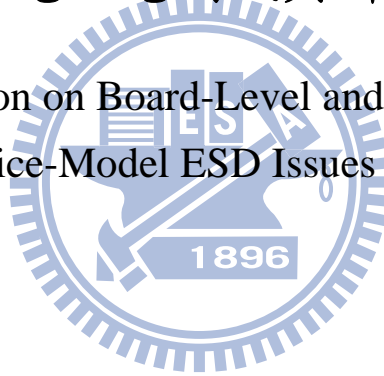# 國 立 交 通 大 學

電機學院　電子與光電學程
碩 士 論 文

積體電路產品之元件充電模式
靜電放電測試與研究

Investigation on Board-Level and Chip-Level

Charged-Device-Model ESD Issues in IC Products

研 究 生：黃志國

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中 華 民 國 九 十 八 年 十 二 月

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摘要

隨著近年來半導體積體電路製程的持續演進，元件尺寸越做越小，已朝向奈米範疇發展，並且由於積體電路運算速度的發展需求，互補式金氧半導體(CMOS)製程技術持續進展，並降低電晶體的閘極氧化層厚度，以提升電路工作頻率；但對靜電放電(Electrostatic Discharge, ESD)而言，較薄的閘極氧化層厚度，意味著電晶體閘極更容易遭受靜電放電轟擊而毀損。在先進互補式金氧半(CMOS)製程中，這種電晶體閘極毀損的情況，在晶片層級元件充電模式(Chip-Level Charged-Device-Model, CDM)靜電放電測試時極為明顯。然而在電子產品應用中，當晶片黏貼至電路板上時，若電路板本身因為摩擦或感應而累積電荷，電路板上累積的電荷，會經由電荷重新分配的過程傳遞至晶片中，並瞬間產生非常大的電流流入晶片，造成晶片損壞，此為電路板層級元件充電模式(Board-Level CDM)靜電放電的成因。

本篇論文的第一部份在探討電路板層級(Board-Level)及晶片層級元件充電模式(Chip-Level CDM)靜電放電在積體電路產品上的行為特性研究與所造成的威脅，且針對實際實驗案例，以故障分析(Failure Analysis)的手法，進行對遭受晶片層級元件充電模式靜電放電測試損害之元件作故障定位(Fault Isolation)，並找出元件充電模式靜電放電所造成之故障機制(Failure Mechanism)比較及探討。

本篇論文的第二部份是以實驗模擬的方式，對數個以互補式金氧半製程製作的測試元件與測試電路進行電路板層級與晶片層級元件充電模式靜電放電測試，首先由於電路板層級元件充電模式靜電放電的電流峰值與電子模組中的電路板尺寸有密切的關係，故針對不同電路板尺寸所產生的電路板層級元件充電模式靜電放電波形進行量測，實驗結果顯示較大的電路板尺寸或將電路板充電至較高電壓，將導致較大的電路板層級元件充電模式靜電放電電流峰值；其次針對測試元件與測試電路進行電路板層級與晶片層級元件充電模式靜電放電測試，測試結果發現電路板層級元件充電模式靜電放電耐受度較低，且證實電路板層級元件充電模式對積體電路產品所造成的損害，遠比晶片層級元件充電模式來的嚴重。

　　本論文透過實驗的結果，成功證明電路板層級元件充電模式靜電放電在積體電路產品所造成的損害，遠比晶片層級元件充電模式來的嚴重，而且常容易被誤認為過度電性應力(Electrical Over Stress, EOS)，由於目前對電路板層級的靜電放電測試尚未有明確規範，經由本論文實驗步驟的建立，可供日後規範建立的參考。

Investigation on Board-Level and Chip-Level Charged-Device-Model ESD Issues in IC Products

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ABSTRACT

With the continuous evolution of semiconductor integrated circuit (IC) process, the device dimension growing narrow down and developing into nanoscale. Moreover, the transistors have been fabricated with thinner gate oxides to achieve higher speed or operation frequency due to the operation speed requirement of integrated circuits (ICs) in advanced process of complementary metal-oxide semiconductor (CMOS). In electrostatic discharge (ESD) events, the transistors are more easily damaged during ESD stress if they are fabricated with thinner gate oxides. The situation of gate oxide damage of transistors is a typical and familiar failure mechanism during chip-level charged-device-model (CDM) ESD test, especially in CMOS process. But in the applications of microelectronic system, IC chips must to be attached to the printed circuit board (PCB). The static charges will be stored in the PCB due to induction or rub and then deliver the charges to the IC chips through redistribution process during the attachment of IC chips to PCB. The instantaneous current flows into the IC chips is huge and will result in the damage of IC chips. It is the cause of board-level CDM ESD event.

In the first part of this thesis, the focus is the investigation of characteristics and threats on board-level and chip-level CDM ESD in IC products. Furthermore, from the experimental results, the technique of failure analysis (FA) with fault isolation is applied to summarize the comparison of CDM failure mechanism caused by ESD event during CDM ESD test.

The second part presents an experiment of ESD test between board-level and chip-level CDM on several samples fabricated with CMOS process. At first, the board-level CDM ESD current waveforms under different sizes of printed circuit boards (PCBs) and different charged voltage are measured. The experiment result has shown that the discharging current strongly depends on the PCB size and the charged voltage. Moreover, chip-level and board-level CDM ESD levels of several test devices and test circuits have been characterized and compared. Test results have shown that the board-level CDM ESD level of the test circuit is lower than the chip-level CDM ESD level, which demonstrates that the board-level CDM ESD event is more critical than the chip-level CDM ESD event. In addition, failure analysis reveals that the failure on the test circuit under the board-level CDM ESD test is much severer than that under the chip-level CDM ESD test.

Based on the experiment results in this thesis, it is successfully proved that the failures caused by board-level CDM ESD event are more server than chip-level CDM ESD event and are easily mistaken for electrical over stress (EOS) related failures. Since the standard for the board-level CDM ESD test is not established so far, the experiment procedures in this thesis can be the reference for the establishment of the board-level CDM ESD test standard.

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CHAPTER 1

INTRODUCTION

1.1 Background of ESD and Electrical Over Stress (EOS)

Electrical Over Stress (EOS) occurs when a voltage greater than the maximum specified voltage is applied to any portion of an electrical device. This excess voltage will often cause excess current to flow along some path. If those current remains too long, heating of the device will occur, very often this heating will result in permanent damage. On an integrated circuit device this damage commonly manifests itself as fused metal lines, melted silicon or other burned circuit elements and is frequently accompanied by carbonized package epoxy deposited on the die surface after decapsulation [1]. On an intrinsic level, EOS testing involves putting increasingly higher voltages across a device until failure occurs. Aside from device geometry, the only variable other than voltage and current that is needed to determine device heating is the duration of the applied overstress (pulse). With smaller voltages, the duration is relatively long (mS to S) but as voltages increase and pulse widths narrow we enter the domain of ESD (Electro Static Discharge).

An ESD (Electro Static Discharge) event occurs when a static electric charge builds up on an object (a person, tool, bench, and device) and is then subsequently discharged to ground. Often a very large potential is built up and the actual discharge generates a spark, the discharge event is commonly referred to as a "Zap". ESD events become a major reliability problem when

this zap discharges through an electronic device. Semiconductor devices, such as CMOS IC's and other computer components are particularly sensitive to damage. A real ESD zap can have greatly variable characteristics. A pulse will typically have a very fast rise time and short duration but voltage and current levels will vary significantly from one event to another.

Fig. 1.1 shows the comparison of the pulse duration generated by ESD and EOS. As it can be seen, the pulse duration (nS to μS) of ESD is shorter than that of EOS (mS to S). The other Key difference between ESD and EOS is the rise time of energy pulse. The typical value of rise time for ESD and EOS is 5 to 20nS and 5μS to 5mS, respectively. Both of them will damage devices by a rapid localized heating of the semiconductor material or by rapidly creating strong electrical fields.
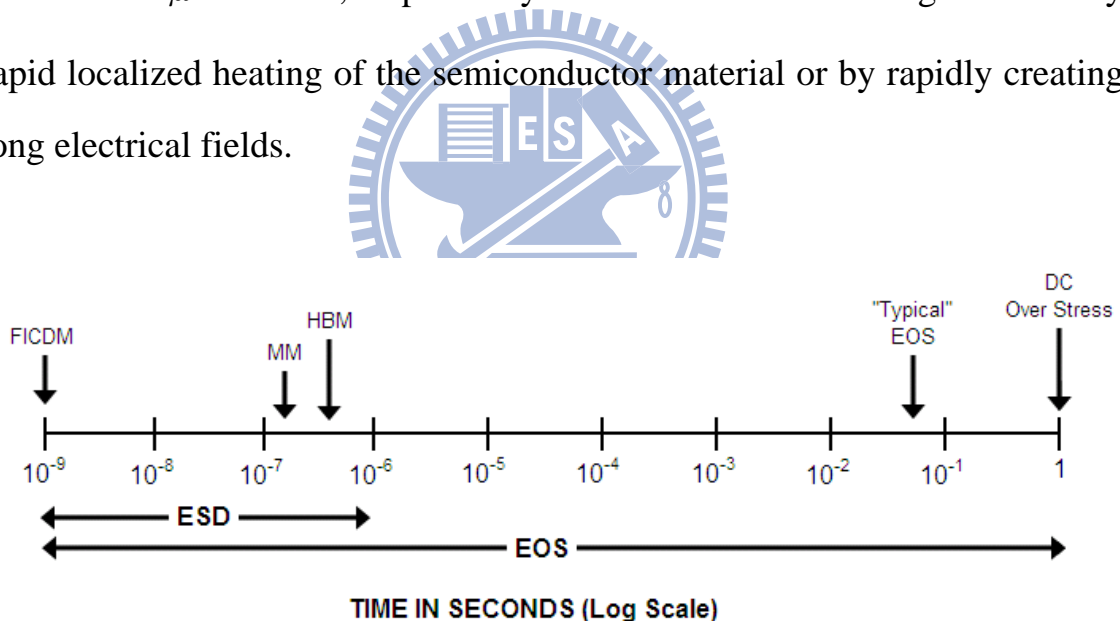
Fig. 1.1   Comparison of pulse duration on ESD and EOS.

Table 1.1 also shows the comparison of pulse duration and failure signature of ESD and EOS. The failure mechanisms of EOS are mostly observed as fused metal lines, melted silicon or other burned circuit elements and are frequently accompanied by carbonized package epoxy deposited on the die surface after decapsulation. These damages can be easily visible

through an OM (Optical Microscope). The ESD failure signatures are most likely observed with junction degradation, contact damage, poly melt filament and gate oxide breaches. These damages may be only visible through delayer and SEM inspection.

Table 1.1 Comparison of ESD and EOS.

ESD vs. EOS	
ESD (ElectroStatic Discharge)	**EOS (Electrical Over Stress)**
ESD	Over Spec.
Rapid Transient	Slow Transient
Hard & Soft Failure	Hard Failure
Not Necessary Obvious Sub-spec. Leakage => Latent Failure	Obvious (Burn out)

1.2   ESD Testing for HBM, MM and CDM

An ESD pulse can be caused by several physical factors, each leading to distinct characteristics depending on the source. These characteristic pulses are broadly grouped into three categories: Human Body Model (HBM), Machine Model (MM), and Charged-Device-Model (CDM). The equivalent circuits of HBM, MM, and CDM ESD tests are shown in Fig. 1.2 and Fig. 1.3, respectively. In the HBM standard, the circuit component to simulate the charged human being is a 100pF capacitor in series with a 1500 ohm resistor [2]. This network has a characteristic rise time and decay time. The characteristic decay time is associated with the time of the network. An MM characteristic time is associated with the electrical components used to emulate the discharge process. In the MM standard, the circuit component is a 200 pF capacitor with no resistive component [3].

(a)

(b)
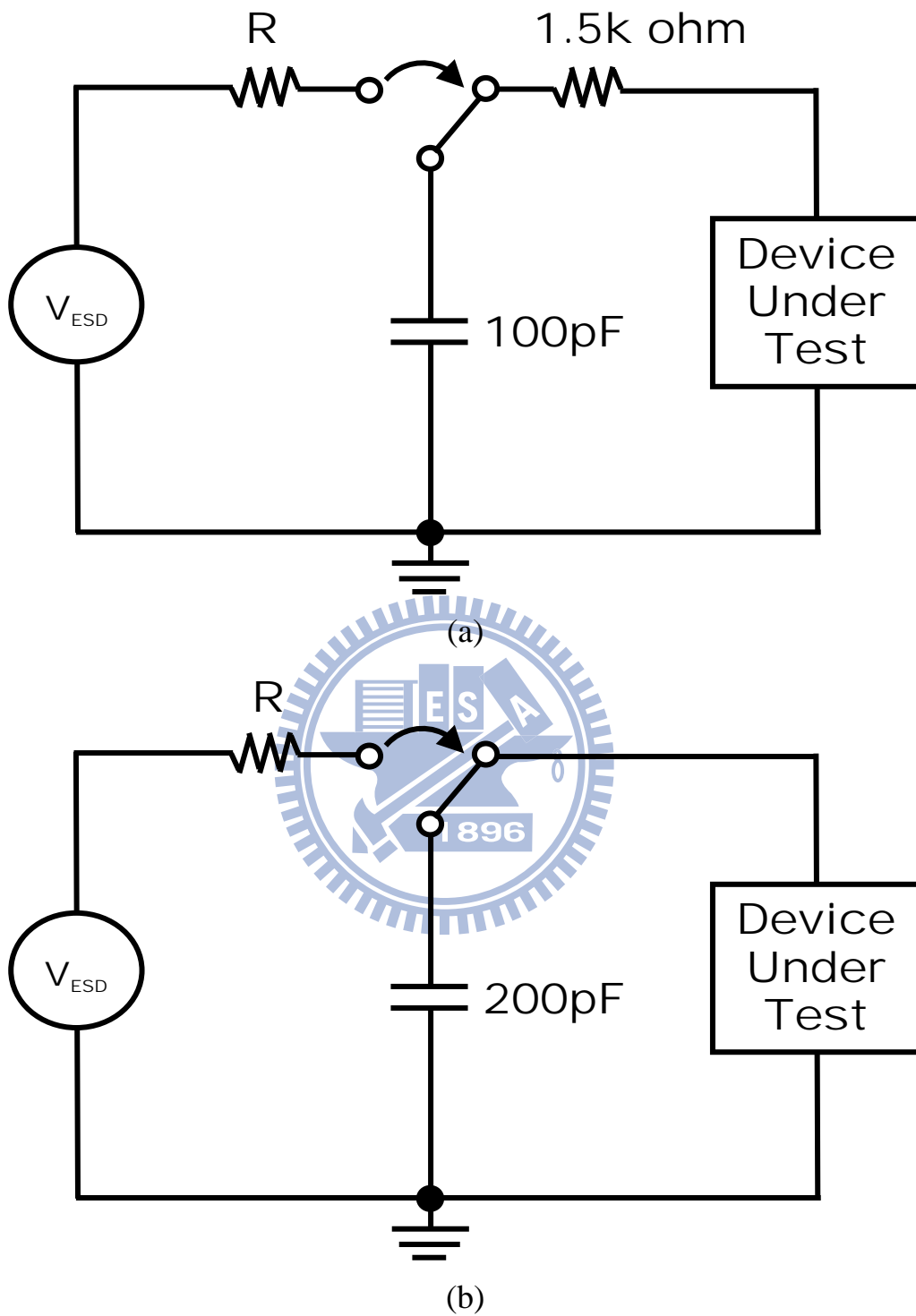
Fig. 1.2   Equivalent circuits of (a) HBM, and (b) MM, ESD tests.

The charged-device-model (CDM) represents IC self-charging (field or triboelectric) and then self-discharge. Currently there are two methods, one supporting the field-induced method and the other supporting a socketed method ESD association standard [4]. The field-induced method is more realistic but less repeatable, whereas the socketed method is less realistic but more repeatable.
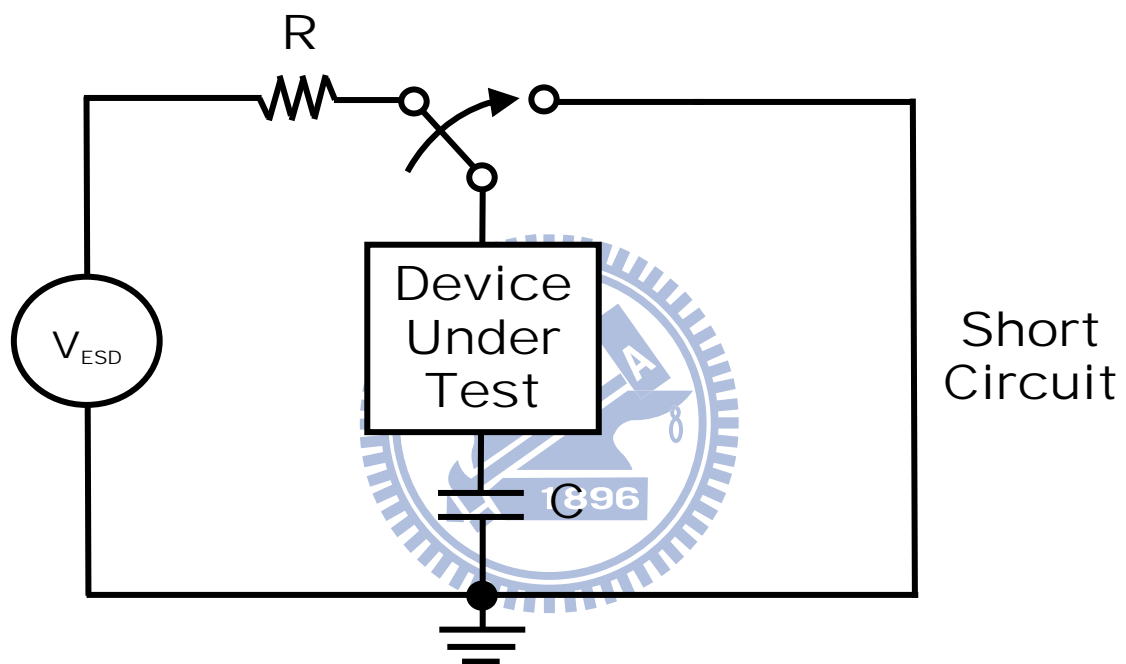
Fig. 1.3   Equivalent circuits of CDM ESD test.

According to the standard test method of HBM and MM ESD tests, the test pin orientations are specified with three kinds of combination as Fig. 1.4 to Fig. 1.6 shown. The stresses may have positive or negative voltages on an I/O pin with respect to the grounded VDD or VSS pin. For comprehensive ESD verification, the pin-to-pin ESD stresses and VDD-to-VSS stresses had also been specified to verify the whole-chip ESD robustness.
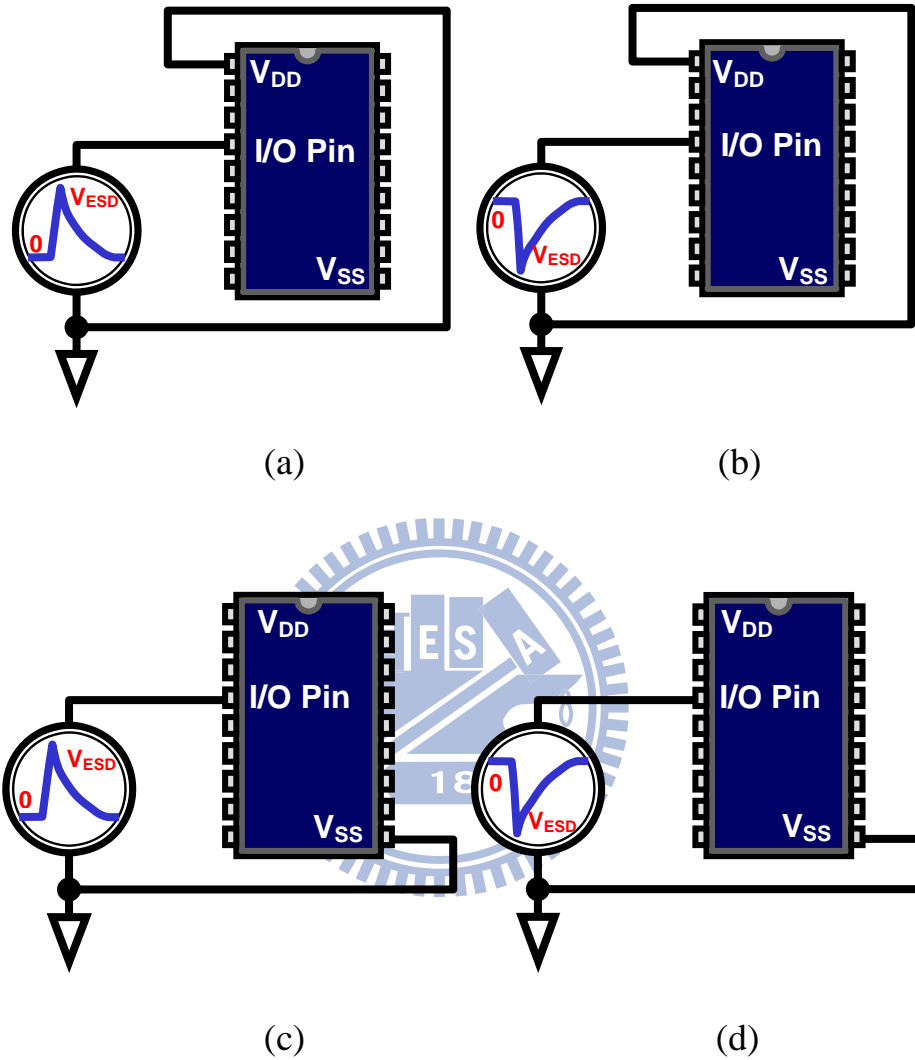
Fig. 1.4 Four ESD test pin combinations for IC products: (a) positive-to-VDD mode (PD Mode), (b) negative-to-VDD mode (ND Mode), (c) positive-to-VSS mode (PS Mode), and (d) negative-to-VSS mode (NS Mode).

