σV_{th} and the calculation formula. (e) The estimation of workfunction fluctuation. The gate area is partitioned into several pieces according to the average grain size. The workfunction of each partitioned area (WK_i) is then randomized according to the probability distribution in (f). The effect workfunction of a single device is obtained and used from estimation of workfunction-variation-induced fluctuation.

devices and circuits. Finally, conclusions are drawn and future work is suggested.

2. The statistical simulation technique

Figure 1(a) illustrates the explored 16 nm gate SOI FinFETs with amorphous-based TiN/HfSiON gate stacks with an equivalent oxide thickness (EOT) of 1.2 nm [10]. The equivalent channel doping concentration is $1.48 \times 10^{18} \text{ cm}^{-3}$. To compare them on the same basis, the device dimension and V_{th} for both n- and p-type devices are the same. Figure 1(b) illustrates the RDF effect in transistors, where the simulation scenario mainly follows our recent work [4, 7]. The device simulation is performed by solving a set of three-dimensional ‘atomistic’ drift–diffusion equations with quantum corrections by the density gradient method [4, 7, 12, 13]. Note that, in ‘atomistic’ device simulation, the resolution of individual charges within classical device simulation using a fine mesh creates problems associated with singularities in the Coulomb potential [5, 6]. The potential becomes too steep with a fine mesh and therefore the majority carriers are unphysically trapped by ionized impurities and the mobile carrier density is reduced [5, 6]. Thus, the density gradient approximation is used to handle discrete charges by properly introducing the quantum mechanical effects [4, 7, 12, 13].

The mobility model is quantified with our device measurements for the best accuracy and the characteristic fluctuation has been validated with the experimentally

measured DC base band data [4, 9]. The device simulation is performed by solving a set of 3D drift–diffusion equations with quantum corrections by the density gradient method [4, 7, 12, 13], which is conducted using a parallel computing system [14]. The mobility model used in the device simulation, according to Matthiessen’s rule [15], can be expressed as

$$\frac{1}{\mu} = \frac{D}{\mu_{\text{surf_aps}}} + \frac{D}{\mu_{\text{surf_rs}}} + \frac{1}{\mu_{\text{bulk}}}, \quad (1)$$

where $D = \exp(x/l_{\text{crit}})$, x is the distance from the interface and l_{crit} is a fitting parameter. The mobility consists of three parts: (1) the surface contribution due to acoustic phonon scattering $\mu_{\text{surf_aps}} = (B/E) + [C(N_i/N_0)^\tau/E^{1/3}(T/T_0)^K]$, where $N_i = N_A + N_D$, $T_0 = 300 \text{ K}$, E is the transverse electric field normal to the interface of semiconductor and insulator, B and C are parameters which are based on physically derived quantities, N_0 and τ are fitting parameters, T is lattice temperature and K is the temperature dependence of the probability of surface phonon scattering; (2) the contribution attributed to surface roughness scattering is $\mu_{\text{surf_rs}} = ((E/E_{\text{ref}})^\chi/\delta + E^3/\eta)^{-1}$, where $\chi = A + \alpha(n + p)N_{\text{ref}}/(N_i + N_1)^v$, $E_{\text{ref}} = 1 \text{ V cm}^{-1}$ is a reference electric field to ensure a unitless numerator in $\mu_{\text{surf_rs}}$, $N_{\text{ref}} = 1 \text{ cm}^{-3}$ is a reference doping concentration to cancel the unit of the term raised to the power v in the denominator of χ , δ is a constant that depends on the details of the technology,