(a)   (b)
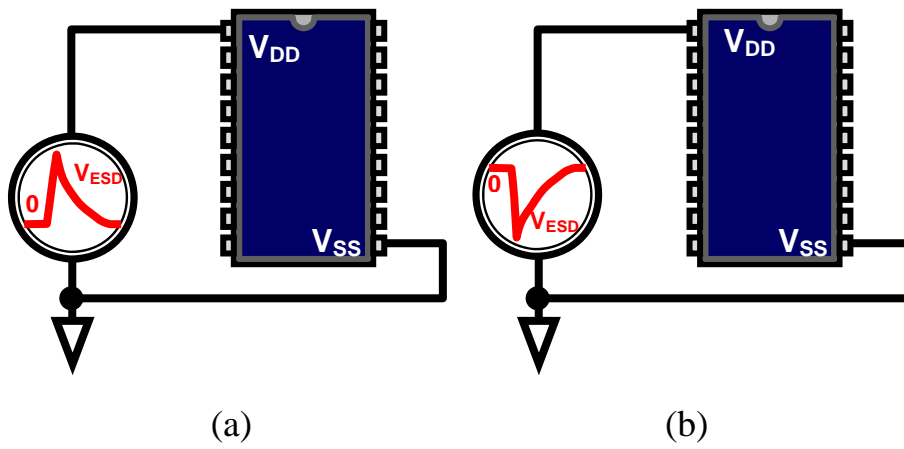
Fig. 1.5   Pin combinations in VDD-to-VSS ESD test: (a) positive mode, and (b) negative mode.

(a)   (b)
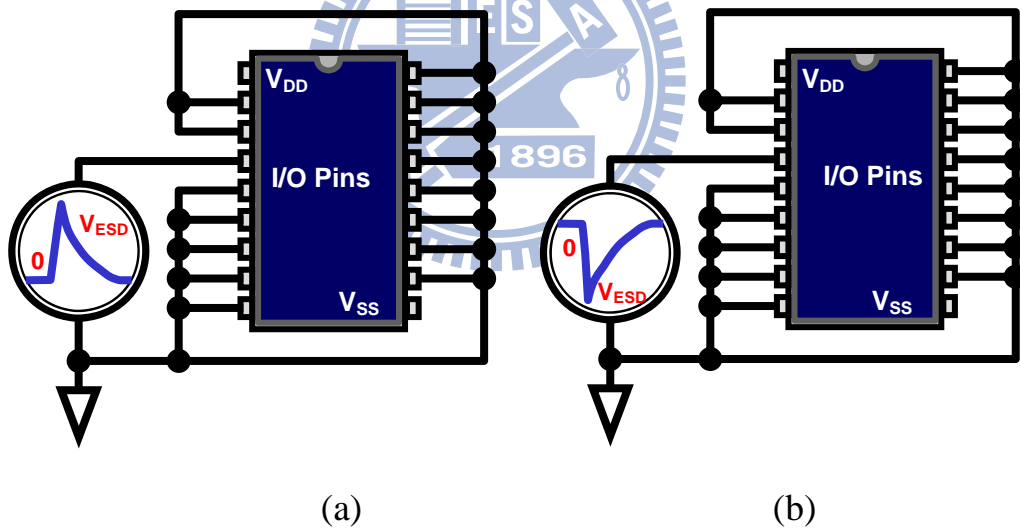
* All power pins are floated.

Fig. 1.6   Pin combinations in pin-to-pin ESD test: (a) positive mode, and (b) negative mode.

Fig. 1.7 and Fig. 1.8 show the CDM ESD test methods of socketed and non-socketed CDM, respectively. In socketed CDM ESD test, the device under test (DUT), socket, test fixture board, HV relays, and other parts of the test simulator are charged and discharged during the test. The charge is uniquely stored in a distributive network of parasitic capacitance and inductance elements starting from the high voltage supply, the high voltage ground relays, the pogo pins, the test fixture board, the socket, and the DUT. Consequently, the discharge currents through the pin under test represent the charge stored in the IC device and socketed CDM test simulator's distributive network.
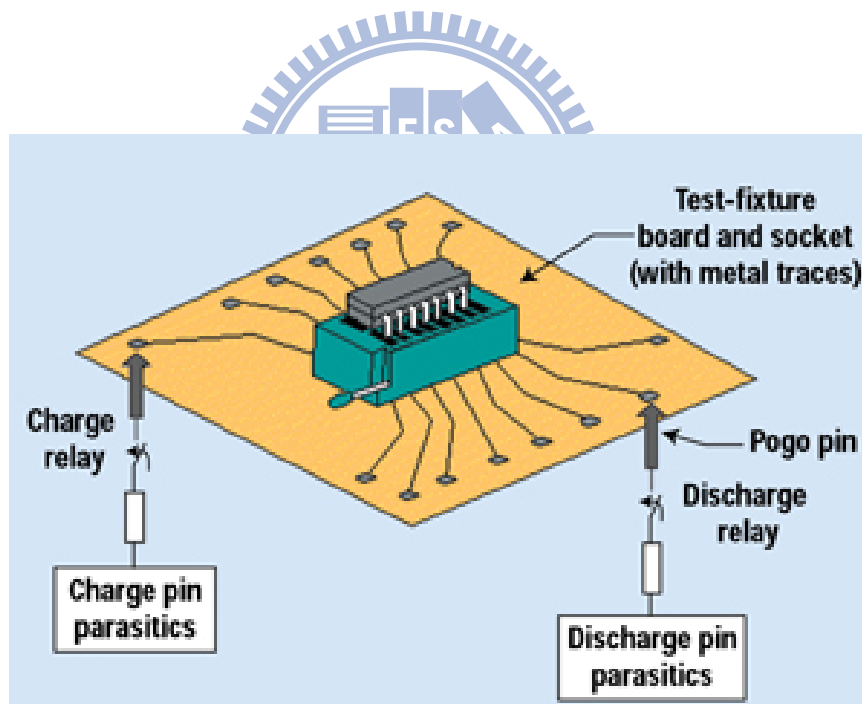
Fig. 1.7   Architecture of socketed CDM (SDM) ESD test.
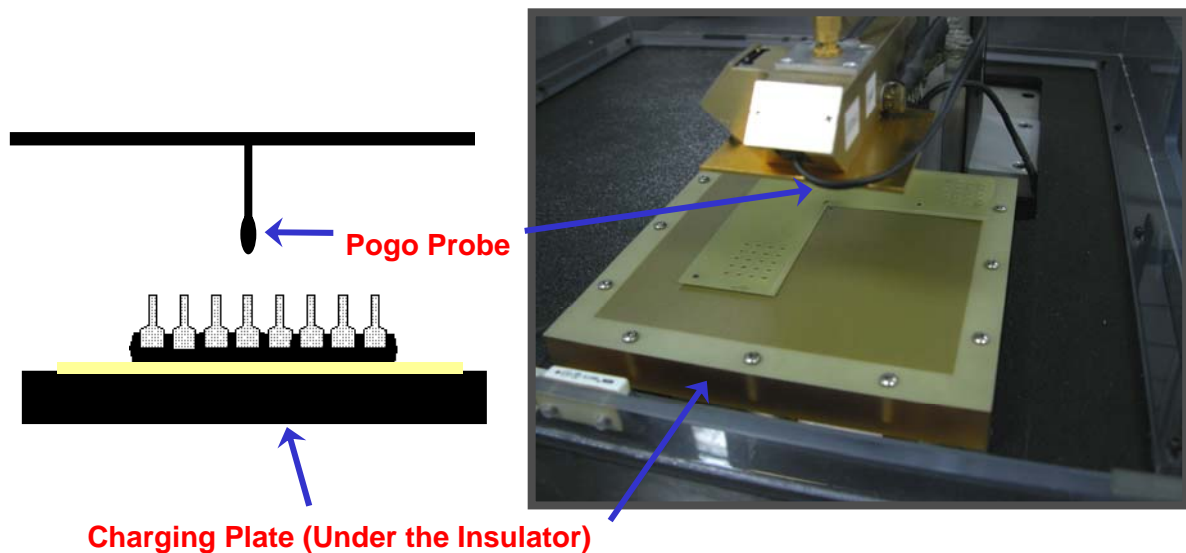
Fig. 1.8   Non-socketed CDM test equipment (Oyrx CDM Orion).

In non-socketed CDM ESD test, two methods may be used to raise the component potential for the subsequent CDM discharge. One is direct-charging method and the other is filed-induced method. Since the field-induced method is more realistic than direct-charging method, the CDM related experiments in this thesis are tested by field-induced method/equipment which is shown in Fig. 1.8. With the CDM ESD test, the tested component was placed on the charging plate. Then raise the potential of the component by energizing the field charging plate. Discharge through all pins, including VDD and VSS pins, one at a time. The CDM discharge is generally completed in a few nanoseconds, and peak currents of tens of amperes have been observed. Fig. 1.9 shows the comparison of characteristic HBM, MM, and CDM pulses, illustrating that three models impose different requirements and as such, three models are mandatory to cover all ESD-induced device failures. The CDM stress clearly causes much faster and higher amplitude discharge current.
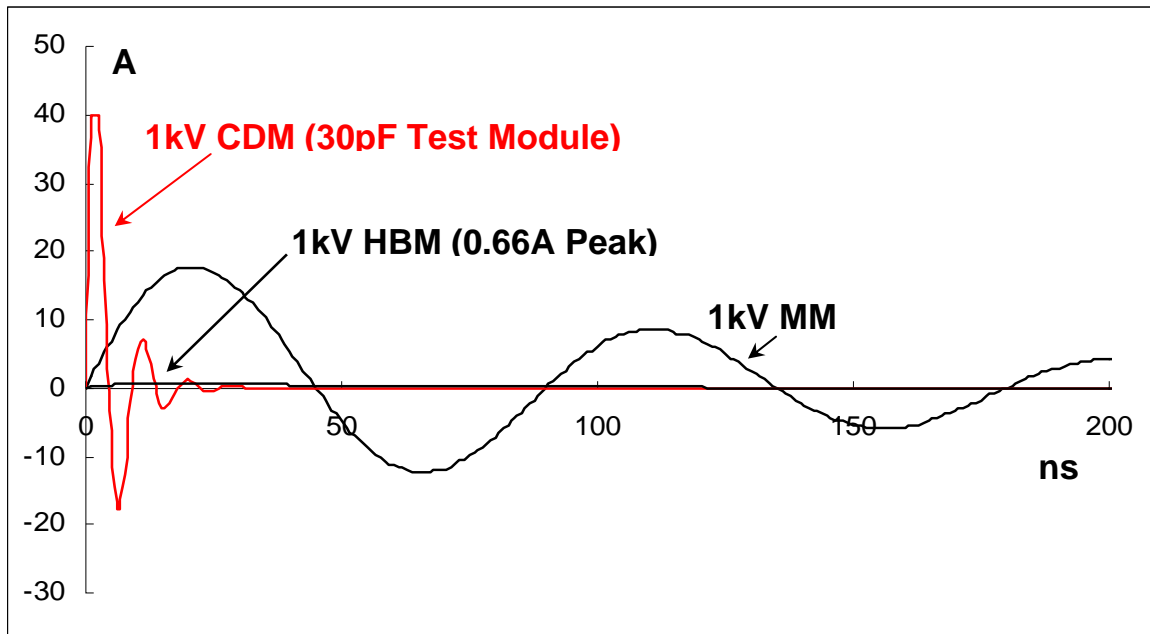
Fig. 1.9   Discharge current waveforms of three ESD stress models (HBM, MM, and CDM).

With progress of automation in manufacturing lines, the HBM and MM damages would decrease and major failures would be explained by the CDM. Furthermore, as the dielectric failure has gained importance with scaled semiconductor devices [5]–[7], the CDM accounts for the majority of ESD failures during chip manufacturing. Table 1.2 is the summary of the ESD zap numbers including the polarity and the minimum zap interval according to the standards, MIL, ESDA, JEDEC, and AEC.

Table 1.2 The summary of the ESD zap numbers including the polarity and the minimum zap interval according to the standards, MIL, ESDA, JEDEC, and AEC.

ESD Model		MIL-STD	ESDA	EIA/JEDEC	AEC
HBM	Standard	MIL-STD-883G Method 3015.7	ESDA STM5.1-2001	JESD22-A114E	AEC-Q100-002-RevD
	Zap Interval [minimum]	1 s	300 mS	100 mS	500 mS
	zap number	3(+), 3(-)	1(+), 1(-)	1(+), 1(-)	1(+), 1(-) or 3(+), 3(-)
MM	Standard		ESDA STM5.2-1999	JESD22-A115A	AEC-Q100-003 RevE
	Zap Interval [minimum]		1 s	500 mS	1 s
	zap number		3(+), 3(-)	1(+), 1(-)	3(+), 3(-)
Non-socketed CDM	Standard		ESDA STM5.3.1-1999	JESD22-C101C	AEC-Q100-011 RevB
	Zap Interval [minimum]		1 s	200 mS	1 s
	zap number		3(+), 3(-)	3(+), 3(-)	3(+), 3(-)
Socketed CDM	Standard		ESD DSP5.3.2-2003 draft		
	Zap Interval [minimum]		1 s		
	zap number		3(+), 3(-)		

1.3 Diagnosis of Possible Damage Sites on ESD Protection Circuit

With the purpose of considering the possible ESD damage sites, the ESD discharging current paths need to be created and constructed. Fig. 1.10 shows the typical on-chip double-diode whole-chip ESD protection scheme in which two ESD diodes at I/O pad are co-designed with the power-rail ESD clamp circuit to prevent internal circuits from ESD damage [8]. In Fig. 1.10, a P+/N-well diode (D_P) and an N+/P-well diode or an N-well/P-substrate diode (D_N) are placed at input pad or output pad. When the D_P and D_N are under forward-biased condition, they can provide discharge paths from I/O pad to VDD and from VSS to I/O pad, respectively.
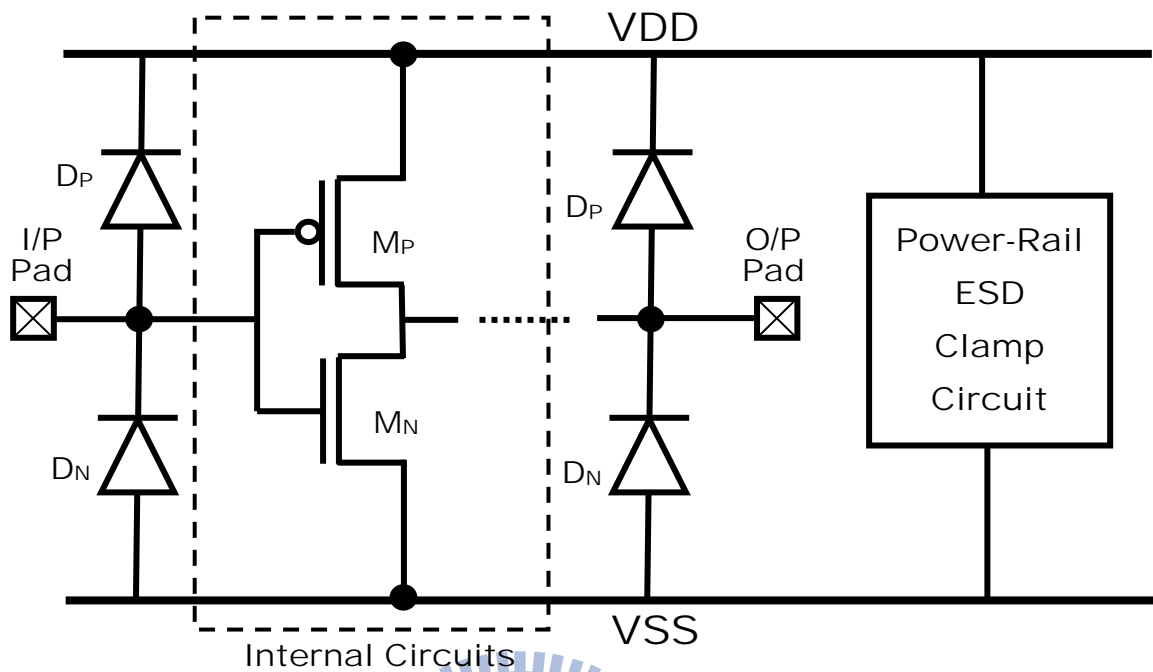
Fig. 1.10 Typical double-diode whole-chip ESD protection scheme for ICs.

During HBM or MM ESD stress with positive-to-VDD (PD) mode and negative-to-VSS (NS) mode, ESD current is discharged through the forward-biased D_P and D_N, respectively. To avoid the ESD diodes from being operated under breakdown condition during HBM or MM ESD stress with positive-to-VDD (PD) mode and negative-to-VSS (NS) mode, which results in a substantially lower ESD robustness, the power-rail ESD clamp circuit is used between VDD and VSS to provide ESD current paths between the power rails [9]. Thus, ESD current is discharged from the I/O pad through the forward-biased D_P to VDD, and discharged to the grounded VSS pin through the turn-on efficient power-rail ESD clamp circuit during PS-mode ESD stress, as shown in Fig. 1.11.
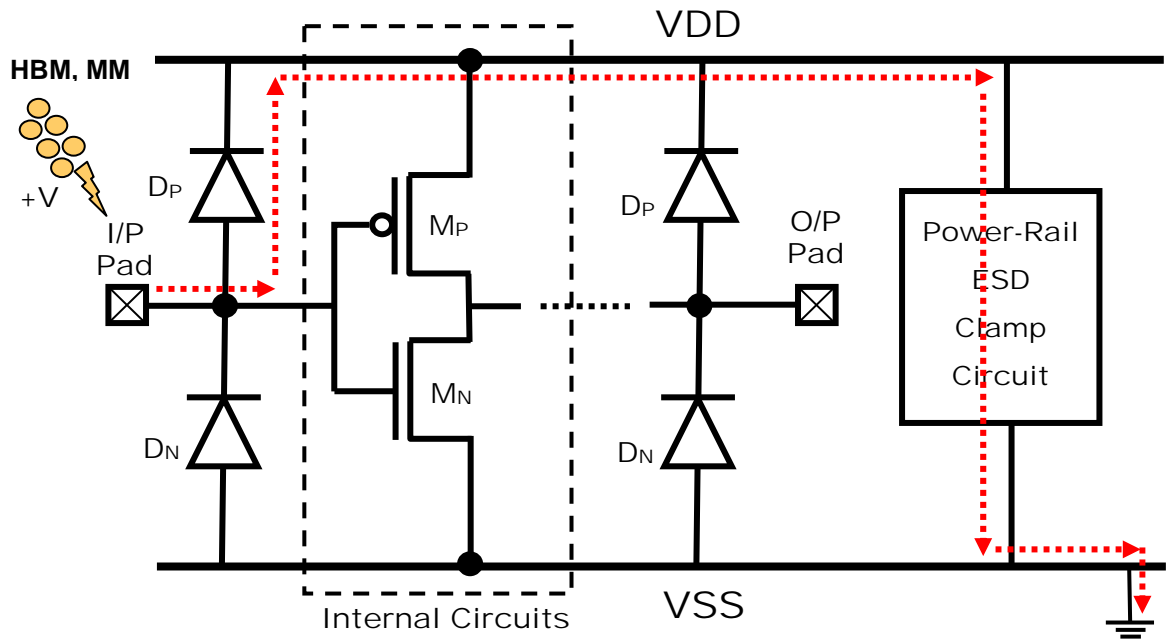
Fig. 1.11 ESD current path in the typical double-diode whole-chip ESD protection scheme under PS-mode ESD stress.