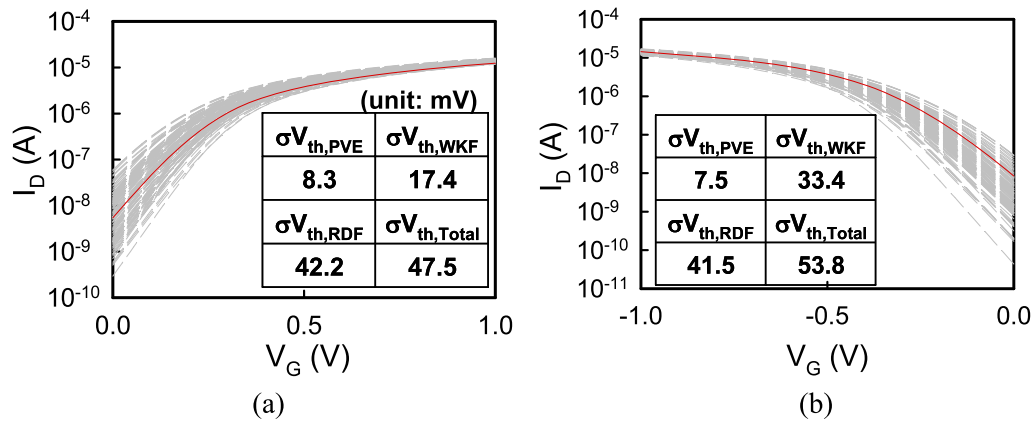


Figure 2. The I_D - V_G characteristics for (a) n-type and (b) p-type SOI FinFETs. The insets are the components of σV_{th} , in which the drain bias is 1 V.

such as oxide growth conditions, $N_1 = 1 \text{ cm}^{-3}$, and A and η are fitted field-dependent coefficients. α is a fitting parameter for doping concentration (3) and the bulk mobility is $\mu_{\text{bulk}} = \mu_L(T/T_0)^{-\xi}$, where μ_L is the mobility due to bulk phonon scattering and ξ is a fitting parameter for the temperature dependence of mobility. The mobility model is quantified with our device measurements for the best accuracy and the characteristic fluctuation has been validated with the experimentally measured DC base band data [4].

The PVE includes gate length deviation (σL_g) and line edge roughness (σL_{ER}) whose magnitude follows the ITRS roadmap that $3\sigma L_g = 0.7 \text{ nm}$ and $3\sigma L_{ER} = 0.8 \text{ nm}$ for the 16 nm technology node [16]. The equation of figure 1(d) presents the estimation formula and σL_g and σV_{th} are obtained from the V_{th} roll-off curve. As the PVE-induced gate length variation ($\sigma L_{g,PVE}$) is obtained, the PVE-induced V_{th} fluctuation ($\sigma V_{th,PVE}$) is calculated by characteristics of the V_{th} roll-off. The detailed simulation techniques for RDF and PVE simulation follow our previous works [17–25]. For WKF, a statistically sound Monte Carlo simulation approach is advanced, as shown in figure 1(e). Based on the average grain size of 4 nm [10], the gate area is first partitioned into several parts. Then, the workfunction of each partitioned area (WK_i) is randomized following the properties of metal in figure 1(f) [11]. We herein use titanium nitride (TiN) and molybdenum nitride (MoN) as gate material for NMOS and PMOS transistors, respectively. After partition and randomization, the effective device workfunction of a single transistor is then obtained and used for estimation of WKF-induced V_{th} fluctuation ($\sigma V_{th,WKF}$). Notably, the different processes of gate formulation, gate first or replacement gate may change the thermal budget and change the grain size of the metal material.

3. Result and discussion

Figures 2(a) and (b) show the characteristics of the drain current (I_D) versus the gate voltage (V_G). The solid lines are the nominal case with supposed device dimensions, workfunction

and doping concentration; the dashed lines are cases with intrinsic parameter fluctuations. The insets of figures 2(a) and (b) show the σV_{th} for n-type and p-type FinFETs. The total threshold voltage fluctuation ($\sigma V_{th,Total}$) is obtained by statistical addition of variances, $(\sigma V_{th,Total})^2 \approx (\sigma V_{th,WKF})^2 + (\sigma V_{th,PVE})^2 + (\sigma V_{th,RDF})^2$. In SOI FinFETs, the RDF and WKF dominate the σV_{th} in both n-type and p-type transistors. The $\sigma V_{th,WKF}$ in p-type FinFETs becomes comparable to $\sigma V_{th,RDF}$ due to the large deviation of workfunction, which implies the importance of workfunction fluctuation in nanoscale device with scaled gate area. Notably, the statistical addition of individual fluctuation sources herein simplifies the variability analysis of nanodevices significantly [3]. It is believed that the dominant source of fluctuation will not be significantly altered.

The WKF-, PVE- and RDF-fluctuated C_g are presented in figures 3(a)–(c), where the solid line shows the nominal case with 16 nm gate, $1.48 \times 10^{18} \text{ cm}^{-3}$ channel doping and the dashed lines are fluctuated cases. The different intrinsic parameter fluctuations induced rather different C_g - V_G characteristics. Figure 4 summarizes the gate capacitance fluctuations (σC_g) with 0, 0.5 and 1.0 V gate bias. Different to the results of the V_{th} fluctuation, the WKF brought less impact on gate capacitance fluctuation. At low gate bias or negative gate bias, the accumulation layer screens the impact of WKF. Additionally, at low gate bias, the total capacitance decreases because of the increased depletion region. The associated value of C_g fluctuation is small. The capacitive response is then dominated by the increment of inversion in the moderate inversion. The device characteristics are then impacted by the intrinsic parameter-fluctuated electrostatic potentials. If the high V_G is achieved, the capacitive response becomes dominated by the inversion layer and the impact of the individual dopants on the device electrostatics is screened by the inversion layer itself. The variation of capacitance now again becomes the variation of gate oxide. The impact of WKF-induced electrostatic potential variations is therefore bringing less impact on the channel surface. Our preliminary results show that the RDF and PVE dominate the gate capacitance fluctuations at all gate bias conditions, respectively. The impacts of the WKF on C_g is reduced