Similarly, ESD current is discharged from the VDD pin through the turn-on efficient power-rail ESD clamp circuit and the forward-biased D_N to the I/O pad during ND-mode ESD stress, as shown in Fig. 1.12.
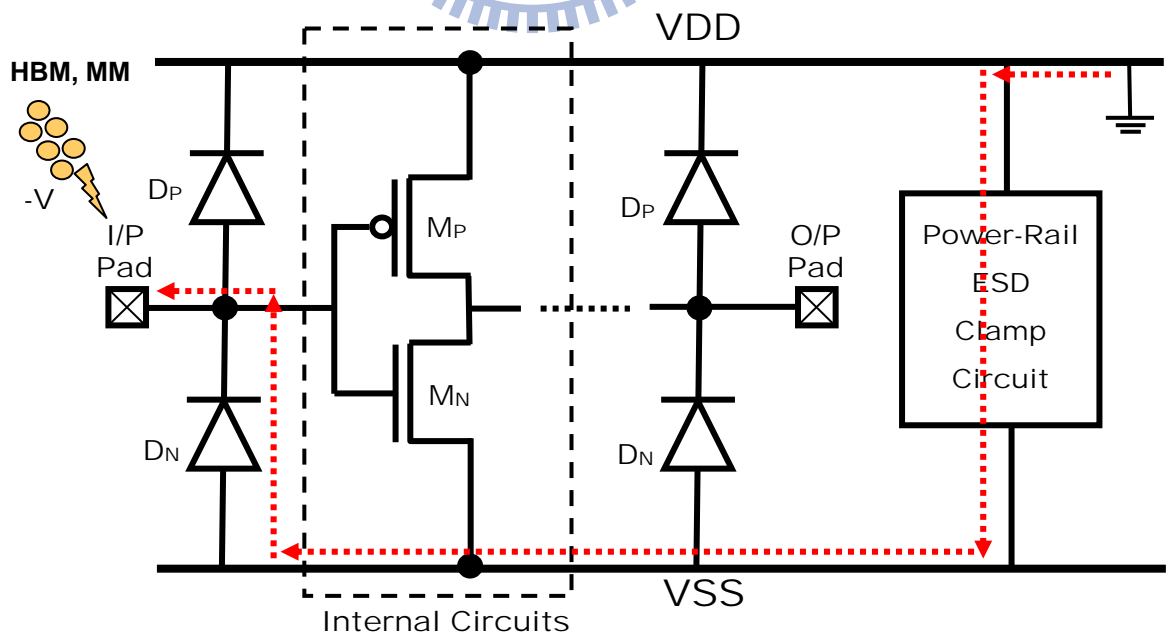
Fig. 1.12 ESD current path in the typical double-diode whole-chip ESD protection scheme under ND-mode ESD stress.

During pin-to-pin ESD stress, ESD current flows from the zapped I/O pad through the forward-biased D_P, the power-rail ESD clamp circuit, and the forward-biased D_N to the grounded I/O pad, as shown in Fig. 1.13.
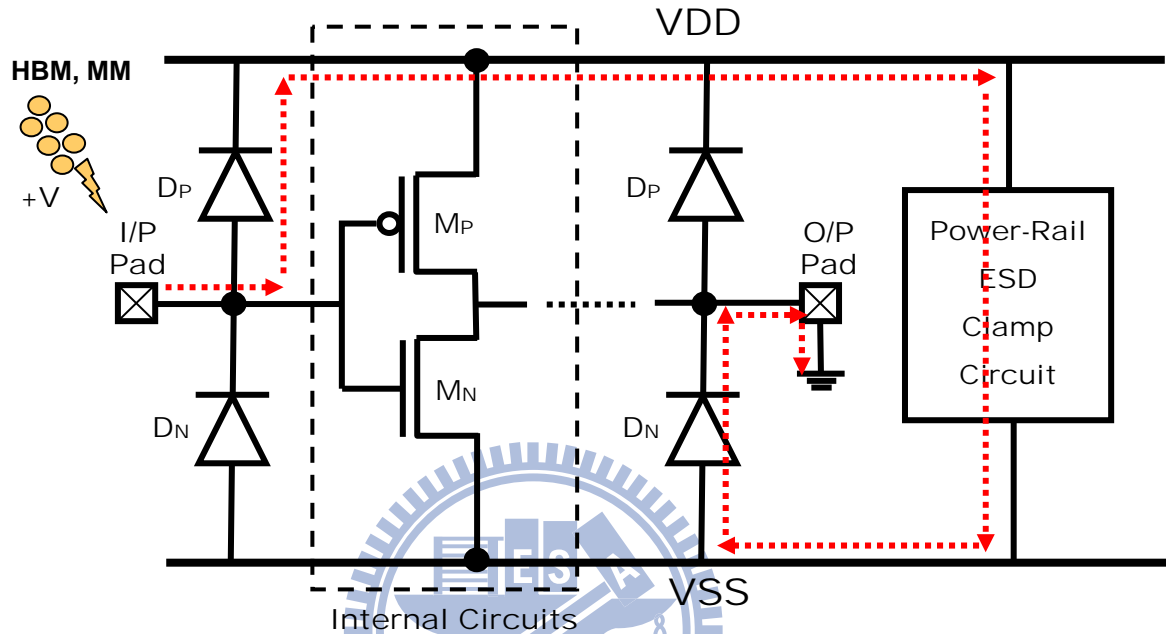
Fig. 1.13   ESD current path in the typical double-diode whole-chip ESD protection scheme under pin-to-pin ESD stress.

Under VDD-to-VSS ESD test, ESD current flows through the power-rail ESD clamp circuit between VDD and VSS. During CDM ESD test, the charges are stored within the body of IC, ESD discharge current flows from the body of IC through M_N to the grounded I/O pad. The gate oxide of M_N would be damaged due to its thinner thickness than that of ESD protection devices. In summary, as Fig. 1.14 shown, the possible ESD damage sites with HBM and MM ESD stresses are most likely on the common drain of double-diode if they are not robust to discharge the ESD current. A typical example is layout or process issue cause non-uniform turn-on of ESD device [10].
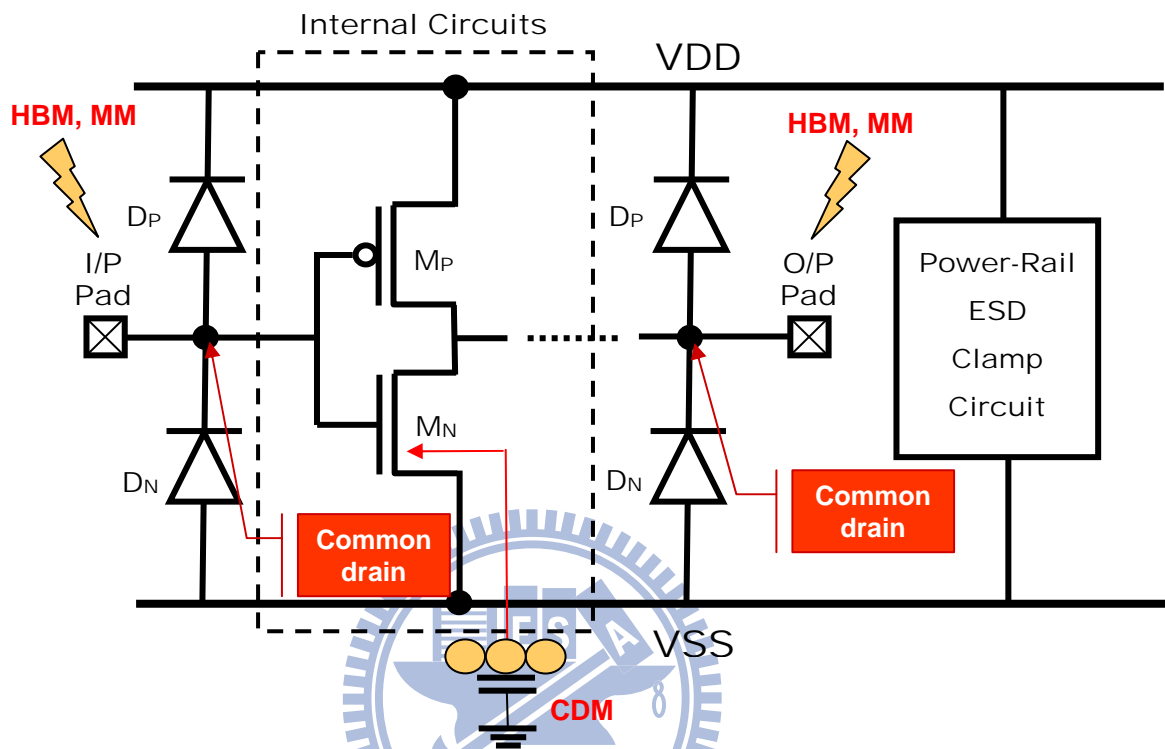
Fig. 1.14 Possible damage locations on the ESD devices under HBM, MM, and CDM ESD tests in whole-chip ESD protection scheme for ICs.

Fig. 1.15 shows the possible ESD damage sites and paths by physical structure illustration of ESD protection device. There are typically six kinds of damage site and mechanism as indicated in the figure. Fig. 1.16 shows the cross-sectional view of ESD device with possible damage sites and paths by transmission electron microscopy (TEM) inspection.
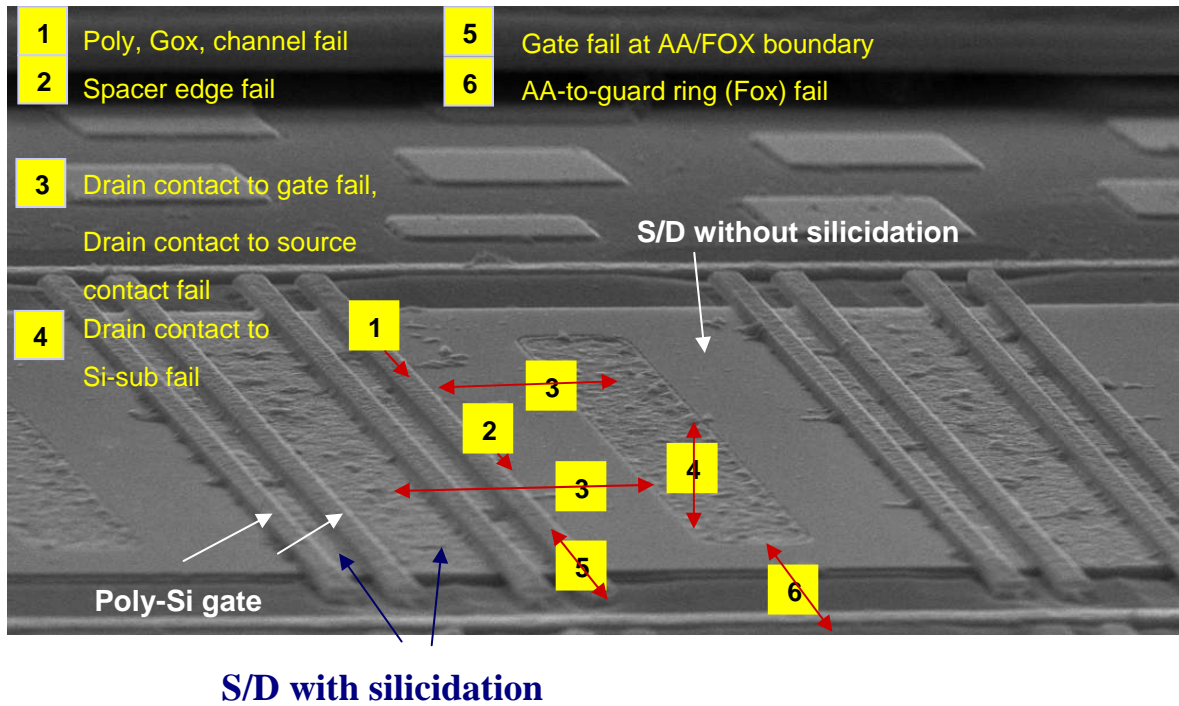
Fig. 1.15 Possible ESD damage sites/paths and mechanisms of the ESD device by physical structure illustration (plane-view of SEM).
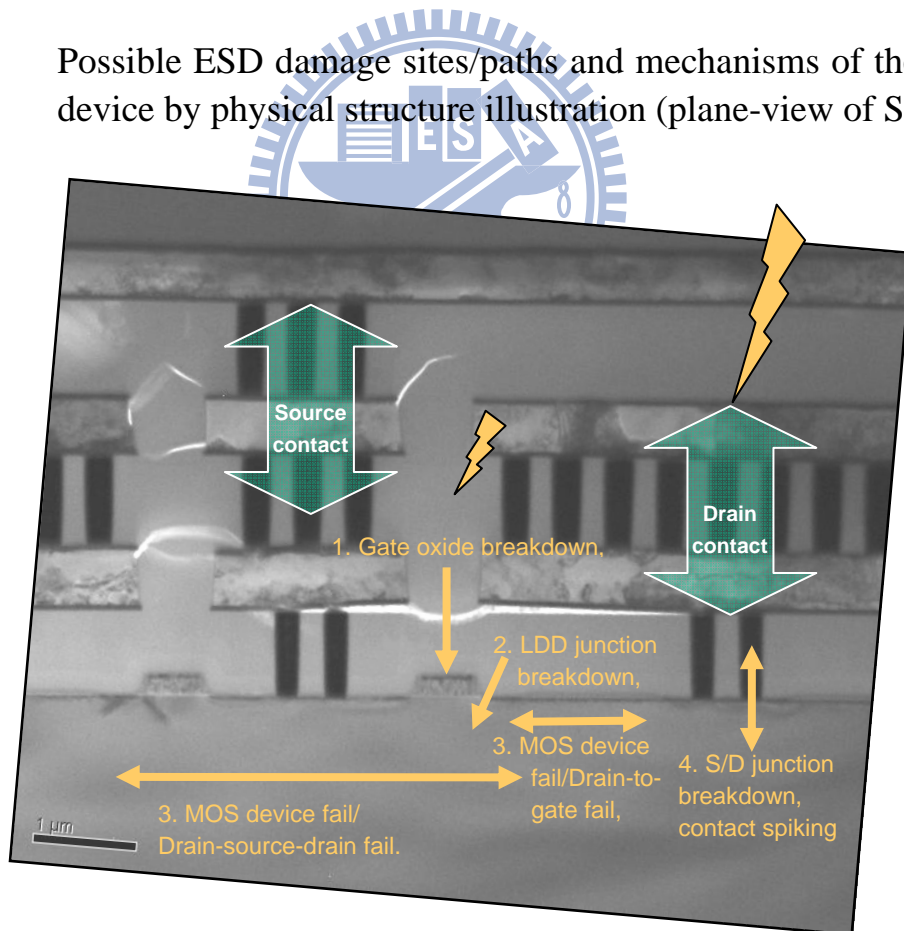
Fig. 1.16 Possible ESD damage sites/paths and mechanisms of the ESD device by physical structure illustration (cross-sectional view of TEM).

During HBM or MM ESD stress, the ESD current flows from the common drain contact through junction and channel to source contact. In this current path, the possible failure mechanisms would be poly damage, GOX damage, D/S punch-through, and drain contact spiking due to the high ESD current and energy. In CDM ESD test, the current path is different from HBM and MM ESD stresses. The typical failure mechanism of CDM fail is GOX damage of the device in internal circuits which is the closest to grounded I/O pads. Table 1.3 lists the summary of ESD failure location with respect to failure mechanism.

Table. 1.3 Summary of ESD failure location with respect to failure mechanism.

Failure Location	Failure Mechanism
Poly, GOX, channel	gate oxide breakdown
Spacer edge	LDD junction breakdown
Contact to gate , S-to-D	MOS device fail
Contact to Si-sub	S/D junction breakdown, contact spiking
Gate, AA/FOX edge	High electric field breakdown
AA-to-guard ring (FOX), STI sidewall	Latch-up fail

1.4 Board-Level Charged-Device-Model (CDM) ESD Issue

In the CDM ESD test, the IC chip is charged first, and then the IC is discharged through the tested pin. During the chip-level CDM ESD test, the charges stored in the substrate or package of the IC chip is suddenly discharged to ground, which leads to huge discharging current flowing

through the test pin. Therefore, CDM becomes more critical among the three component-level ESD test standards because of the thinner gate oxide in nanoscale CMOS devices and the larger die size for the application of system on chip (SoC). The thinner gate oxide causes a lower GOX breakdown voltage, which makes the MOS transistor more sensitive to ESD. Moreover, an IC with larger die size can store more static charges, which leads to larger discharging current during CDM ESD events. CDM ESD current has the features of huge peak current and short duration. Furthermore, CDM ESD current flows from the chip substrate to the external ground, whereas HBM and MM ESD currents are injected from the external ESD source into the zapped pin. Thus, effective ESD protection design against CDM ESD stress has gotten more requests in IC industry.

In addition to chip-level CDM ESD issue, board-level CDM ESD issue becomes more important recently, because it often causes the ICs to be damaged after the IC is installed to the circuit board of electronic system. For example, board-level CDM ESD events often occur during the assembly of microelectronic modules or module function test on the circuit board of electronic system. Even though the IC has been designed with good chip-level ESD robustness, it would still be very weak in board-level CDM ESD test. The reason is that the discharging current during the board-level CDM ESD event is significantly larger than that of the chip-level CDM ESD events on real IC products [11], [12]. In these two previous works, the ICs which already passed the component-level ESD specifications were still returned by customers because of ESD failure. After performing the field-induced CDM ESD test on the ICs which have been mounted on the printed circuit board (PCB), the failure is the same as that happened in the customer returned ICs.

This indicates that the real-world charged-board-model (CBM) ESD damage can be duplicated by the board-level CDM ESD test. These previous works have demonstrated that the board-level CDM ESD events indeed exist, which should be taken into consideration for all IC products.

1.5   Organization of This Thesis

To investigate the failure behaviors and comparisons of CDM ESD issues for board-level and chip-level CDM ESD events, this thesis consists of four chapters. In chapter 2, chip-level and board-level CDM ESD issues in IC products are investigated and several case studies with FA methodology and fault isolation on chip-level CDM ESD failure are also included.

In chapter 3, based on the mechanism of board-level CDM ESD event, an experiment has been performed to investigate the board-level CDM ESD current waveforms under different sizes of PCBs and charged voltages in the discharging path. Experimental results have shown that the discharging current strongly depends on the PCB size and the charged voltage. Moreover, chip-level and board-level CDM ESD levels of several test devices and test circuits fabricated in CMOS processes have been characterized and compared. Test results have confirmed that the board-level CDM ESD level of the test circuit is lower than the chip-level CDM ESD test.

Finally, chapter 4 summarizes the main results of this thesis and future works are also addressed in this chapter.

CHAPTER 2

FAILURE ANALYSIS OF CDM ESD EVENT

2.1 Chip-Level CDM ESD Event in IC Products

A chip-level CDM ESD event is a charge driven phenomenon [13]. An IC can become charged by triboelectrification and stored within the body of itself. If this charged device come into contact with a metallic surface through a certain pin, the charge initially distributed all over the conductive parts of the device is collected and leaves it via the grounded pin within a few nanoseconds. This process is called as CDM ESD event and shown in Fig. 2.1.
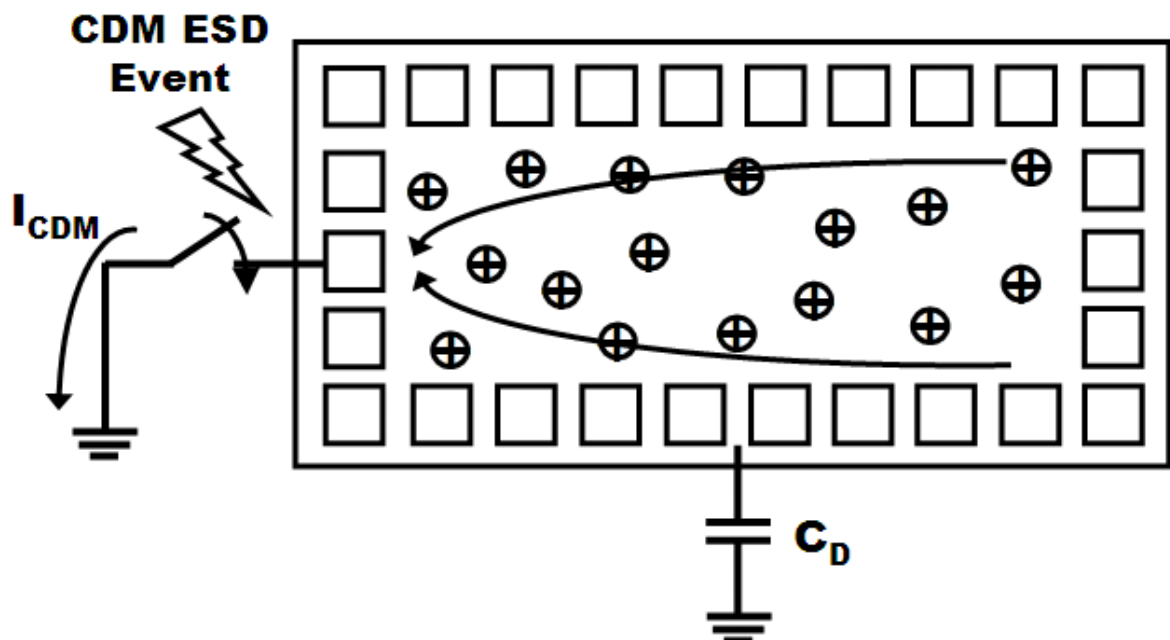
Fig. 2.1   CDM ESD event: When a certain pin is grounded, the stored charges in the IC will be quickly discharged through the grounded pin.