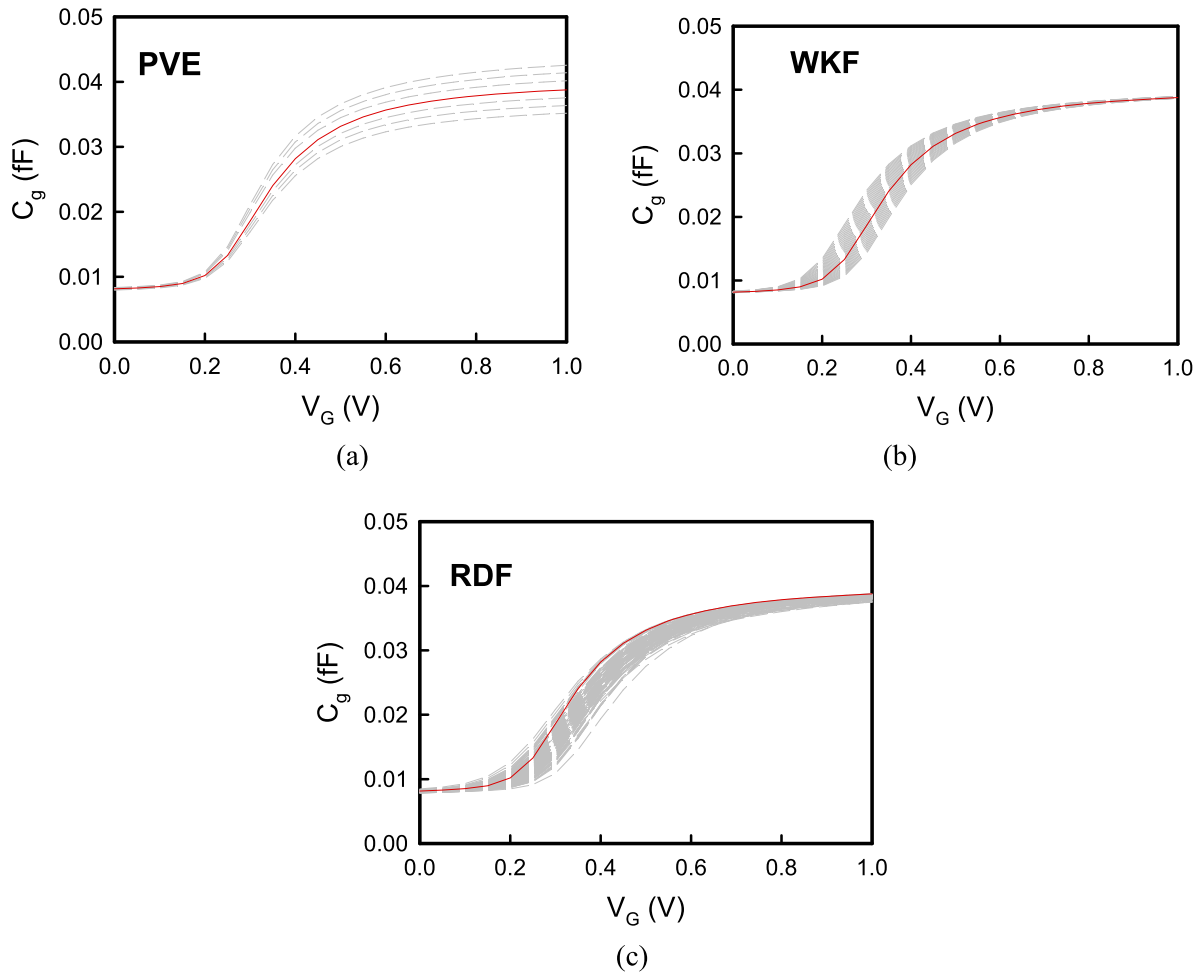


Figure 3. The C_g - V_G characteristics for the explored devices with (a) PVE, (b) WKF and (c) RDF.

significantly at high gate voltage (V_G) due to the screening effect of the inversion layer of the device, which screens the variation of surface electrostatic potential and decreases the fluctuation of gate capacitance. The screening effect resulting from the inversion layer also decreases the RDF-induced gate capacitance fluctuation at high gate bias; however, the screening effect of the inversion layer is weakened by discrete dopants positioned near the channel surface. Notably, the PVE brings direct impact on gate length and therefore influences the gate capacitance. The PVE-induced gate capacitance fluctuation is independent of screening effect and should be noticed when the transistor operated in high gate bias, as shown in figure 4.

Figures 5(a)–(c) describe the PVE-, WKF-, and RDF-induced characteristic fluctuation of the cutoff frequency ($F_T = v_{\text{sat}}/2\pi L_g = g_m/2\pi C_g$) for the n-type transistors, where g_m and v_{sat} are the transconductance and the saturation velocity, respectively. The solid lines are the nominal case; the dashed lines are the cases with intrinsic parameter fluctuation; the symbol lines are the averaged result. At low V_G , the cutoff frequency of FinFET increases as V_G increases due to the enhancement of carrier velocity. The cutoff frequency is saturated and then degrades in the high electric field due to the velocity saturation and scattering of carriers. The

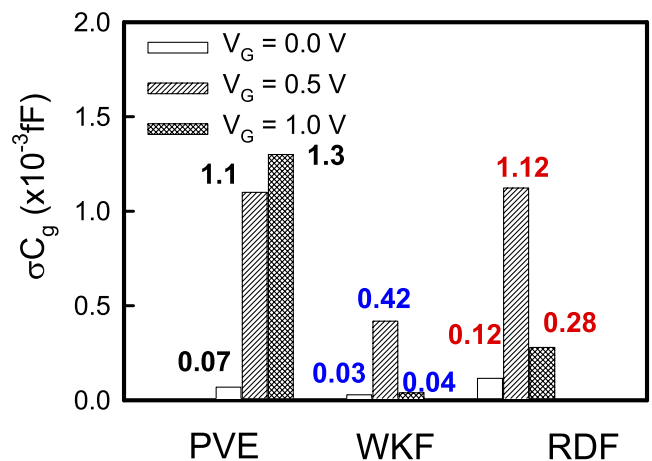


Figure 4. The C_g fluctuation at $V_G = 0, 0.5$ and 1 V for 16 nm gate FinFETs with intrinsic parameter fluctuations WKF, PVE and RDF.

PVE-induced σF_T in figure 5(a) is significant in all bias conditions owing to the direct impact of gate length variation on F_T . Different to the PVE-induced σF_T , the WKF-induced σF_T , figure 5(b), diminished as the velocity saturation occurs. The inversion layer is formed at this time and