Parasitic capacitance (C_D) is the equivalent capacitance with respect to the die size of different ICs. Thus, different ICs have different peak currents and different CDM ESD levels. Since the discharging current in CDM ESD event is larger and faster than that of HBM and MM ESD events, the internal circuits may be damaged during CDM ESD events before the ESD protection circuit is turn-on. Thus, CDM ESD current is most likely to flow through the gate oxides of MOS transistors due to their capacitive structures in high signal frequency. So the gate oxide is most likely to be damaged during CDM ESD events. This kind of failure mechanism is caused by high current densities and high electric fields [14]. While current-induced damage occurs because joule heating melts a region of the structure, there are two failure modes associated with high electric fields. One is dielectric rupture and the other is charge injection. Dielectric rupture is the case where an induced voltage creates an electric field greater than the dielectric strength of the material. Charge injection occurs because high electric fields at the surface of a junction accelerate the electrons, gaining enough energy to surmount the oxide–silicon energy barrier [15], [16].

In nanoscale CMOS processes, the gate oxide thickness becomes thinner and derives from the increase of parasitic capacitance (C_D). Therefore, the gate oxides of MOS transistors in nanoscale CMOS processes are more vulnerable to CDM ESD stress. Furthermore, more functions and systems are integrated into a single chip such as SoC, will increase the die area/size and the die capacitance. It would cause more static charges stored in IC and induce larger CDM ESD current. With larger die size and MOS transistors using thinner gate oxide, nanoscale CMOS ICs are very sensitive to ESD, especially CDM ESD events.

2.2 Fault Isolation Techniques on CDM ESD Failure

In ESD failure analysis, the typical procedure and flow is shown as Fig. 2.2. The failure analysis flow is structured to gather as much data about the failure in a nondestructive manner period to proceeding with a destructive technique. As to fault isolation techniques and tools, they are essential to the process of designing and debugging ESD protection circuits and solving ESD problems in existing circuits. These tools include EMMI (Emission Microscopy), OBIRCH (Optical Beam Induced Resistance Change), TIVA (Thermally Induced Voltage Alterations) and LIVA (Light Induced Voltage Alterations) are characterized in Table 2.1.
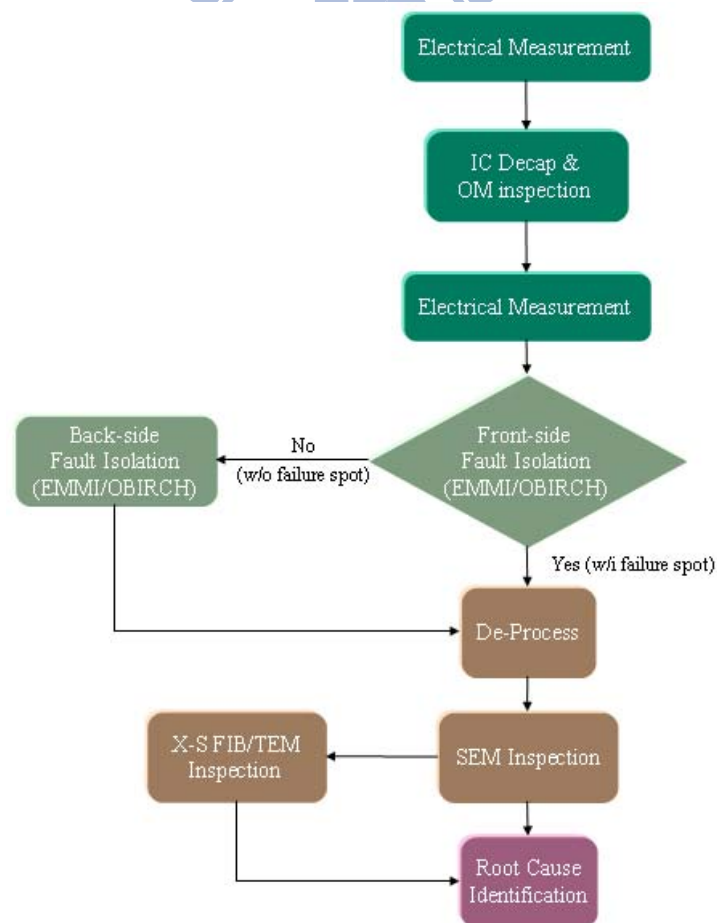
Fig. 2.2　　Typical FA procedure and flow of ESD failure analysis.

Table 2.1　　Characteristics of four fault isolation tools.

Fault Isolation Tools	Physical Principle	Bias Model	Applications
Emission Microscopy (EMMI)	Photos are generated as a result of electron-hole pair recombination and generation.	Depends on the failure test conditions.	Devices damage including junction, contact and substrate damage or short.
Optical Beam Induced Resistance Change (OBIRCH)	Resistance variation due to laser heating. (Thermally induced the variation of TCR, Temperature coefficient of restistance)	Force constant voltage and measure the variation of current.	Metal shorts or high resistance, junction, contact and substrate damages.
Thermally Induced Voltage Alterations (TIVA)	Resistance variation due to laser heating. (Thermally induced the variation of TCR, Temperature coefficient of restistance)	Force constant current and measure the variation of voltage.	
Light Induced Voltage Alterations (LIVA)	Currents are changed as a result of laser induced electron-hole pair generation.	Force constant current and measure the variation of voltage.	Open junctions and substrate damage.

With the failure mechanism of CDM ESD event is gate oxide damage of MOS transistor in internal circuits, in failure analysis, a fault isolation technique is need to localize the failure location. Emission Microscopy (EMMI) is the most popular tool which takes advantage of the electro-luminescent characteristics of silicon devices [17], [18]. Fig. 2.3 shows the appearance pictures of EMMI equipment with HAMAMATSU Phemos-1000 system.

Fig. 2.3　　HAMAMATSU Phemos-1000 Emission Microscope system.

For ESD FA, the evaluation of the electro-luminescent characteristics in both forward- and reverse-biased states provides information about defects, faults, failures, and device operation. Photons are generated as a result of electron–hole pair (EHP) recombination and generation. As a minority carrier recombines with the majority carrier, a photon is emitted in EHP recombination. Additionally, photon emission can be used to find oxide and dielectric failures. The CDM ESD events can introduce "pin hole" defects. Oxide defects can be observed by using electro-luminescent techniques. Fig. 2.4 shows the diagram of EMMI operation principle.
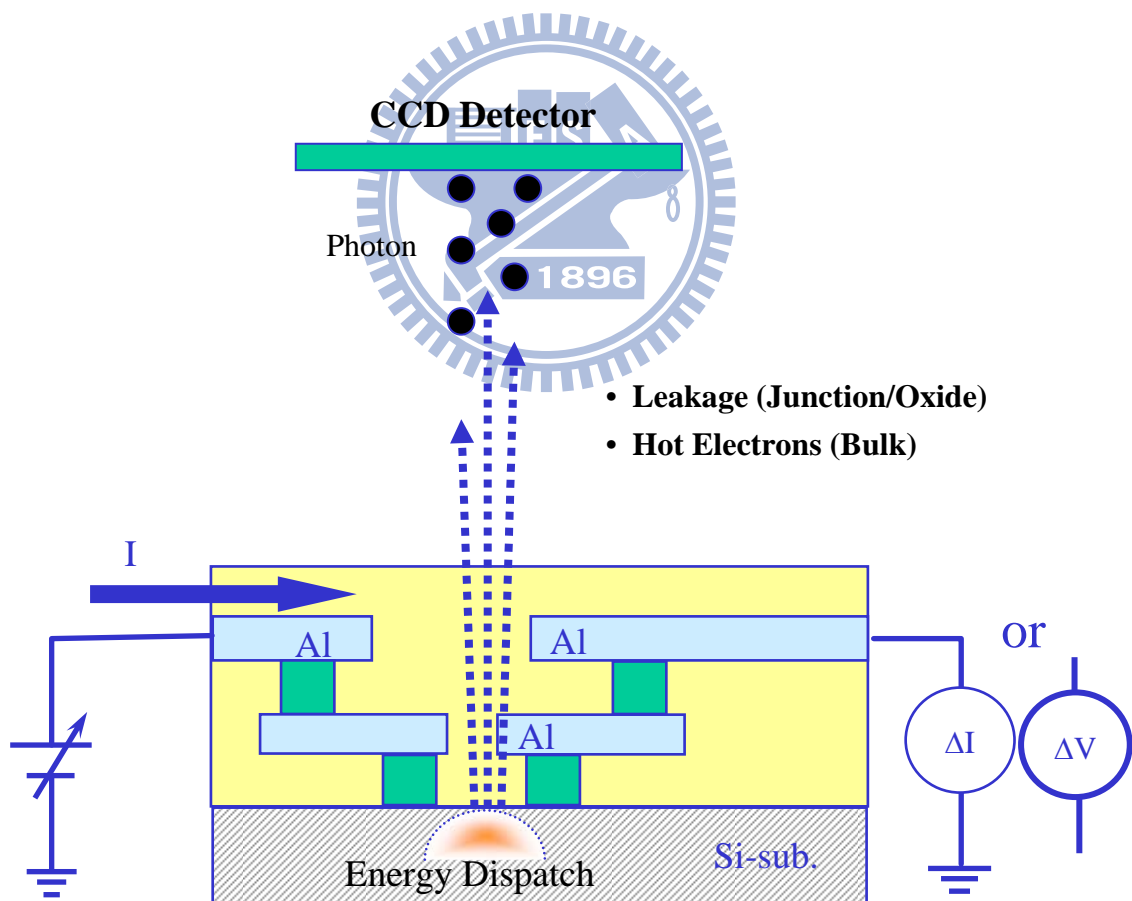
Fig. 2.4   Diagram of EMMI operation principle.

When gate oxide of MOS transistor in internal circuits is damage or defective during CDM ESD events, the emitted photons in EHP recombination can be detected by EMMI in failure device. The emission detection mechanism is that electrons were injected into the gate oxide by Fowler-Nordheim (F-N) tunnelling. The electrons in the gate oxide conduction band can gain energy while travelling through the gate oxide and become hot electrons. Hot electrons that reach the poly gate can either recombine directly or produce additional electron-hole pairs in the poly, which may subsequently recombine. The electron-hole recombination will lead to photon emission and detected by EMMI detector.

2.3 Case Study on Chip-Level CDM ESD Damage

An input buffer fabricated in a 0.6-μm CMOS process is shown in Fig. 2.5. This chip is equipped with ESD protection circuit at its input pad, but it is still damaged after 1000V CDM ESD test. Due to the consideration of noise isolation between I/O cells and internal circuits, the VSS of I/O cells (VSS_I/O) and the VSS of internal circuits (VSS_Internal) are separated in the chip layout. After de-process the chip to poly layer, the failure site after CDM ESD test is at the gate oxide of the NMOS (M_N) in the input buffer, as shown in Fig. 2.6 with the inspection of scanning electron microscopy (SEM). As a result, the ESD protection device at the input pad can not efficiently protect the gate oxide during CDM ESD stress, because there is no connection between VSS_I/O and VSS_Internal. The CDM ESD current path which damaged the gate oxide of NMOS (M_N) is shown by dash line in Fig. 2.5.
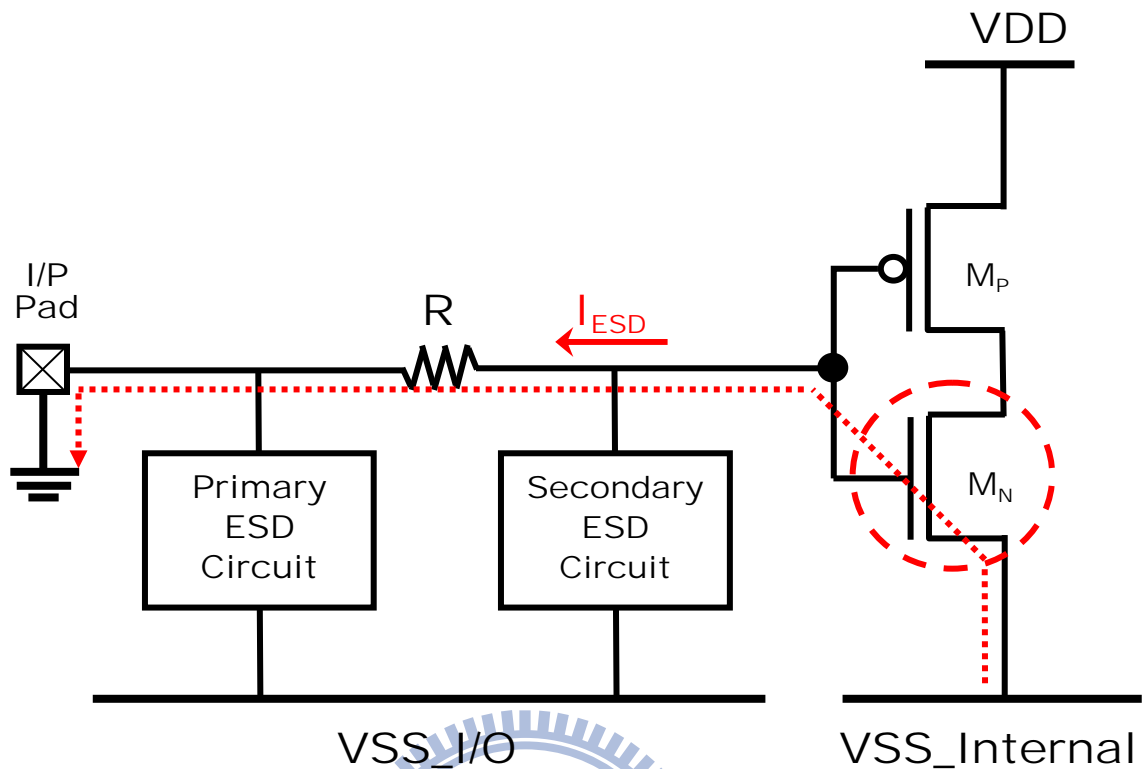
Fig. 2.5 CDM ESD current path in an input buffer.
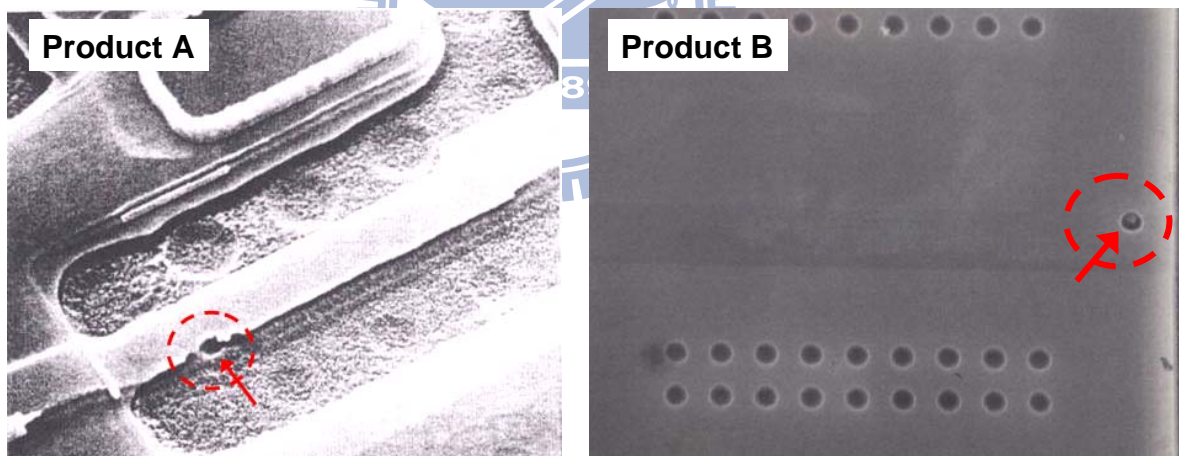
Fig. 2.6 SEM failure pictures of the failure site with gate oxide damage in the input NMOS (M_N).

An output buffer fabricated in a 0.5-μm CMOS process is shown in Fig. 2.7(a). This chip is damaged after 100V CDM ESD test. Fig. 2.8 is the failure picture inspected by SEM. The SEM pictures had proven that the failure caused by CDM ESD event is located at the poly gate of a NMOS (M_N)

transistor in the internal circuit that is connect to the output pad. The CDM ESD current which damaged the gate oxide of NMOS (M_N) is shown by dash line in an output buffer circuit and a diagram of cross-sectional view in Fig. 2.7(a) and (b), respectively.

(a)

(b)
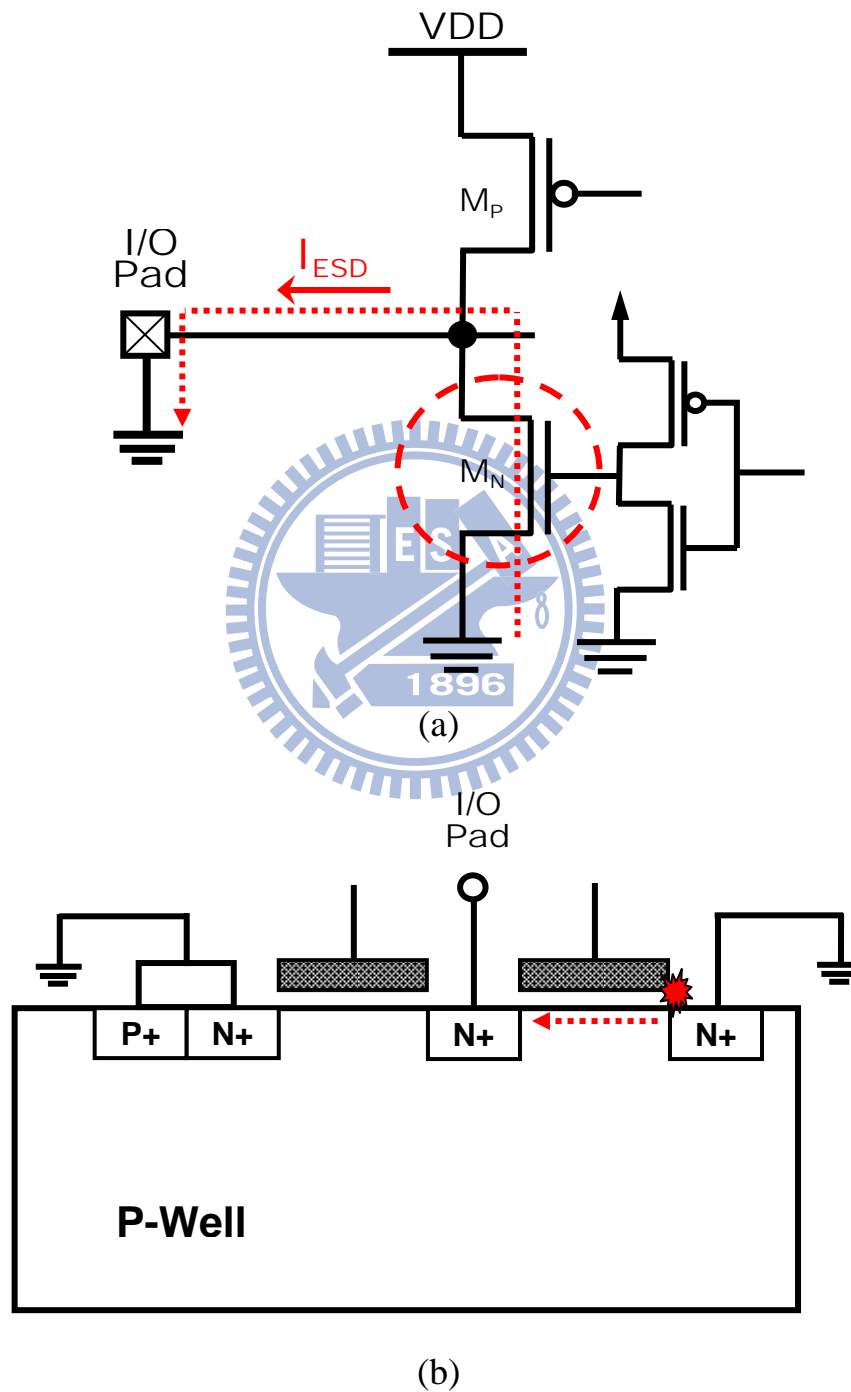
Fig. 2.7   CDM ESD current path in (a) an output buffer circuit, and (b) the diagram of cross-sectional view.
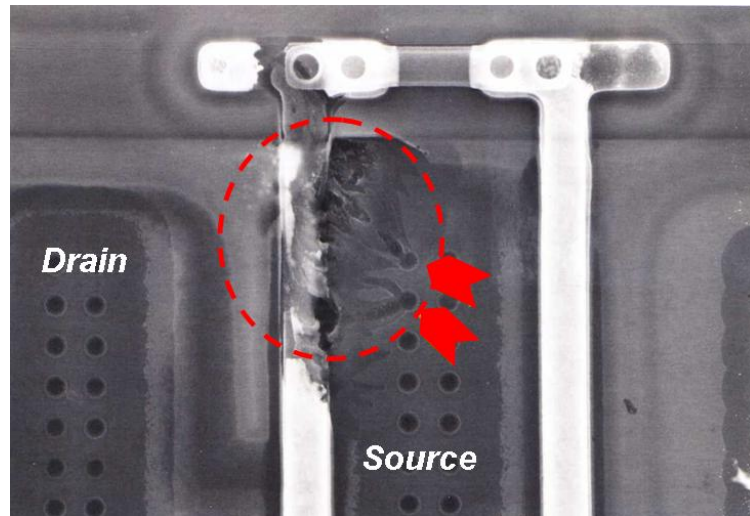
Fig. 2.8   SEM failure pictures of poly gate damage on a NMOS (M_N) transistor in the internal circuit.

From Fig. 2.9 to Fig. 2.11, there are two cases shown that the gate oxide damages by using the EMMI to localize the failure device are also located in the internal circuit after CDM ESD tests. In these two cases, the charges stored in the body of chip still flow through the gate terminal of the input or output MOS transistor in the internal circuits and damage its gate oxide during CDM ESD stresses, even though ESD protection circuits have been applied to the I/O pads. The pins near the corners in IC products are more often to suffer CDM ESD events, because the corner pins are usually first touched by external ground during transportation or assembly [19].
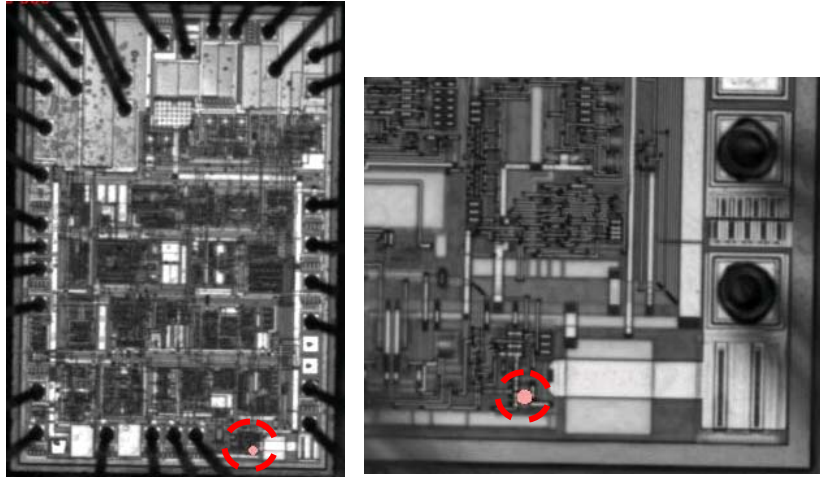
(a)
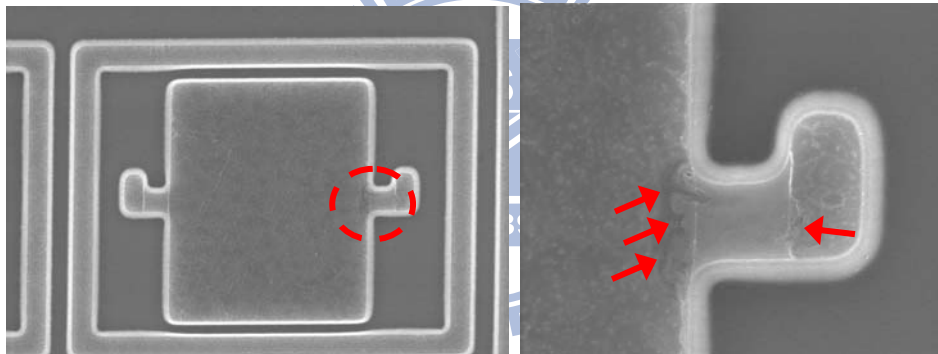
(b)

Fig. 2.9   After chip-level CDM ESD test, the failure point is located by
(a) EMMI detection, and (b) SEM inspection.
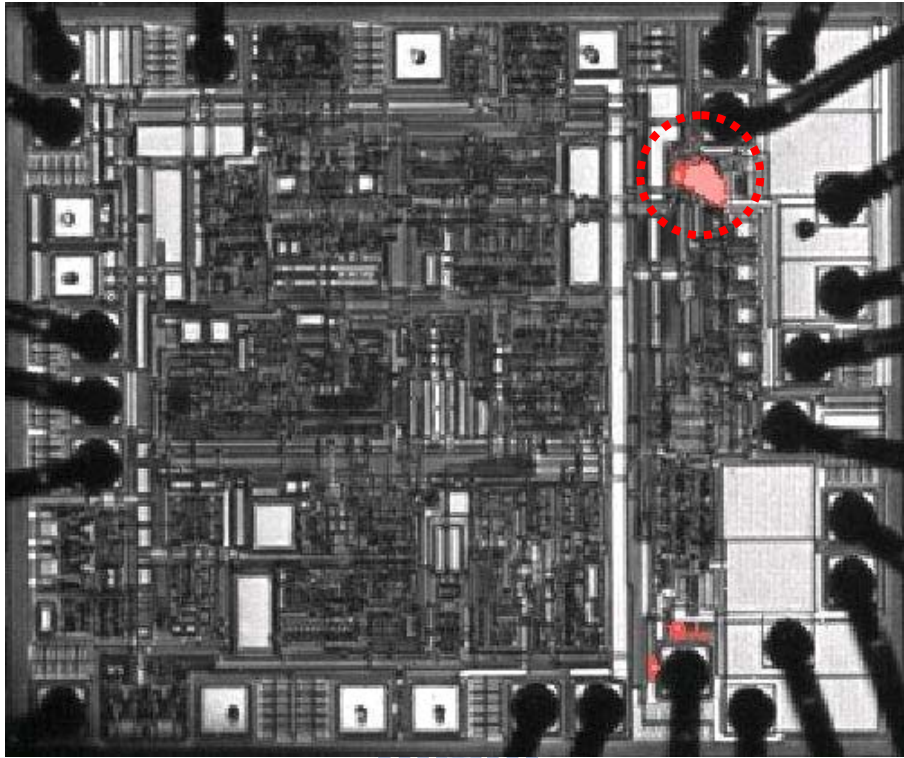
(a)
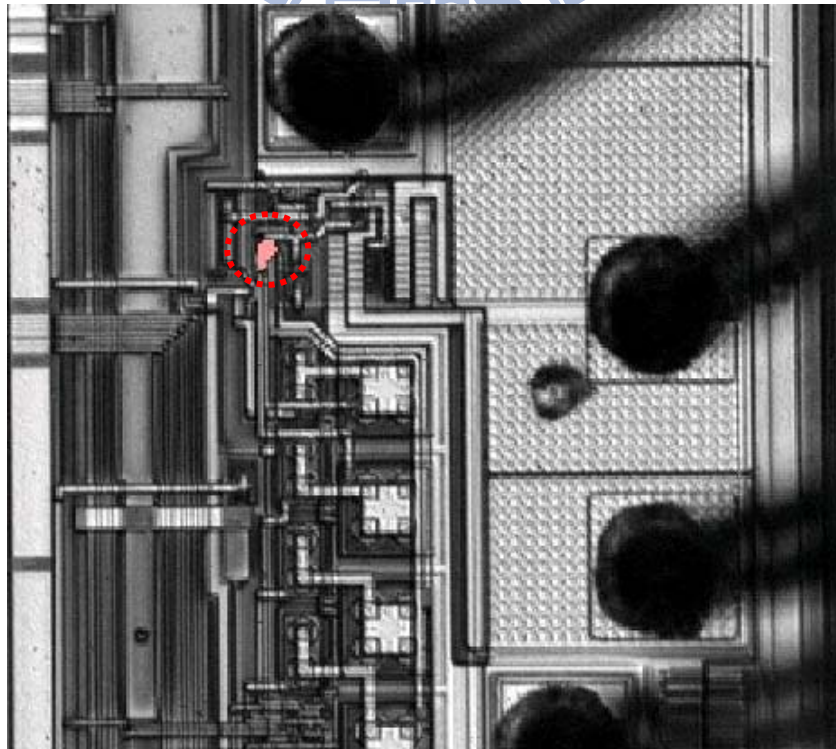
(b)

Fig. 2.10   After chip-level CDM ESD test, the failure point is located by (a) EMMI detection. (b) The enlarged image of its ESD failure location.
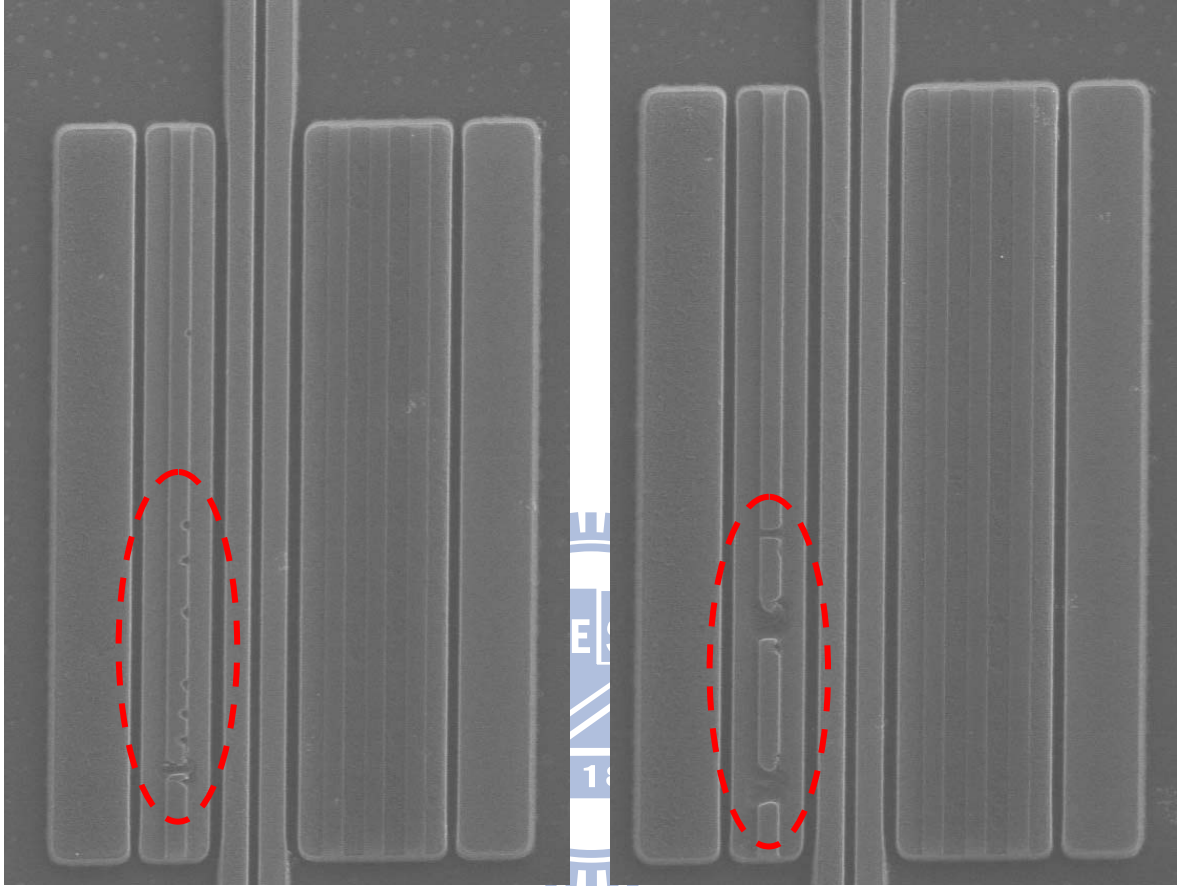
Fig. 2.11   SEM failure pictures of gate oxide damage in internal circuit.

2.4   Board-Level CDM ESD Event in IC Products

In microelectronic system, IC chips must be attached to the PCB. Before the attachment, static charges could be stored in the body of the chip or the metal traces on the dielectric layer in the PCB. During the attachment, the static charges originally stored in the chip and the PCB will be redistributed [20], as illustrated in Fig. 2.12. Fig. 2.13 shows the charge redistribution mechanism. When two capacitors with different voltages are shorted, charge

redistribution will occur. C_{chip} and C_{board} denote the parasitic capacitances of IC chip and the printed circuit board (PCB), respectively. C_{chip} and C_{board} are not connected in the beginning. When the IC chip is attached to the PCB, C_{chip} and C_{board} are shorted. Consequently, the voltages across C_{chip} and C_{board} will become

$$V_{Final} = \frac{C_{chip} \times V_{chip} + C_{board} \times V_{board}}{C_{chip} + C_{board}} \tag{2.1}$$

after they are connected together. The instantaneous current during the attachment of IC chip to PCB will be significantly increased if the initial voltage difference between the IC chip and PCB is larger. The peak current during the charge redistribution can easily damage the IC to cause a CDM-like failure. After the chip is attached to the PCB, certain pins in the PCB may be connected to low potential or accidentally grounded during module function test, as illustrated in Fig. 2.14. In this situation, the charges originally stored in the chips and PCB will be quickly discharged through the grounded pin to damage the chips on the PCB. If the voltages across the equivalent capacitances of the chips and PCB are larger, more charges are stored, which leads to larger discharging current.

Fig. 2.12   Charges stored in the PCB and the charges stored in chip will be redistributed when the chip is attached to the PCB.
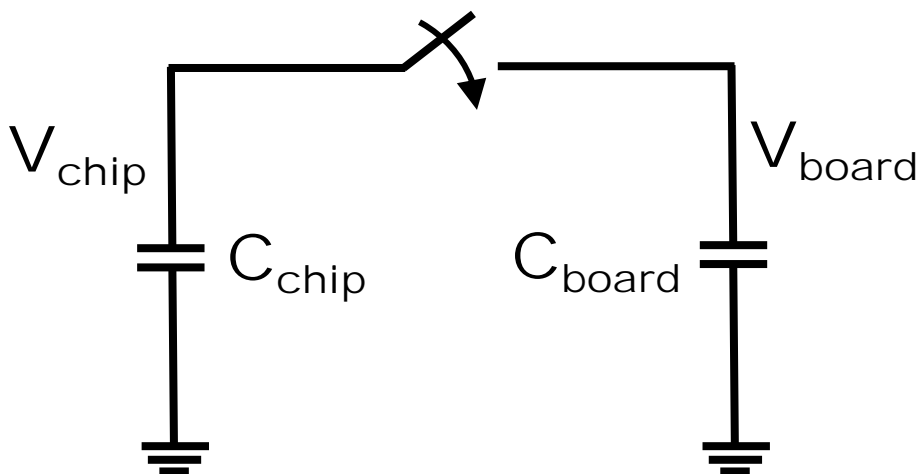
Fig. 2.13   When two capacitors with different voltages are shorted, charge redistribution will occur.
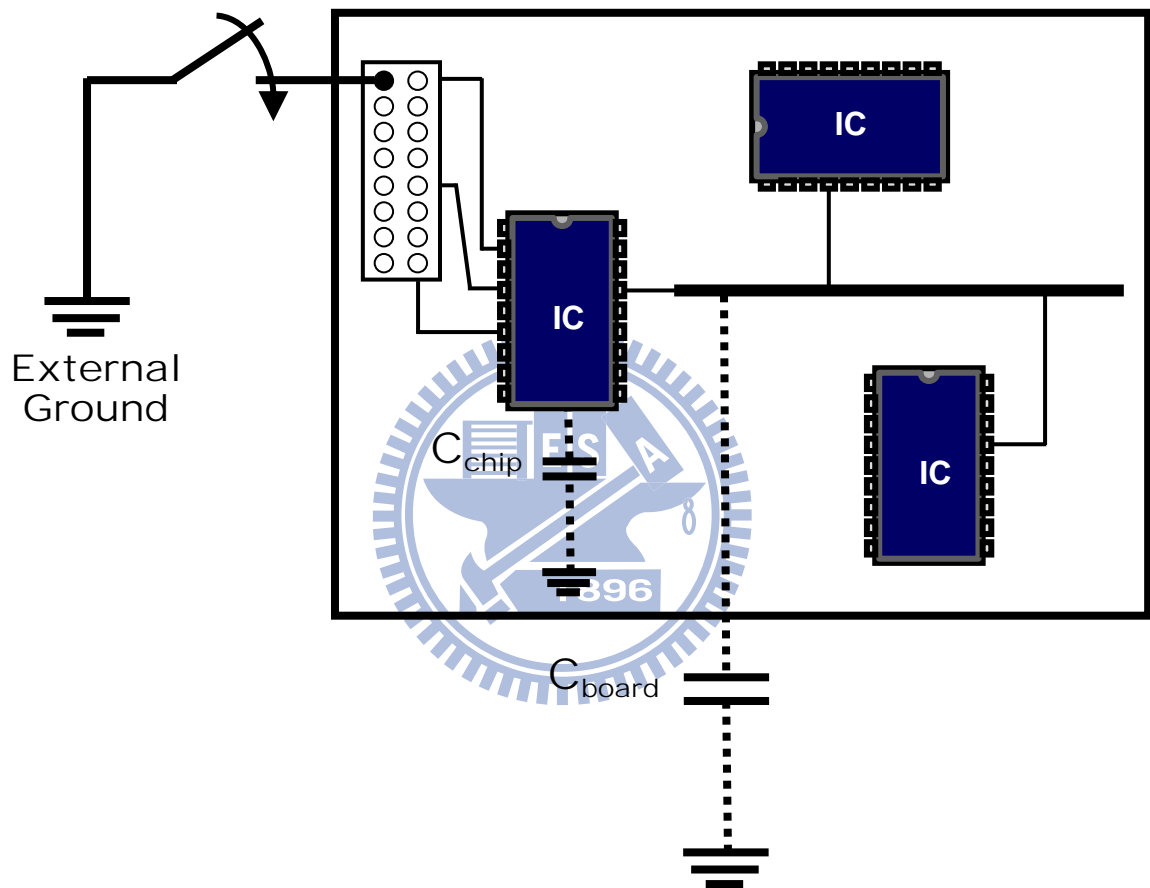
Fig. 2.14   When a certain pin of the PCB is grounded during the module function test, huge current will flow from the PCB through the IC to the external ground.
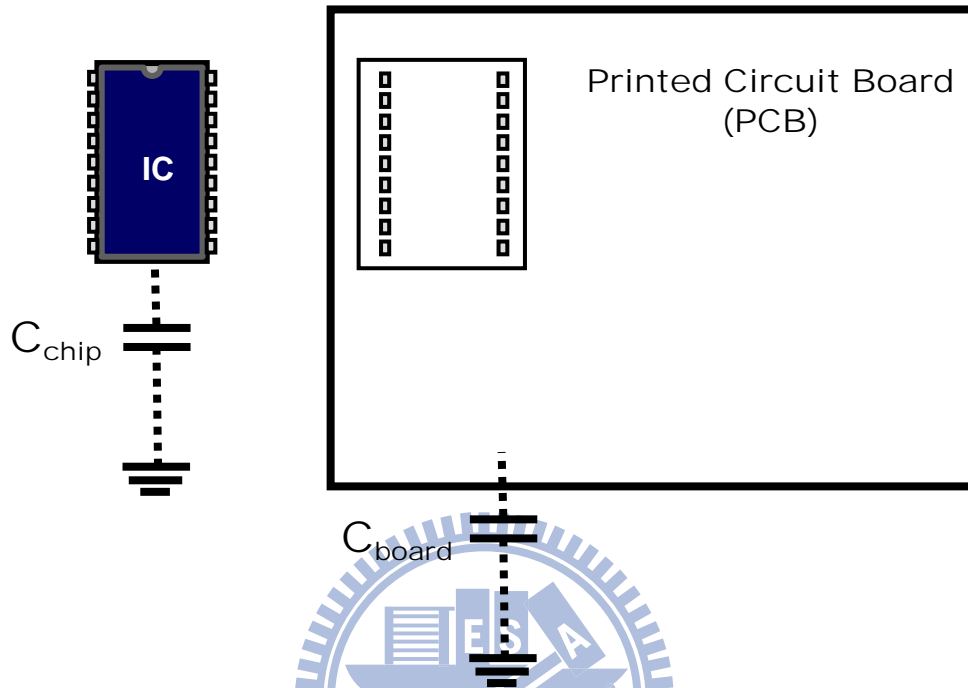
CHAPTER 3

CHIP-LEVEL AND BOARD-LEVEL CDM ESD

TESTS ON IC PRODUCTS

Recently, it was informed from the IC industry that some IC products which already passed the component-level ESD specifications were still damaged by the CDM-like ESD events in the field applications. Besides, it had been reported that the customer-returned ESD damages can be duplicated by the board-level CDM ESD test [11], [12]. Some studies which evaluated the discharging current under different charged board dimensions in the board-level CDM ESD tests for different IC applications had been reported [21], [22]. Board-level CDM ESD event often causes the ICs to be damaged after the ICs are installed to the circuit boards of electronic systems. For example, board-level CDM ESD events often occur during the module function test on the circuit board of electronic system. Even though the IC has been designed with good chip-level ESD robustness, it could have a reduced ESD level in board-level CDM ESD event.