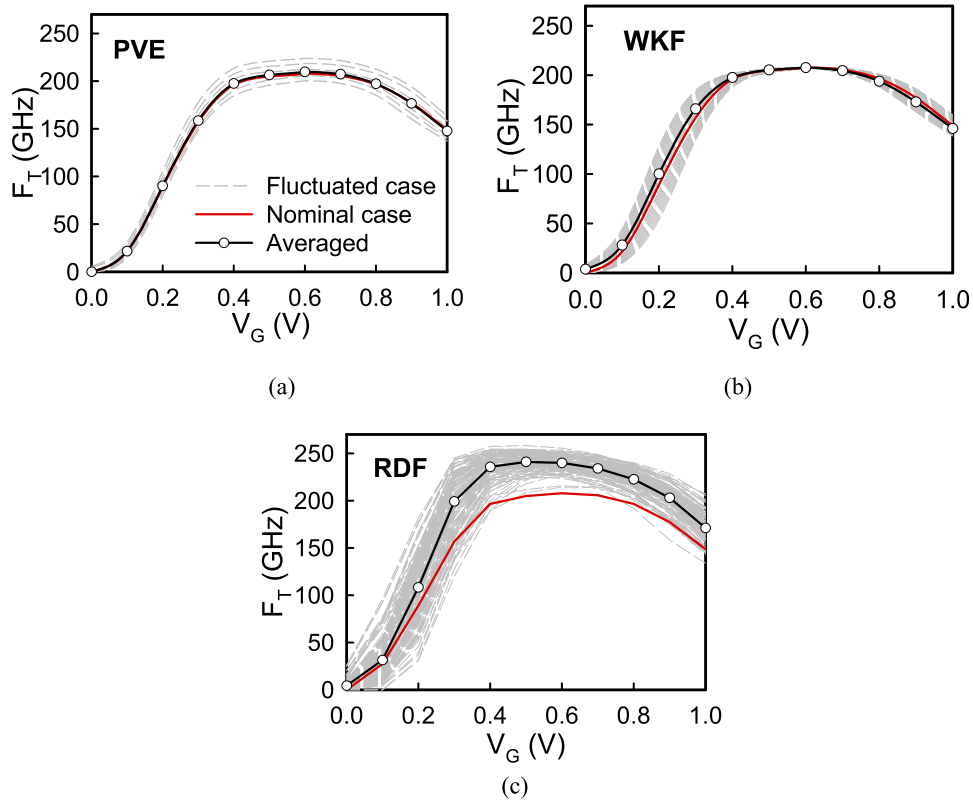


Figure 5. The F_T - V_G characteristics for the explored devices with (a) PVE, (b) WKF and (c) RDF.

screens the variation of surface electrostatic potential induced by WKF. The σF_T then becomes significant at high field because of the carrier scattering. As for the RDF-induced σF_T , as shown in figure 5(c), the σF_T does not diminish when the saturation of the carrier velocity occurs due to the randomness of carrier-impurity scattering events and carrier velocity variation. Moreover, though the screening effect at high field may also reduce the RDF-induced σF_T , the shielding effect may be broken by discrete dopants positioned near the channel surface. The RDF-induced σF_T thus dominates the fluctuation of cutoff frequency. The σF_T is summarized in figure 6, in which the RDF dominates the σF_T . Both the RDF- and WKF-induced σF_T are significant in the subthreshold region. Then the σF_T is reduced due to the screening effect of the inversion layer. The scattering effect occurring in the high-field region then increases the σF_T induced by RDF.

Notably, our preliminary result shows that the nominal and the averaged values of F_T are similar for the results of PVE and WKF. However, in RDF, the difference of the nominal and the averaged F_T becomes significant as V_G increases. For the discrete dopants in the channel, the channel dopants will induce a relatively negative potential in the channel and twist the electric field nearby, as plotted in figures 7(a) and (b). The arrows in figure 7(a) present the direction of the electrical field. Figure 7(c) presents the electric field for three cut lines at the channel surface. Line 1 is with one dopant located at channel surface; line 2 is without the impurity at the conducting path; line 3 is also has one dopant located at the channel surface, but the dopant is located 0.5 nm below the channel surface. The result shows that the dopant inside the channel may enhance

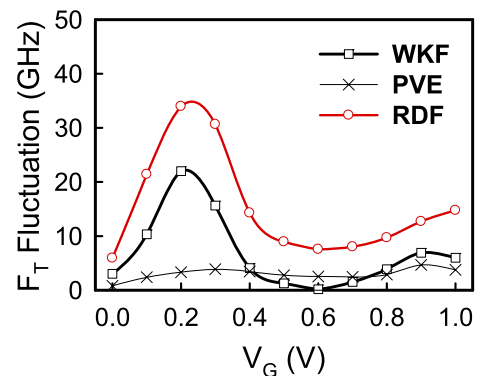


Figure 6. The summarized σF_T for the explored devices with (a) PVE, (b) WKF and (c) RDF.

the electric field nearby. The electric field strength is location-dependent and its magnitude decreases as the distance from a location to the dopant increases, as shown in line 1 and line 3 of figure 7(c). The twisted electric field then enhances the electron velocity, v_{sat} , as shown in figure 7(d) and therefore increases the average value of F_T . The average value of F_T is larger than the nominal F_T due to the enhanced electric field near the channel dopants.