The reason is that the discharging current during the board-level CDM ESD event is significantly higher than that during the chip-level CDM ESD event. The board-level CDM ESD issue becomes more important in the real-world applications of IC products which are fabricated in nanoscale CMOS processes with the much thinner gate oxide. In this chapter, three kinds of PCBs are used to compare the equivalent board capacitances (C_B), discharging current waveforms, and peak discharging currents under board-level CDM ESD tests. Moreover, a two-layer PCB with FR4 dielectric

layer is employed to perform the board-level CDM ESD tests on the test circuits fabricated in 0.25-μm and 130-nm CMOS processes.

The electrostatic discharge (ESD) transient currents and failure analysis (FA) between chip-level and board-level charged-device-model (CDM) ESD tests are also investigated in this chapter. The discharging current waveforms of three different printed circuit boards (PCBs) are characterized first. Then, the chip-level and board-level CDM ESD tests are performed to an ESD testkey of GGNMOS, an ESD-protected dummy NMOS, and a high-speed receiver front-end circuit, respectively. Optical beam induced resistance change (OBIRCH) is using to detect and localize the failure sites. Scanning electron microscope (SEM) failure pictures show that the board-level CDM ESD test causes much severer failure than that caused by the chip-level CDM ESD test.
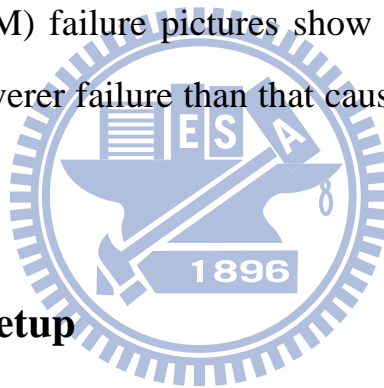
3.1 Measurement Setup

A CDM ESD test system, Oyrx CDM Orion, was used for field-induced chip-level and board-level CDM ESD tests. The equipment picture is as Fig. 3.1 shown. The experimental setups of chip-level and board-level CDM ESD tests are shown in Fig. 3.2 and Fig. 3.3, respectively. In the traditional chip-level CDM ESD test, only the IC chip (DUT) is put on the charging plate of the field-induced CDM ESD tester. On the contrary, both the IC chip and the test board on which the IC chip is mounted are put on the charging plate of the field-induced CDM ESD tester in the board-level CDM ESD test. With a 40-pin dual-in-line-package (DIP) socket soldered on the PCB, the

packaged test circuit can be mounted on the PCB to perform the board-level CDM ESD test.
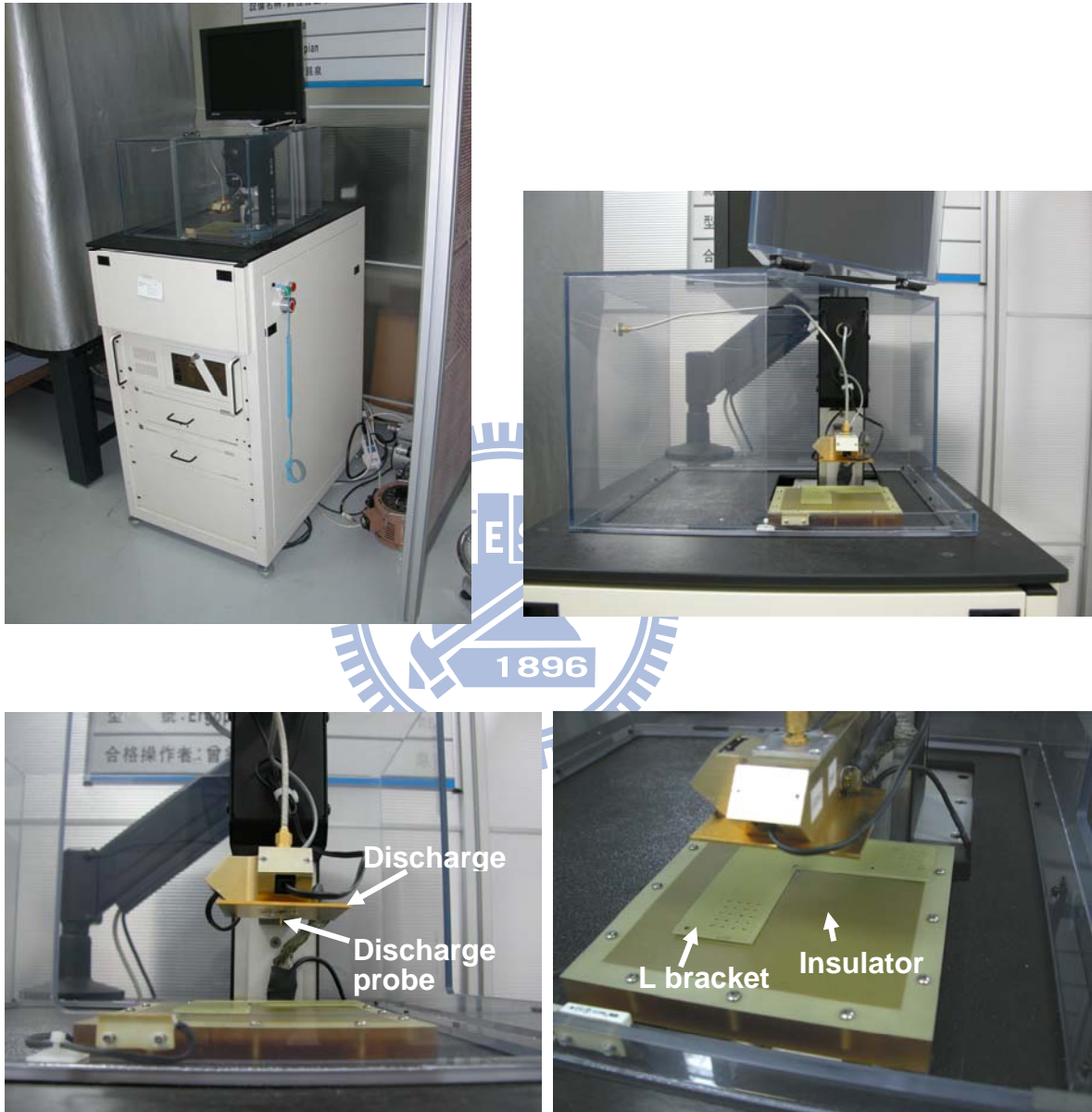
Fig. 3.1　　The non-socket CDM tester (Oyrx CDM Orion).
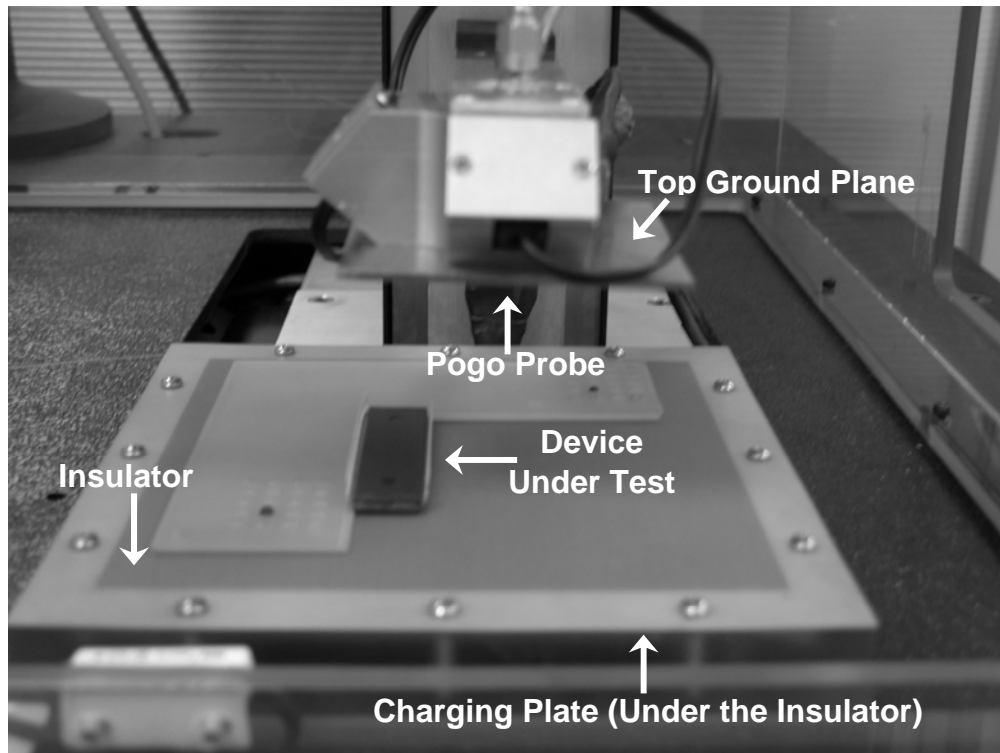
Fig. 3.2   Measurement setup of field-induced chip-level CDM ESD test.
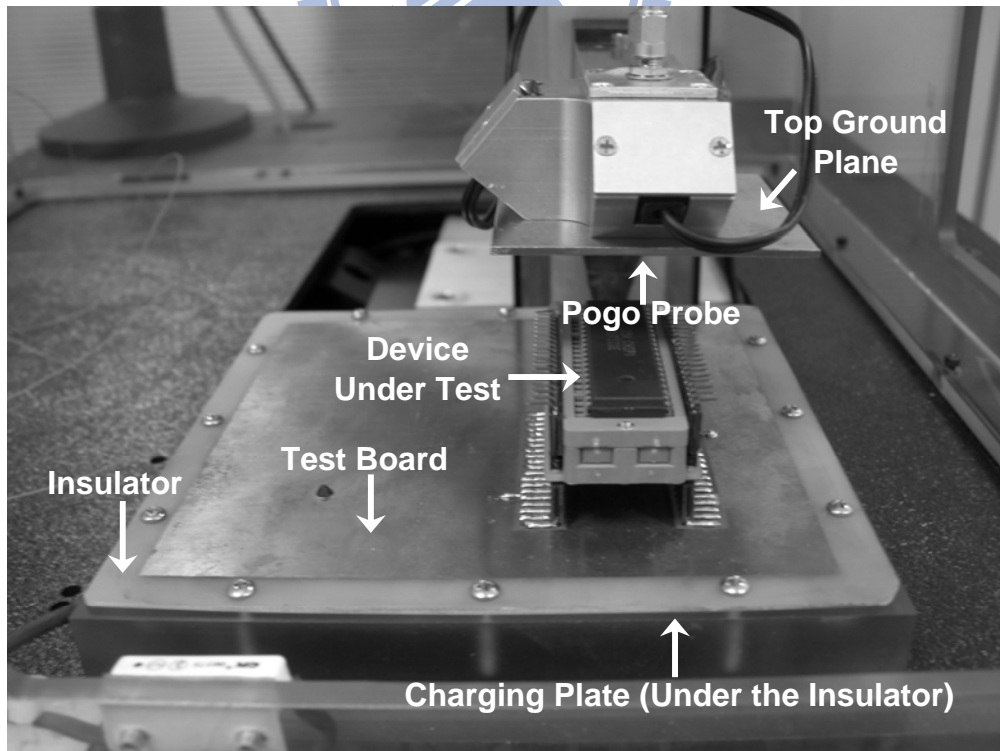
Fig. 3.3   Measurement setup of field-induced board-level CDM ESD test.

Three different two-layer PCBs were chosen to investigate their board capacitances (C_B), because the board capacitance is a key factor in board-level CDM ESD tests. The characteristics of these three PCBs are listed in Table 3.1. The board capacitances and discharging current waveforms were monitored by Agilent 4275A LCR meter at 1 MHz and Tektronix 680C oscilloscope, respectively, as Fig. 3.4 and Fig. 3.5 shown.

Table 3.1   Characteristics of three different kinds of PCBs.
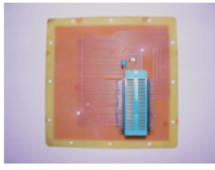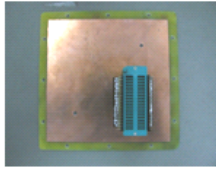
	PCB_1		PCB_2		PCB_3	
PCB Type						
PCB Size	12.5cm x 12.5cm		15cm x 15cm		15cm x 15cm	
Board Capacitance	274 pF		324 pF		390 pF	
Charge Voltage	+500 V	+1000 V	+500 V	+1000 V	+500 V	+1000 V
Peak Current	5.71 A	9.16 A	8.65 A	13.8 A	13.41 A	22.96 A

Fig. 3.4   The LCR meter which is used for the measurement of capacitance (Agilent 4275A).

Fig. 3.5   The high-frequency oscilloscope (TDS 680C).

3.2   Experimental Results and Discussion

3.2.1   Board-Level CDM ESD Current Waveforms in Different PCBs

Because the board capacitance (C_B) is a key factor in board-level CDM ESD tests, the higher capacitance and the higher ESD discharging energy, three PCBs with different effective area of top plate are used to get which one has the largest capacitance. As Fig. 3.6 shown, the larger area of top plate will contribute the larger board capacitance (C_B). Due to the area limitation of bottom plate, 15 X 15 cm, the top plate is designed to with the area of the same as bottom plate. Table 3.1 lists the measured board capacitance and peak discharging current among these three PCBs under +500V and +1000V charged voltages. Fig. 3.7, 3.8, and 3.9 show their corresponding discharging current waveforms under +500V charged voltage. Higher peak currents were observed in PCB_3 due to the largest board capacitance and lowest resistance along the discharging path on PCB. In the board-level CDM ESD tests with IC products, PCB_3 was chosen as the test board.

PCB

Discharging Pin

GND

GND Plate (Top Plate)

C_B

Charging Plate (Bottom Plate)
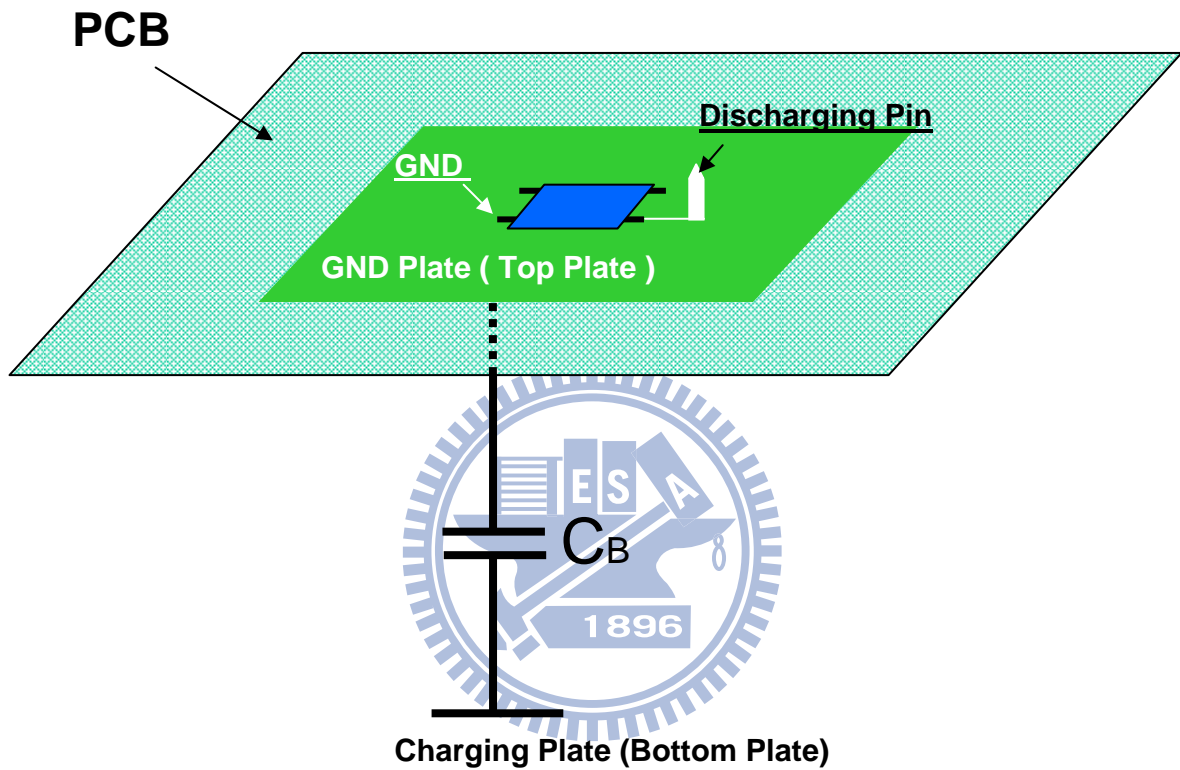
Fig. 3.6   The top plate of board capacitance (C_B) is connected to the GND plate of PCB. The bottom plate of C_B is the charging plate of CDM ESD tester.
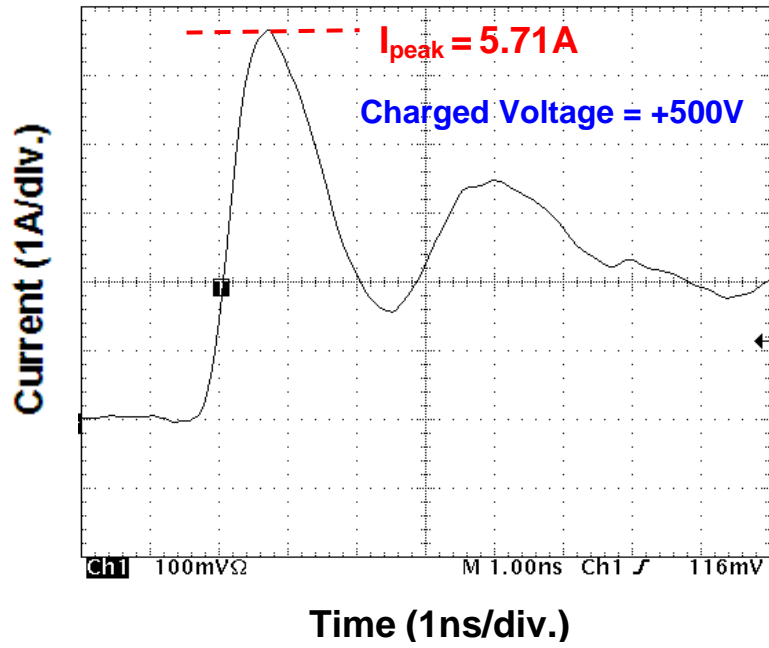
Fig. 3.7   Discharging current waveform of PCB_1 under +500V charged voltage.
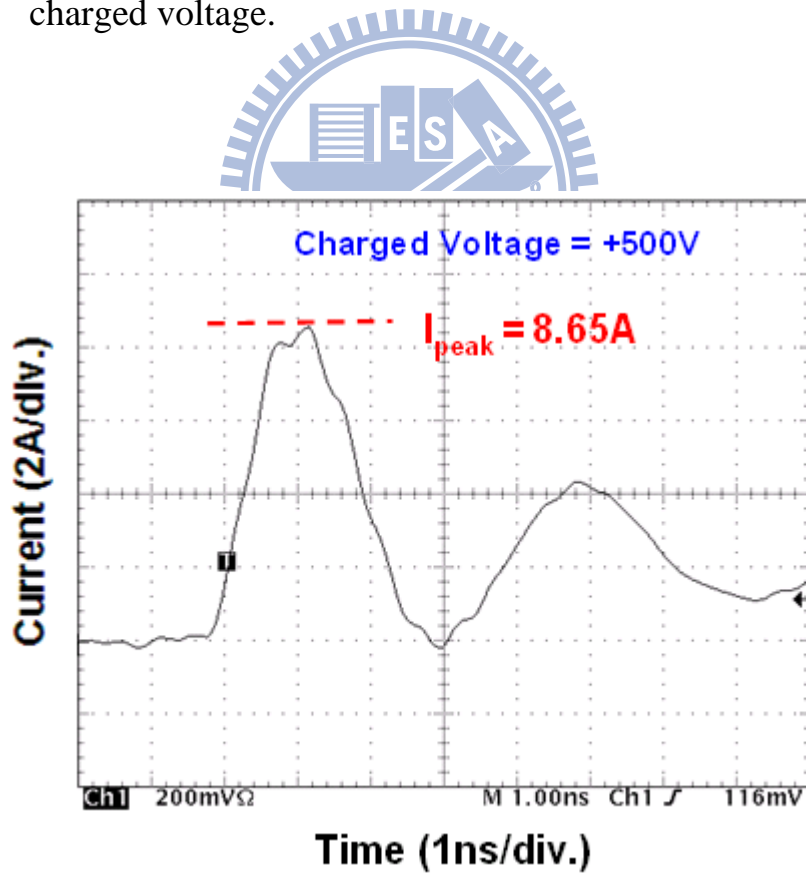
Fig. 3.8   Discharging current waveform of PCB_2 under +500V charged voltage.
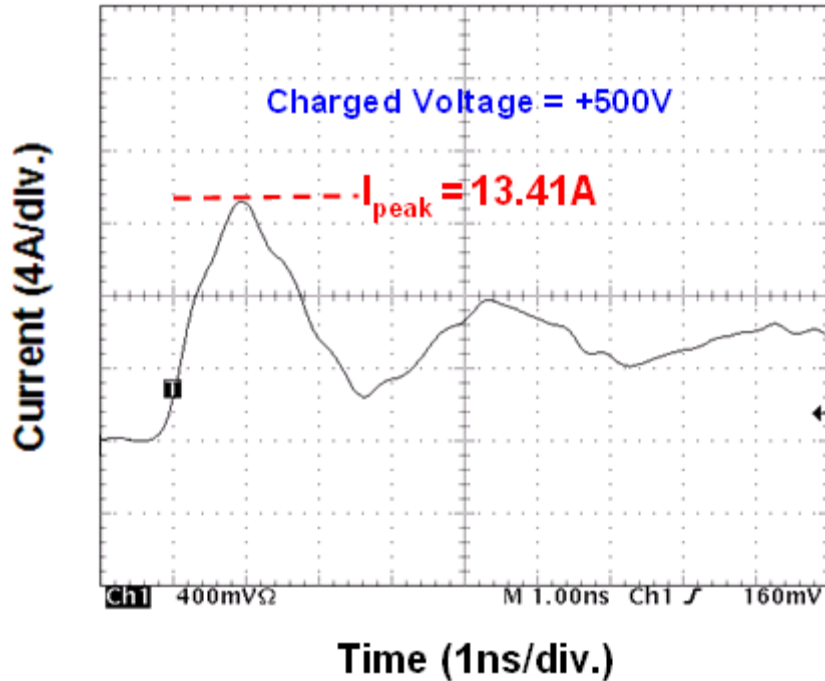
Fig. 3.9   Discharging current waveform of PCB_3 under +500V charged voltage.

3.2.2 Test With ESD Testkey of Grounded Gate NMOS (GGNMOS)

As shown in Fig. 3.10 and 3.11, the ESD protection device GGNMOS fabricated in a 0.25-um CMOS process was used as the test circuit. The gate and source terminals of the GGNMOS are connected to the GND. The drain terminal of the GGNMOS is connected to the input pad.
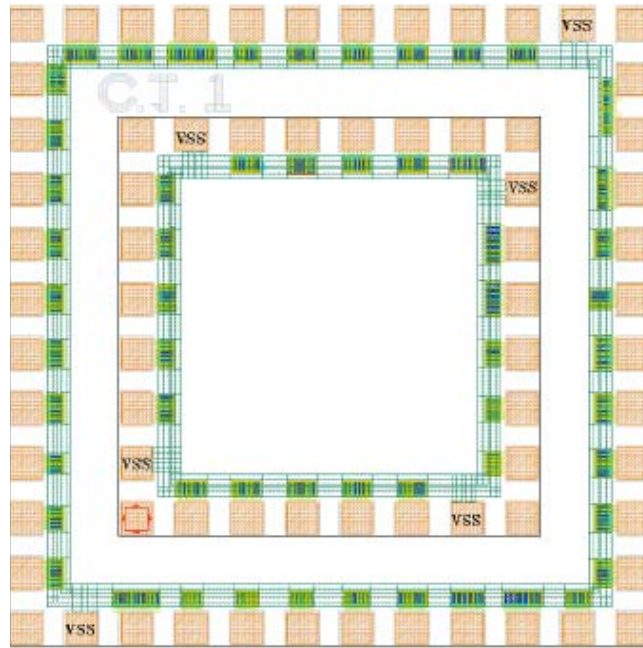
Single finger width = 15 um
Clearance from SAB edge to ploy edge = 0.3 um
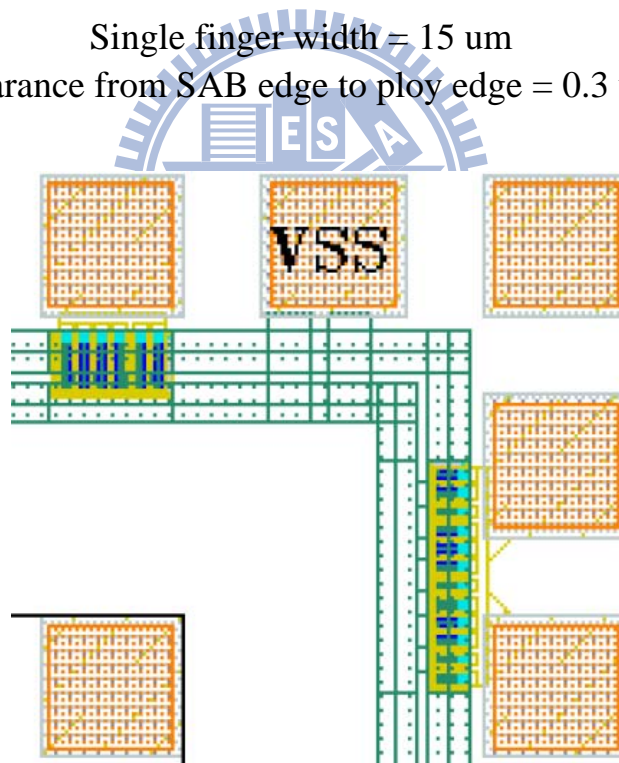
Fig. 3.10   Layout top view of ESD testkey with ESD protection device of GGNMOS.

Fig. 3.11 Test circuit of ESD testkey with GGNMOS for chip-level and board-level CDM ESD tests.

In the board-level CDM ESD test with PCB_3, the bottom layer of PCB_3 was connected to the test system as the charging plate, whereas the top layer was connected to the ground node of the test circuit. The tested pin under CDM ESD tests is the input pad. The measured results on the chip-level and board-level CDM ESD robustness of the GGNMOS are listed in Table 3.2. The GGNMOS passes +2000V for chip-level and board-level CDM ESD tests. The result indicates that the GGNMOS has high ESD robustness of chip-level and board level CDM tests.

Table. 3.2 Chip-level CDM and board-level CDM ESD robustness of GGNMOS.

JEDEC	CDM (Chip-Level)	CBM (Board-Level)
Start Voltage (V)	250	250
Step (V)	250	125
Zap Counts at each voltage	5	5
Pass Voltage (V)	2000	2000
Failure Criterion	± 1uA Voltage shift 30%	± 1uA Voltage shift 30%

Fig. 3.12 shows the discharging current path during chip-level and board level CDM tests. In chip-level CDM test, the discharging current flows from bulk Si to drain input pad due to the lowest resistance path. Similarly, the discharging current flows from source (GND) to drain input pad under board-level CDM test. Thus, the GGNMOS is not the suitable ESD protection device to compare the difference of chip-level and board-level CDM events.
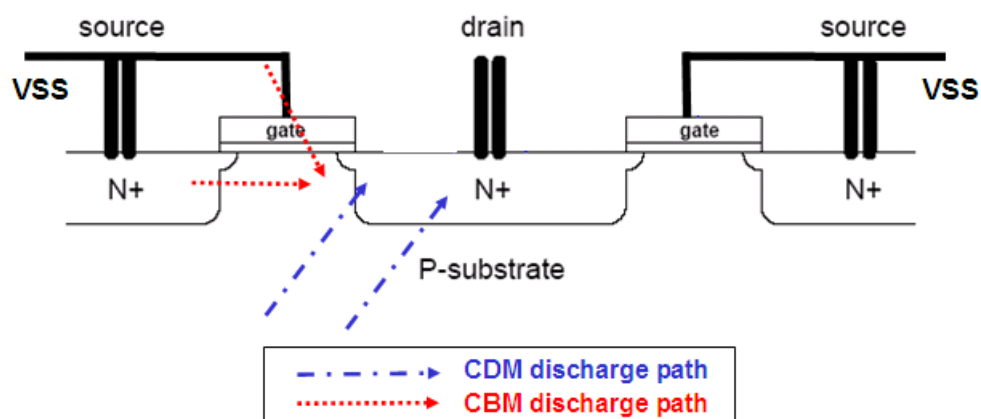
Fig. 3.12   Discharge current path of GGNMOS under CDM ESD test.

3.2.3   Test With Dummy Receiver NMOS (RX_NMOS)

As shown in Fig. 3.13, the dummy receiver NMOS (RX_NMOS) fabricated in a 130-nm CMOS process was used as the test circuit. The gate terminal of the RX_NMOS is connected to the input pad to emulate the connection of a typical input NMOS in a receiver. The drain, source, and bulk terminals of the RX_NMOS are connected to VSS. On-chip ESD protection circuits are applied in the chip with the RX_NMOS together. The typical double-diode ESD protection scheme is applied to the input pad. The power-rail ESD clamp circuit consists of an RC timer, an inverter, and an

ESD clamp NMOS. The equivalent capacitance between the input pad and substrate of the RX_NMOS in the 40-pin DIP package is ~ 6.8 pF.
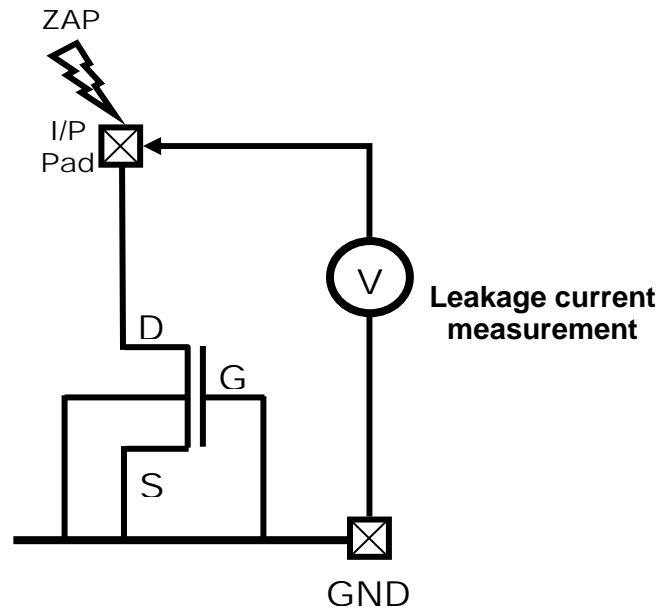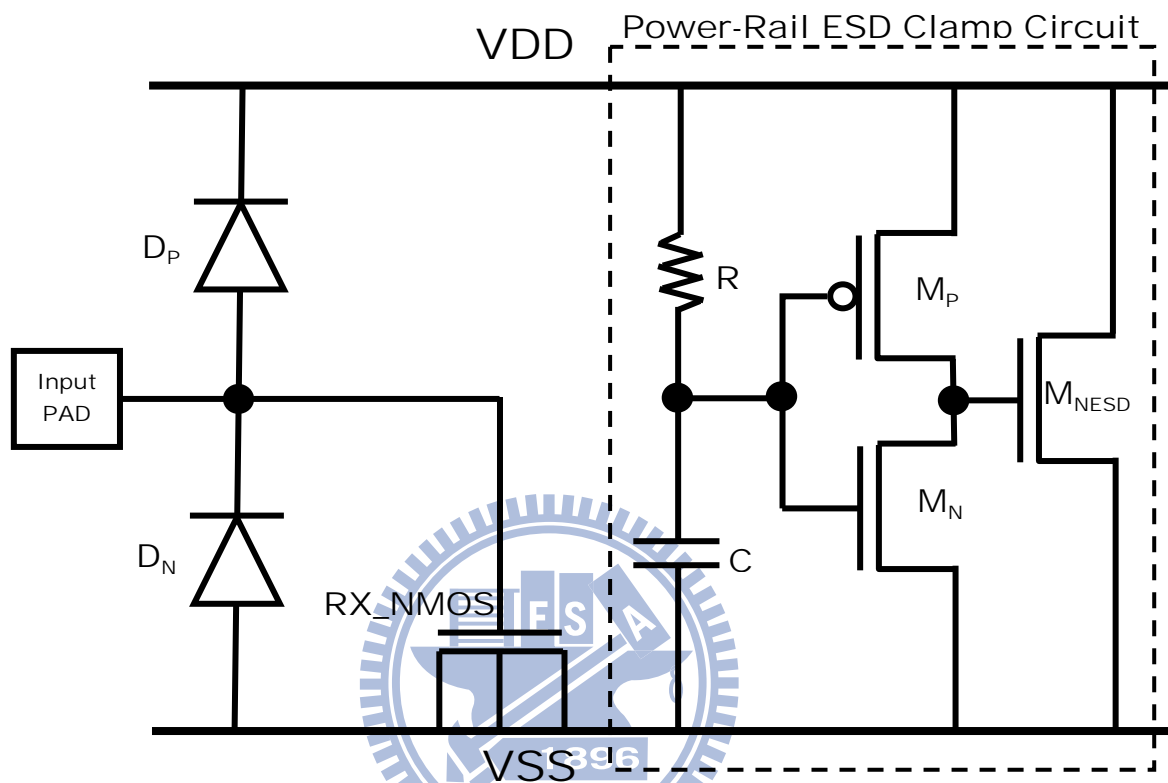

Fig. 3.13   Test circuit of RX_NMOS (dummy NMOS) for chip-level and board-level CDM ESD tests.

In the board-level CDM ESD test with PCB_3, the bottom layer of PCB_3 was connected to the test system as the charging plate, whereas the top layer was connected to the ground node of the test circuit. The tested pin under CDM ESD tests is the input pad. The discharging current waveforms under +200V chip-level and +200V board-level CDM ESD tests are shown in Fig. 3.14(a) and 3.14(b), respectively.
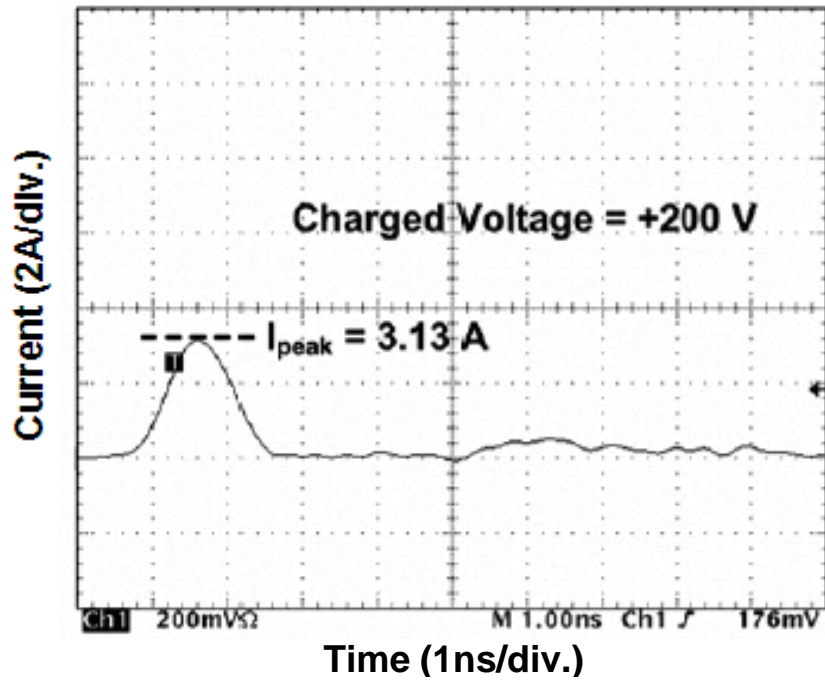
(a)
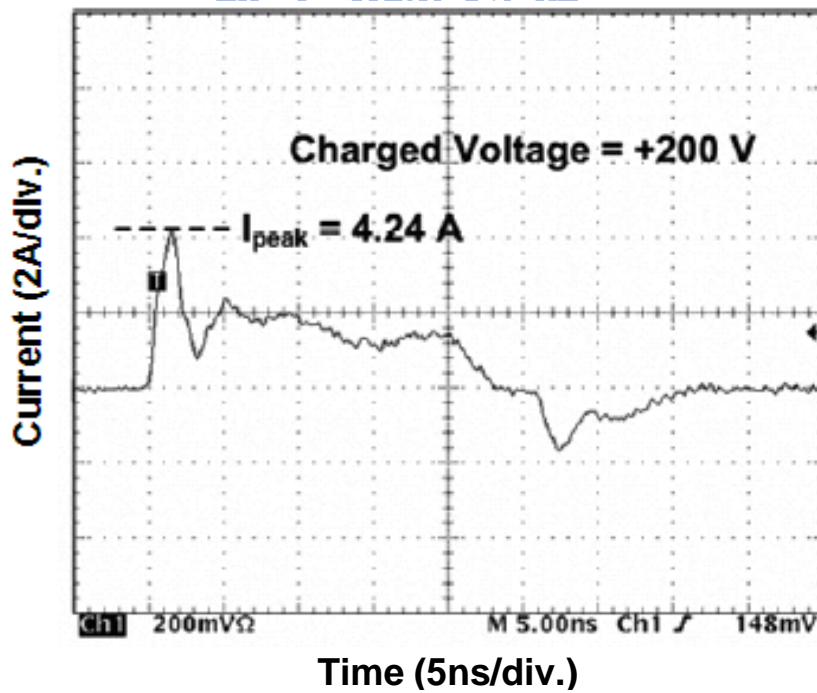
(b)

Fig. 3.14   Discharging current waveforms of RX NMOS under (a) +200V chip-level CDM, and (b) +200V board-level CDM, ESD tests.

As compared with the chip-level CDM ESD test, larger charging capacitance exists in the board-level CDM ESD test. Thus, the board-level CDM ESD test has higher peak discharging current, which results in lower ESD robustness of the IC. The peak discharging currents and measured results on the chip-level and board-level CDM ESD robustness of the RX_NMOS are listed in Table 3.3. The RX_NMOS passes +200V chip-level CDM ESD test, but fails at +200V board-level CDM ESD test. This result demonstrates that the board-level CDM ESD robustness is lower than the chip-level CDM ESD robustness, because the board-level CDM ESD event has much larger discharging current than that in the conventional chip-level CDM ESD event.

Table 3.3   Chip-level CDM and board-level CDM ESD robustness of RX_NMOS.