4. Conclusions

This study has extensively explored the metal-gate device's DC and AC variability. The dominant variability sources

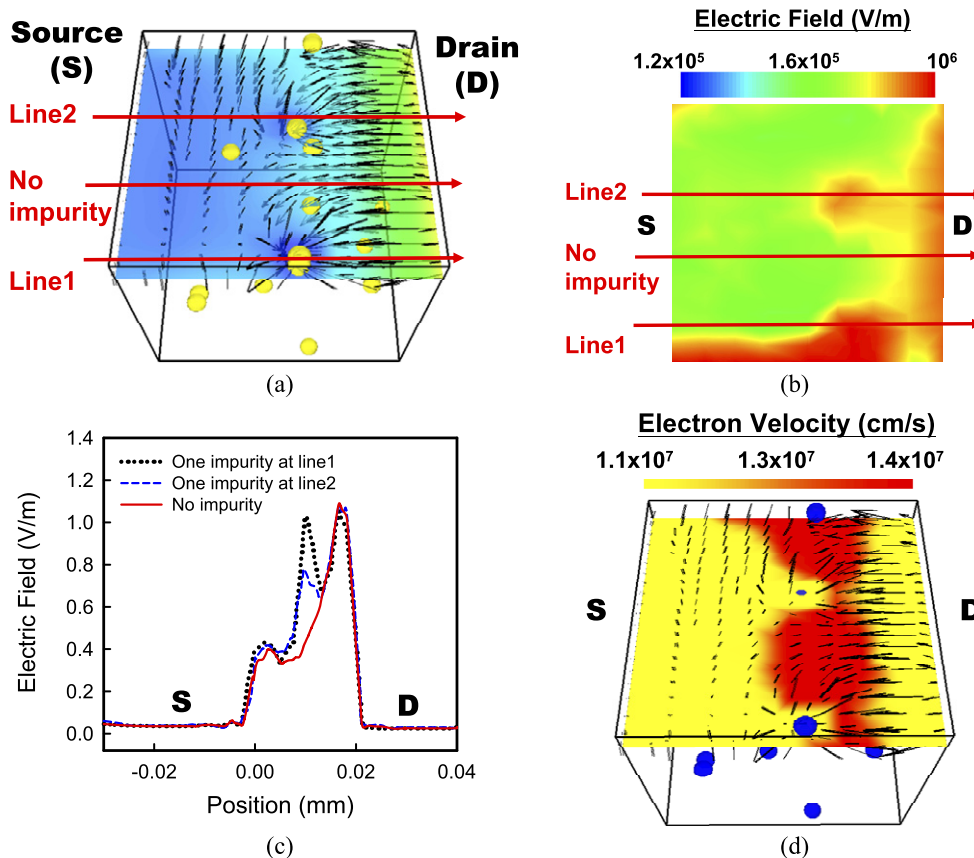


Figure 7. The distributions (a) potential and (b) electric field of the channel for a discrete-dopant-fluctuated transistor. (c) 1D electric field distribution for three cut lines at channel surface. Line 1 is with one dopant located at channel surface; line 2 is without impurity at the conducting path; line 3 also has one dopant located at channel surface, but the dopant is located 0.5 nm below the channel surface. (d) The distribution of electron velocity, where the velocity is enhanced by electric field near dopants.

in nanoscale FinFETs have been identified. The RDF and WKF are the dominating factors in σV_{th} for n-type and p-type FinFETs. The large deviation of workfunction difference in PMOS transistors has impacted the device threshold voltage and therefore should be noticed in the nanoscale transistors era. For the device AC characteristics, the influence of the emerging fluctuation source, WKF, is reduced due to the screening effect of the inversion layer. The screening effect is less effective in RDF because the discrete dopant positioned in the channel may break the shielding and induce random scattering events. The RDF therefore dominates the σF_T . The result of this study is important in understanding the intrinsic parameter fluctuation in DC and AC characteristics, which is useful for nanodevice design in different applications. Currently we are working on the modeling of WKF and developing a fluctuation suppression approach for WKF, such as reduction of grain size.

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