	Charged Voltage	Chip-Level CDM	Board-Level CDM
Peak Current	+100 V	N/A	1.72 A (Pass)
	+150 V	N/A	2.71 A (Pass)
	+200 V	3.14 A (Pass)	4.24 A (Fail)
	+400 V	8.43 A (Fail)	N/A

By using the optical beam induced resistance (OBIRCH) detection, as Fig. 3.15 shown, the failure sites caused by CDM ESD test were detected and located at the gate of the RX_NMOS. Fig. 3.16 shows the SEM failure pictures. The test samples were de-layered to the substrate layer so the damages at the gate oxide can be clearly observed. Comparing to Fig. 3.16(a) and 3.16(b), the ESD damage caused by board-level CDM ESD event is much worse than that caused by the chip-level CDM ESD event, because the board-level CDM ESD event has much higher discharging energy than that of chip-level CDM ESD event under the same charged voltage.
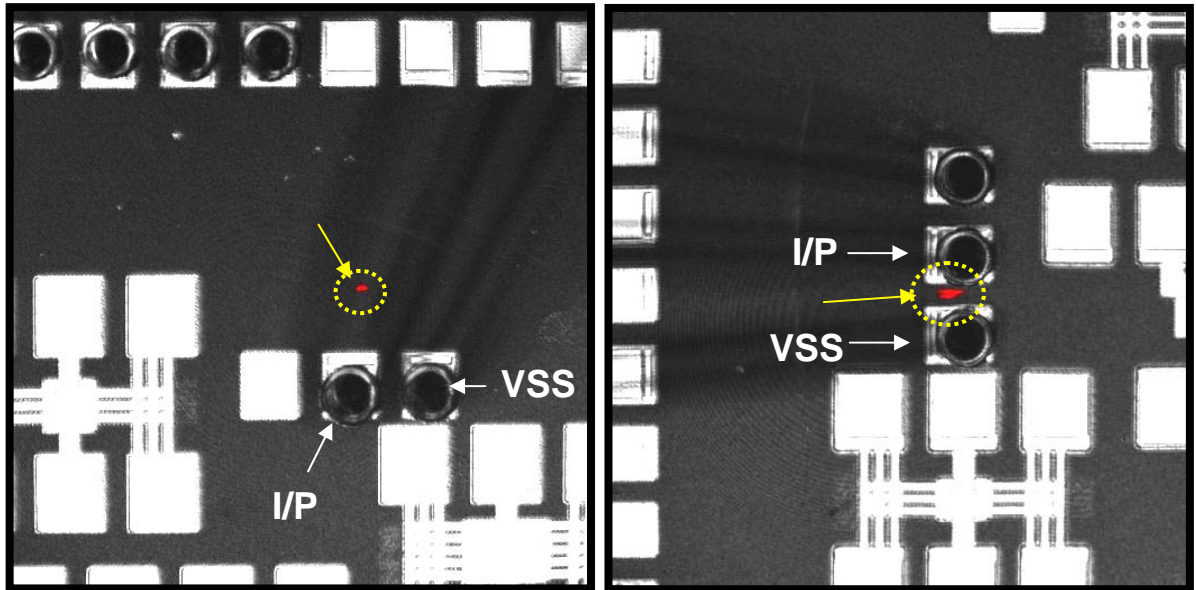
Fig. 3.15   OBIRCH pictures show the failure sites caused by CDM ESD test were detected and located at the gate of the RX_NMOS.

(a)   (b)
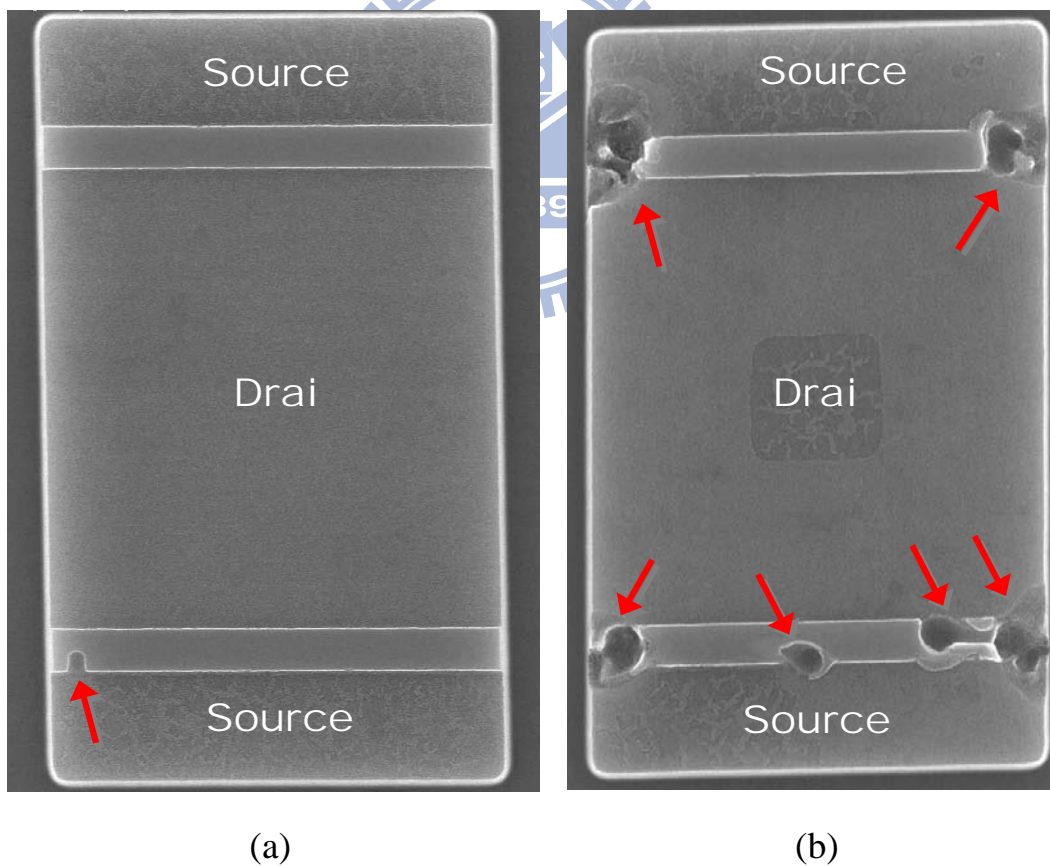
Fig. 3.16   SEM failure pictures of gate oxide damages at RX NMOS after (a) chip-level CDM, and (b) board-level CDM, ESD tests.

3.2.4 Test With 2.5-GHz High-Speed Receiver Interface Circuit

A 2.5-GHz differential high-speed receiver interface circuit fabricated in a 0.13-um CMOS process was also verified with the chip-level and board-level CDM ESD tests. Fig. 3.17 shows the circuit schematic of the 2.5-GHz differential high-speed receiver interface circuit with on-chip ESD protection design. The differential receiver interface circuit has the differential input stage realized by MOS transistors. The double-diode ESD protection scheme is applied to each differential input pad, and the P-type substrate-triggered silicon-controlled rectifier (P-STSCR) [21] is used in the power-rail ESD clamp circuit. Because of the high-speed application, the dimensions of ESD diodes under the input pads are limited to reduce the parasitic capacitance at the pads. The equivalent capacitance between the Vin pad and the substrate of the ESD-protected 2.5-GHz differential high-speed receiver interface circuit in a 40-pin DIP package is ~ 5.4pF.
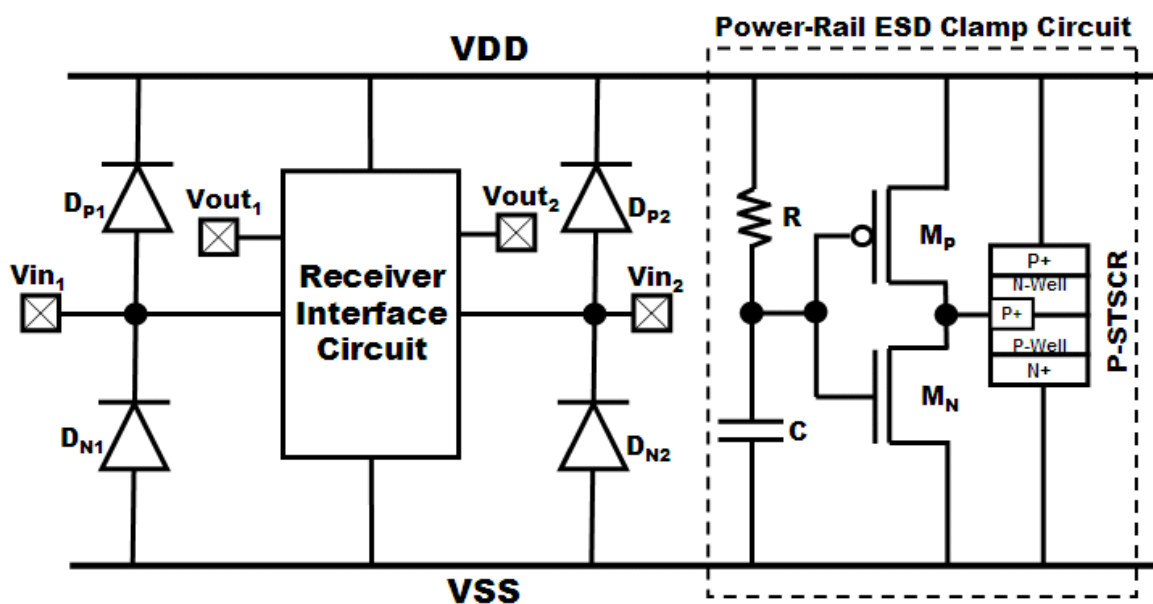
Fig. 3.17 Test circuit of 2.5-GHz high-speed receiver interface circuit for chip-level and board-level CDM ESD tests.

Besides, a reference high-speed receiver interface circuit without on-chip ESD protection circuit was also fabricated in the same process to compare their ESD robustness. The tested pin under CDM ESD tests is the Vin pad. The measured chip-level and board-level CDM ESD levels of the 2.5-GHz high-speed receiver circuits with and without on-chip ESD protection circuits are listed in Tables 3.4 and 3.5, respectively. The chip-level and board-level CDM ESD levels of the reference high-speed receiver interface circuit are quite poor, which fail at ±100V and ±50V, respectively. With the on-chip ESD protection circuits, the failure voltages of the high-speed receiver circuit during chip-level and board-level CDM ESD tests can be greatly improved to -1300V and -900V, respectively. Similarly, the board-level CDM ESD level is lower than the chip-level CDM ESD level. Failure analysis was performed on the ESD-protected high-speed receiver interface circuits after -1300V chip-level CDM ESD test and -900V board-level CDM ESD test.

Table. 3.4   Chip-level CDM ESD robustness of 2.5-GHz high-speed receiver interface circuit.

	Without ESD Protection		With ESD Protection	
Polarity	+	-	+	-
Failure Voltage	100 V	100 V	2000 V	1300 V

Table. 3.5   Board-level CDM ESD robustness of 2.5-GHz high-speed receiver interface circuit.

	Without ESD Protection		With ESD Protection	
Polarity	+	-	+	-
Failure Voltage	50 V	50 V	1300 V	900 V

The SEM failure pictures after chip-level and board-level CDM ESD tests are shown in Figs. 3.18 and 3.19, respectively. The failure points are located at the P+/N-well ESD diode D_{P1}. Although the ESD protection devices are successfully turned on during CDM ESD tests, the huge current during CDM ESD tests still damages the ESD protection devices. According to the SEM failure pictures, the failure is much worse after board-level CDM ESD test than that after chip-level CDM ESD test. This result has confirmed again that board-level CDM ESD events are more critical than chip-level CDM ESD events.
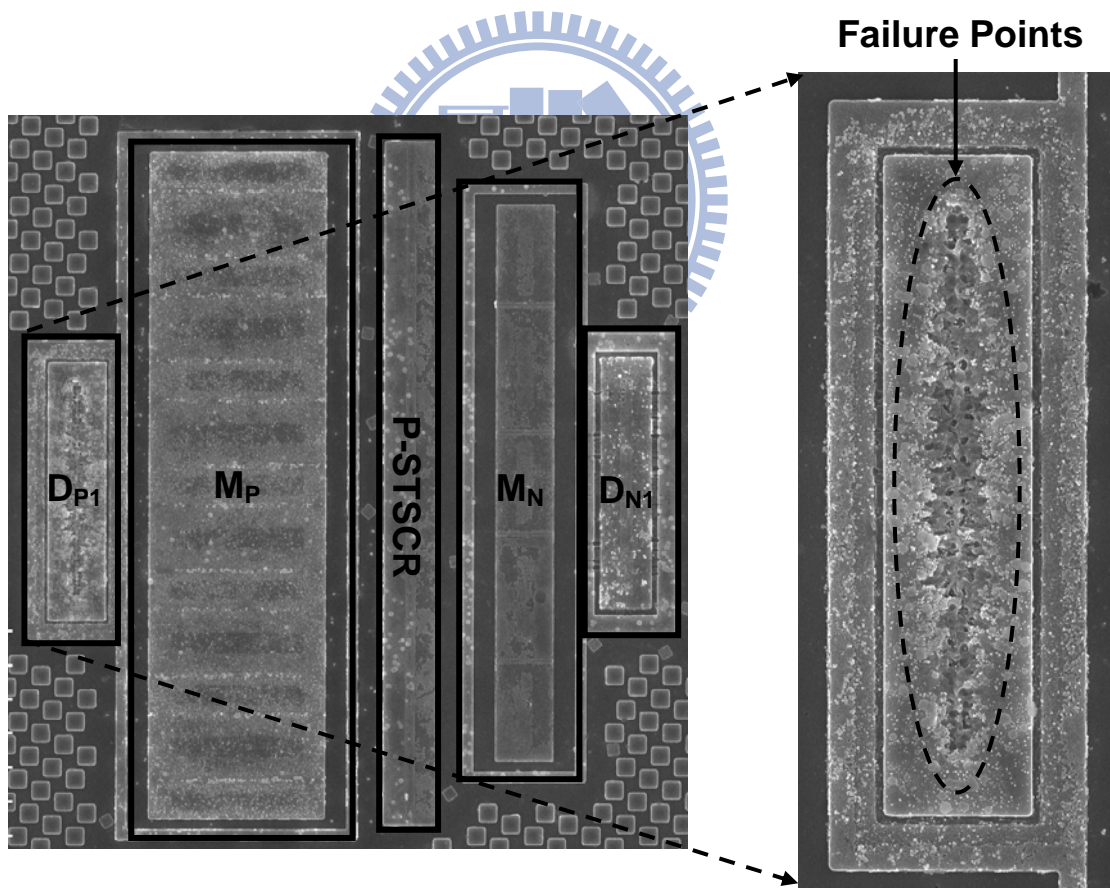
Fig. 3.18 SEM failure picture of the failure points on 2.5-GHz high-speed receiver front-end circuit after -1300V chip-level CDM ESD test.
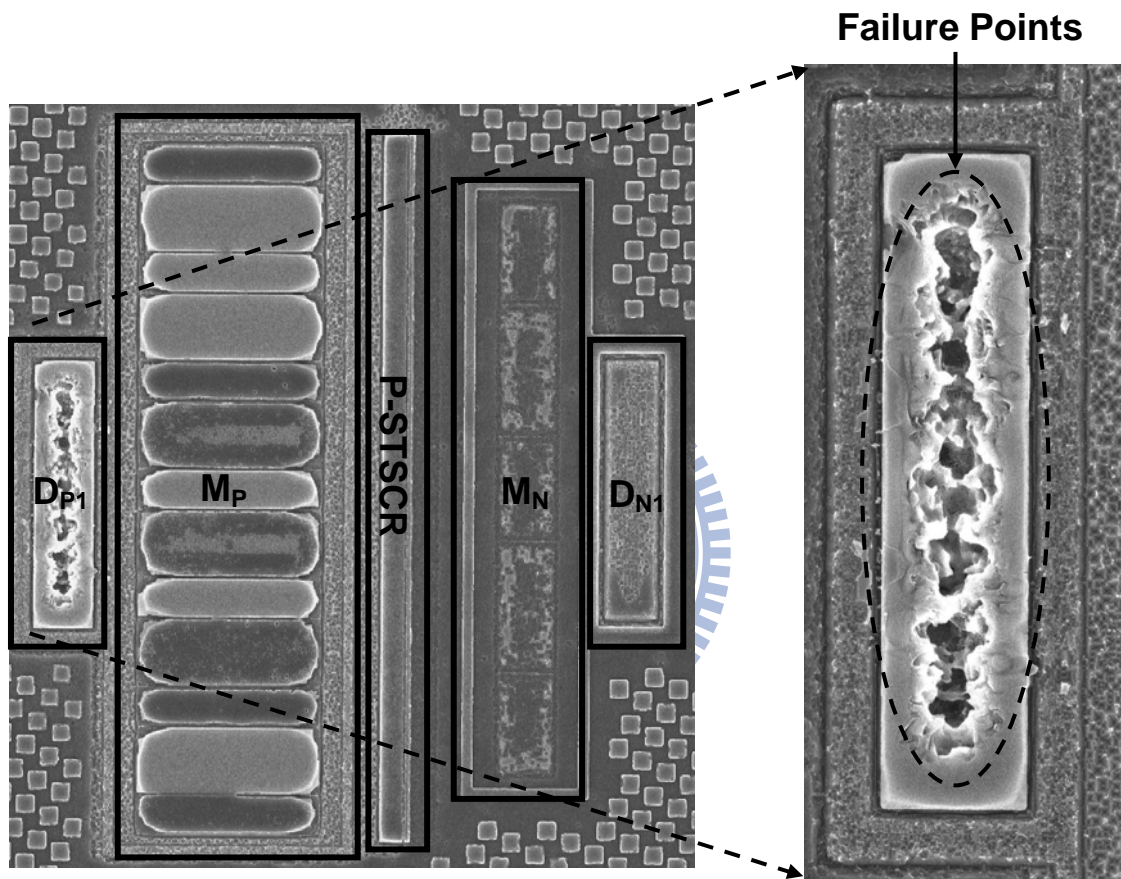
Fig. 3.19 SEM failure picture of the failure points on 2.5-GHz high-speed receiver front-end circuit after -900V board-level CDM ESD test.

3.3 Summary

Since the board capacitance is much larger than the IC chip capacitance, the board-level CDM ESD event has higher discharging energy than that of chip-level CDM ESD event. Thus, the board-level CDM ESD robustness is lower than the chip-level CDM ESD robustness. Failure analysis on the IC samples shows that ESD damage caused by board-level CDM ESD is much worse than that caused by chip-level CDM ESD. This result indicates that the board-level CDM ESD event is more critical than the chip-level CDM ESD event to IC products in field applications. The test standard on board-level CDM ESD event should be established for IC industry to verify ESD robustness of their IC products in real-world applications.

Recently, one draft of board-level CDM ESD test standard has been suggested, as shown in http://proj.moeaidb.gov.tw/sipo/files/Tec/Board-Level _CDM_Standard.pdf. We can follow this suggested board-level CDM ESD test standard to quickly de-bug the CDM ESD failures in IC products.

CHAPTER 4

CONCLUSIONS AND FUTURE WORKS

4.1   Main Results of This Thesis

The chip-level and board-level CDM ESD issues in IC products are investigated in this thesis. And based on the failure analysis result of several case studies on chip-level CDM ESD events, the possible failure or damage sites on ESD protection device during CDM ESD events are characterized and summarized.

The mechanism of board-level CDM ESD event is also introduced here. Based on this mechanism, an experiment has been performed to investigate the board-level CDM ESD current waveforms under different sizes of PCBs and charged voltages in the discharging path. Experimental results have shown that the discharging current strongly depends on the PCB size and the charged voltage. Moreover, chip-level and board-level CDM ESD levels of several test devices and test circuits fabricated in CMOS processes have been characterized and compared. Test results have confirmed that the board-level CDM ESD level of the test circuit is lower than the chip-level CDM ESD test, which indicates that the board-level CDM ESD event is more critical than the chip-level CDM ESD event. In addition, failure analysis reveals that the failure on the test circuit under board-level CDM ESD test is much severer than that under chip-level CDM ESD test.

4.2　Future Works

In nanoscale CMOS processes, the gate oxide of MOS transistor becomes thinner, which degrades the CDM ESD robustness of CMOS ICs. Moreover, the die size becomes larger in SoC applications, so more charges will be stored in the body of the chip. Consequently, CDM ESD issues, including chip-level and board-level CDM ESD events, will become more critical and should be taken into consideration in the ICs and microelectronic systems, especially when they are realized in nanoscale CMOS processes. Since the board-level CDM ESD damages are easily mistaken for EOS damages and no effective design against board-level CDM ESD events was reported so far, the formal board-level CDM ESD test standard and method should be considered and developed. Moreover, due to the threat of the board-level CDM ESD event in real-world failures, the test standard of board-level CDM ESD should be established in the near future for IC industry to verify their products.

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